

élantec

HIGH PERFORMANCE ANALOG INTEGRATED CIRCUITS

EL2041/EL2041C

Wideband, Fast Settling, Unity Gain Stable, Operational Amplifier

ELANTEC INC

T.79-07-10

Features

- Open loop unity bandwidth—90 MHz
- Unity gain stable
- High gain—10k typ.
- High slew rate—250 V/ μ s
- Low offset voltage—0.5 mV typ., 2 mV max.
- Low supply current—13 mA typ., 17 mA max.
- Wide supply operation $\pm 5V$ to $\pm 15V$
- Output voltage swing— $\pm 11V$
- Power bandwidth—4 MHz
- Fast settling time
- MIL-STD-883 Rev. C compliance
- Pin compatible with HA2541

Applications

- Pulse and video amplifiers
- Fast integrators
- Wideband filters
- High speed sample and hold circuits
- Fast, precise D/A converter output amplifier
- High speed A/D input amplifier

Ordering Information

Part No.	Temp. Range	Package	Outline #
EL2041CG	0°C to +75°C	12-Pin TO-8	MDP0002
EL2041CJ	0°C to +75°C	14-Pin CerDIP	MDP0014
EL2041G	-55°C to +125°C	12-Pin TO-8	MDP0002
EL2041G/883B	-55°C to +125°C	12-Pin TO-8	MDP0002
EL2041J	-55°C to +125°C	14-Pin CerDIP	MDP0014
EL2041J/883B	-55°C to +125°C	14-Pin CerDIP	MDP0014
EL2041CN	0°C to +75°C	8-Pin P-DIP	MDP0031

5962-87785 is the SMD version of this device.

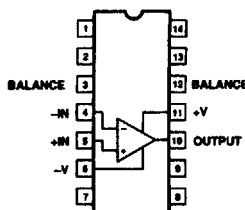
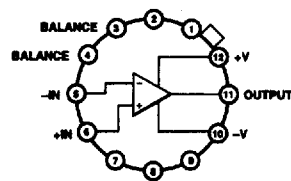
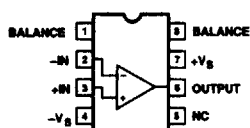
General Description

The EL2041 is a unity gain stable monolithic operational amplifier with a 90 MHz open loop unity bandwidth. This unprecedented bandwidth is accomplished with a 45° phase margin and a 6.5 dB gain margin. Unlike other wideband amplifiers, the patented EL2041 operates on standard $\pm 15V$ supplies, swings $\pm 11V$ at its output, and maintains an 80 dB open loop gain into a 1k load.

In addition, the EL2041 has a 250 V/ μ s slew rate while drawing only 13 mA of supply current. Zener Zap techniques are used to trim the offset voltage to 2 mV maximum, making the EL2041 an excellent choice for applications requiring both speed and accuracy.

Elantec's EL2041/883B complies with MIL-STD-883 Revision C in all aspects, including burn-in at 125°C. Elantec's facilities comply with MIL-I-45208A and other applicable quality specifications. For information on Elantec's military processing, see the Elantec document, QRA-2: *Elantec's Military Processing—Monolithic Products.*

Connection Diagrams

CerDIP Package**Top View****12-Lead TO-8****Top View****Plastic DIP****Top View**

Note: Non-designated pins are no connects and are not electrically connected internally.

Manufactured under U.S. Patent No. 4,837,523

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Absolute Maximum Ratings ($T_A = 25^\circ\text{C}$)

Voltage between V+ and V-	35V	Storage Temperature Range	-65°C to +150°C
Differential Input Voltage	6V	Maximum Junction Temperature	
Output Current	Continuous 25 mA	CerDIP, TO-8	175°C
	Peak 50 mA	Plastic DIP	150°C
Internal Power Dissipation	See Curves	Lead Temperature (Soldering, 5 seconds)	300°C
Operating Temperature Range			
EL2041	-55°C to +125°C		
EL2041C	0°C to +75°C		

Important Note:

All parameters having Min/Max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Inspection. Elantec performs most electrical tests using modern high-speed automatic test equipment, specifically the LTX77 Series system. Unless otherwise noted, all tests are pulsed tests, therefore $T_J = T_C = T_A$.

Test Level	Test Procedure
I	100% production tested and QA sample tested per QA test plan QCX0002.
II	100% production tested at $T_A = 25^\circ\text{C}$ and QA sample tested at $T_A = 25^\circ\text{C}$, T_{MAX} and T_{MIN} per QA test plan QCX0002.
III	QA sample tested per QA test plan QCX0002.
IV	Parameter is guaranteed (but not tested) by Design and Characterization Data.
V	Parameter is typical value at $T_A = 25^\circ\text{C}$ for information purposes only.

DC Electrical Characteristics $v_S = \pm 15\text{V}$; $R_L = 1\text{ k}\Omega$, unless otherwise specified

Parameter	Description	Temp	EL2041				EL2041C				Units
			Min	Typ	Max	Test Level	Min	Typ	Max	Test Level	
V _{OS}	Offset Voltage	+25°C		0.5	2	I		0.5	5	I	mV
		Full			10	I			10	III	mV
TCV _{OS}	Average Offset Voltage Drift	Full		20		V		20		V	$\mu\text{V}/^\circ\text{C}$
I _B	Bias Current	+25°C		5	15	I		5	15	I	μA
		Full			20	I			20	III	μA
I _{OS}	Offset Current	+25°C		1	4	I		1	4	I	μA
		Full			6	I			6	III	μA
R _{IN}	Input Resistance	+25°C		20		V		20		V	k Ω
C _{IN}	Input Capacitance	+25°C		1		V		1		V	pF
V _{CM}	Common Mode Input Range	Full	± 8	± 11		I	± 8	± 11		I	V
e _{IN}	Input Noise Voltage ($f = 1\text{ kHz}$, $R_G = 0\Omega$)	+25°C		10		V		10		V	nV/ $\sqrt{\text{Hz}}$
A _{VOL}	Large Signal Voltage Gain (Notes 1, 2)	+25°C	7k	10k		I	5k	10k		I	V/V
		Full	5k			I	4k			III	V/V
CMRR	Common-Mode Rejection Ratio (Note 3)	Full	70	80		I	60	80		II	dB
V _O	Output Voltage Swing	Full	± 11	± 12		I	± 11	± 12		II	V
I _O	Output Current (Note 11)	Full	± 25	± 50		I	± 25	± 50		I	mA
R _O	Output Resistance	+25°C		40		V		40		V	Ω
I _S	Supply Current	Full		13	17	I		13	17	II	mA
PSRR	Power Supply Rejection Ratio (Note 7)	Full	60	80		I	60	80		II	dB

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AC Electrical Characteristics $V_S = \pm 15V$; $R_L = 1\text{ k}\Omega$, unless otherwise specified

Symbol	Parameter	Temp	EL2041				EL2041C				Units
			Min	Typ	Max	Test Level	Min	Typ	Max	Test Level	
f_u	Open Loop Unity Bandwidth (Notes 4, 10)	+25°C		90		V		90		V	MHz
FPBW	Full Power Bandwidth (Notes 1, 5)	+25°C	2.8	4		I	2.8	4		I	MHz
t_r	Rise Time (Note 6)	+25°C		4		V		4		V	ns
OS	Overshoot (Note 6)	+25°C		10		V		10		V	%
SR	Slew Rate (Note 6)	+25°C	180	250		I	180	250		I	V/ μ s
t_s	Settling Time (Notes 8, 9, 10) 10V Step to 0.05%	+25°C		90		V		90		V	ns

Note 1: $V_O = \pm 10V$.Note 2: $R_L = 1\text{ k}\Omega$.Note 3: Two tests are performed. $V_{CM} = 0V$ to $+8V$ and $V_{CM} = 0V$ to $-8V$.Note 4: $V_O = 90\text{ mV}$.Note 5: Full power bandwidth guaranteed based on slew rate measurement using: $FPBW = \frac{\text{Slew Rate}}{2\pi V_{peak}}$

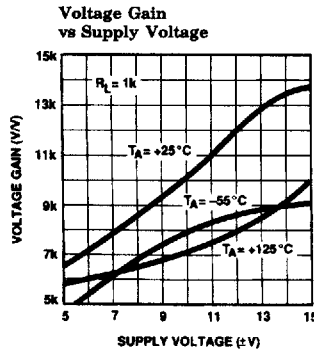
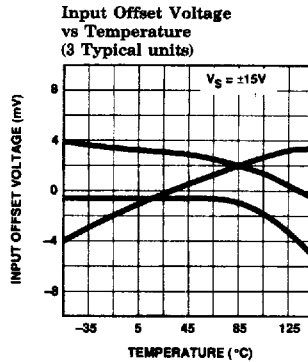
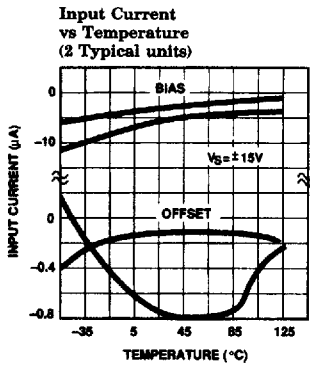
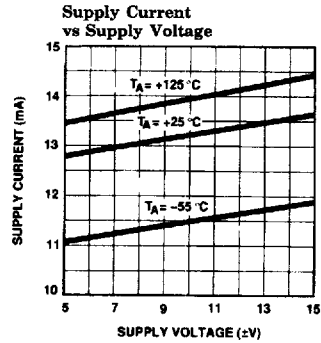
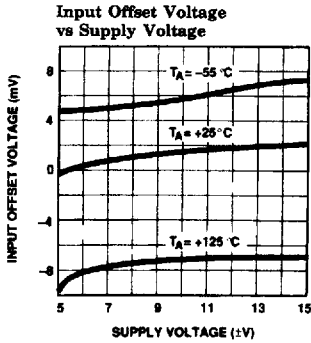
Note 6: Refer to Test Circuits section of data sheet.

Note 7: Two tests are performed. $V_+ = +15V$, and V_- is changed from $-7V$ to $-15V$. $V_- = -15V$, and V_+ changed from $+7$ to $+15V$.

Note 8: Settling time measurements are made with techniques in the following reference: "Take The Guesswork Out of Settling-Time Measurements," EDN, September 19, 1985.

Note 9: $A_V = +1$, $R_L = 1k$.Note 10: 200Ω , 20 pF output snubber, see application section.Note 11: $R_L = 200\Omega$.

Typical Performance Curves



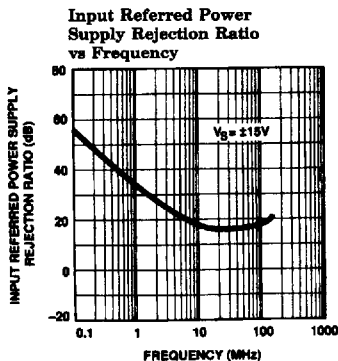
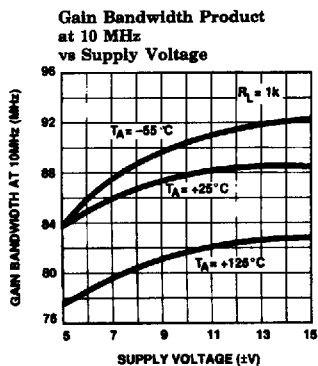
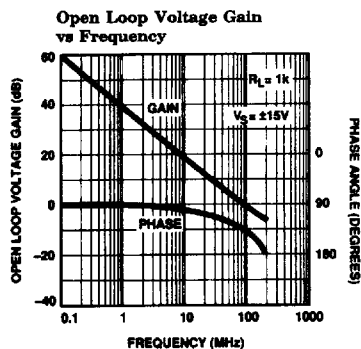
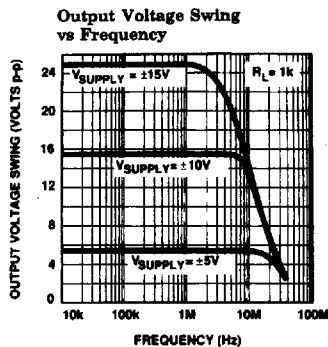
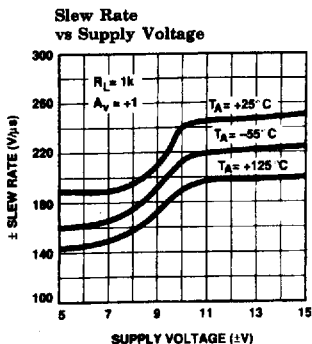
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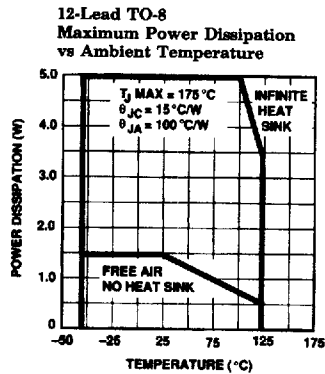
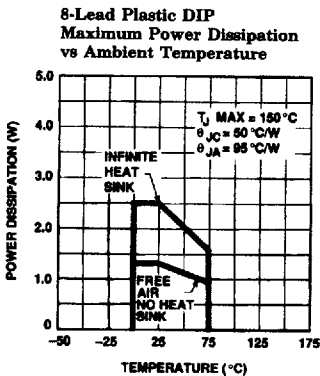
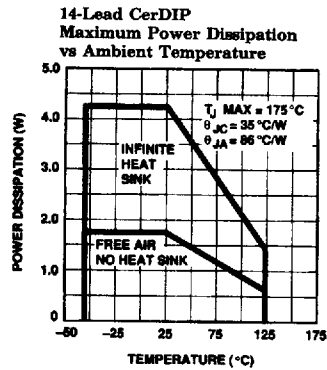
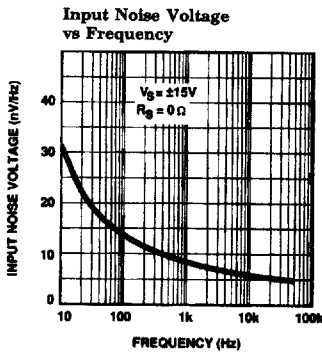
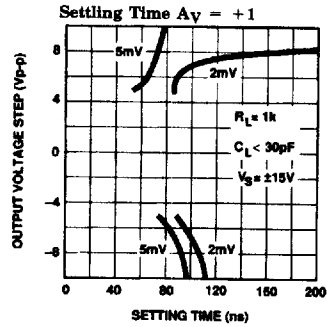
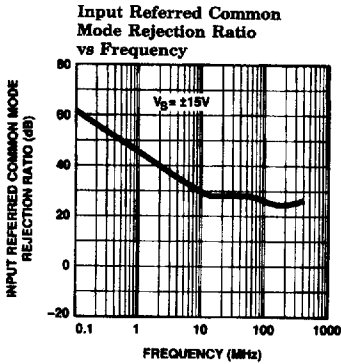
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Typical Performance Curves — Contd.



Typical Performance Curves — Contd.



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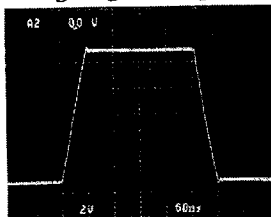
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Typical Performance Curves — Contd.

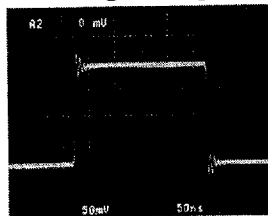
Large Signal Response



$A_V = +1$
 $V_{IN} = \pm 5V$
 $V_O = \pm 5V$

2041-7

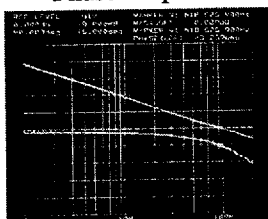
Small Signal Response



$A_V = +1$
 $V_{IN} = \pm 100 mV$
 $V_O = \pm 100 mV$

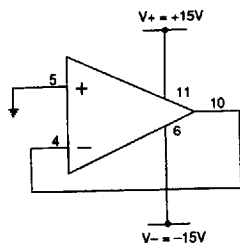
2041-8

Open Loop Gain and Phase Response



2041-9

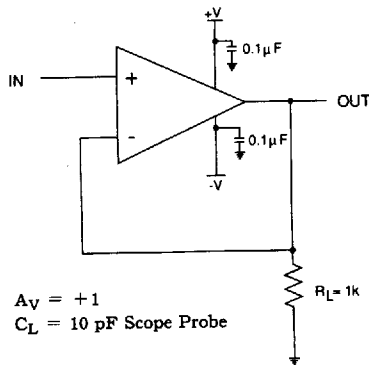
Burn-In Circuit



Pin numbers are for 14-Lead cerDIP. Burn-in circuit is identical for all package types.

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Test Circuit



$A_V = +1$
 $C_L = 10 pF$ Scope Probe

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Application Hints**Product Description**

The EL2041 is a wideband monolithic operational amplifier built on Elantec's proprietary Complementary Bipolar process. Unlike many $\pm 5V$ wideband op amps available today, the EL2041 operates from $\pm 5V$ to $\pm 15V$ and is capable of driving $\pm 11V$ at its output. The large signal swing and open loop voltage gain of 80 dB with a 1 k Ω load, differentiate the EL2041 from other op amps that do not have sufficient load isolation. Another unusual characteristic of the amplifier is the extremely wide unity gain bandwidth of 90 MHz. This bandwidth is accomplished with a 45° phase margin, a 6.5 dB gain margin, and a slew rate of 250 V/ μs . These AC characteristics are realized with a 13 mA supply current, which means lower power dissipation and higher reliability than competing products.

Power Supply Bypass

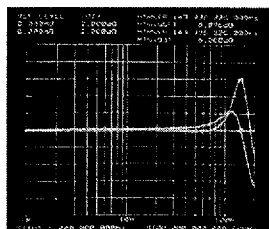
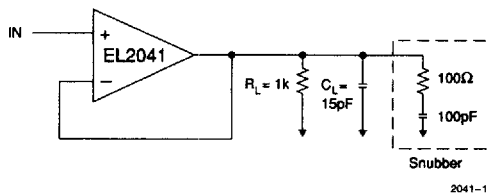
It is important to bypass the power supplies of the EL2041 with 0.1 μF or 0.01 μF ceramic disc capacitors. Failure to do this will result in oscillation or signal distortion. Although the lead length is not critical, it should not be more than $\frac{1}{2}$ " from the IC pins.

Capacitive Loading

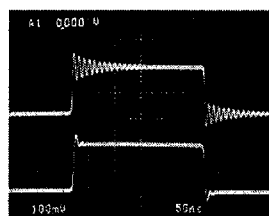
Like all high speed op amps, the EL2041 is sensitive to capacitive loading. There are at least two ways to approach this problem: The use of a snubber (Q spoiler), or the use of feedback isolation.

The first approach is to consider the output stage of the amplifier as a highly inductive element due to the application of feedback. When this output stage is loaded with a capacitance a natural resonance occurs. By putting a series RC at the output of the amplifier, the energy of the tank can be absorbed, quenching the instability. The way to select the RC values for the Q spoiler is to drive a small signal (few 100 mV) squarewave into the desired capacitive load. Place a small resistor (few 100 Ω) at the output to ground, and note the reduction in ringing. When the desired

response has been obtained, the capacitance value can be chosen. Start with a few 10's of pico farads in series with the selected resistor. Adjust the capacitor for the desired response. The capacitor value cannot be chosen arbitrarily large because of the reduction in open loop gain the series resistor will cause. In the example shown, the effects of a 15 pF load have been eliminated. Larger values of load capacitance can be tamed with a different RC value.



Frequency Response +6 dB Peak without Snubber and +2 dB with 200 Ω , 20 pF Snubber.



Top Trace is without Snubber; Bottom Trace is 100 Ω , 100 pF Snubber.

Another way to look at the effect of capacitive loading is in the frequency domain. The open loop output impedance of the EL2041 is about 40 Ω ; when the output is loaded with 15 pF, an output pole is formed at 265 MHz. This pole sounds innocent until it's realized that it causes a phase shift of $\tan^{-1} \omega RC$, and at 100 MHz that is 21°. If the amplifier has a 45°

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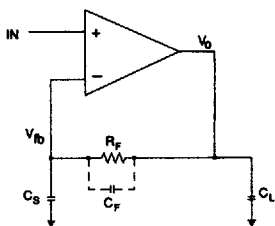
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Application Hints — Contd.

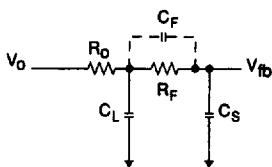
phase margin with no capacitive loading, then with 15 pF it will reduce to 24° and considerable ringing will occur. Some help can be obtained by isolating the output from the capacitance on the inverting input.

Voltage Follower with Feedback Isolation



2041-15

Equivalent Circuit for Signal, Fed Back



2041-16

The signal fed back is:

$$\frac{V_{FB}}{V_O} = \frac{1}{(1 + SC_L R_O)(1 + SR_F C_S)}$$

The situation now appears to have been made worse with an output pole and a feedback pole, but with the addition of a capacitor C_F , the effects of the stray capacitance at the inverting input can be swamped.

$$\frac{V_{FB}}{V_O} = \frac{1 + SC_F R_F}{(1 + SC_L R_O)(1 + SR_F [C_F + C_S])}$$

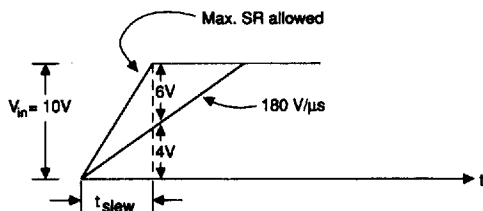
The trick here is to pick C_F large enough to overwhelm C_S and cancel the feedback pole. However C_F cannot be made too large or it will look like an AC short and C_S will again appear in parallel

with C_L . Some typical values to begin design work are: $R_F = 200$, $C_F = 15$ pF, for $C_L = 15$ pF, and C_S depends on board layout (try to minimize). It should also be realized that these values of R_F and C_F will begin to roll-off the close loop gain at 40 MHz.

Input Overdrive

It is important not to overdrive the input of the EL2041. Input slew rates in excess of 180 V/ μ s can cause distortion in the large signal square wave response, and this will show up as an increase in settling time (see typical performance curves). There are several solutions to this: Slew rate limit the input source, put clamp diodes across the amplifier inputs, or take some voltage gain in the amplifier.

Slew rate limit the input: For example with a 10 V_{p-p} step at the input, the input rate should be limited to:



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$$t_{SLEW} = \frac{V_{IN} - V_{ZENER}}{\min SR}$$

$$\text{Max source SR} = \frac{V_{IN}}{t_{SLEW}}$$

or

$$\frac{10 - 6}{180 \text{ V}/\mu\text{s}} = 22 \text{ ns} \quad , \quad \text{Max SR} = \frac{10}{22 \text{ ns}} = 450 \text{ V}/\mu\text{s}$$

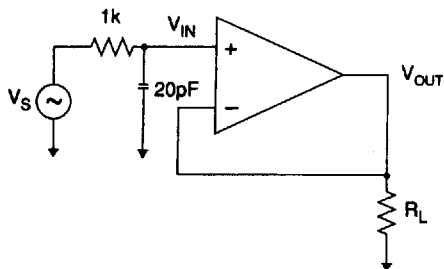
If the input slew rate is limited by a 1k resistor, how large a capacitor is needed?

$$\frac{10\text{V}}{1\text{k}} = 10 \text{ mA} = C \frac{dv}{dt} \quad , \quad C = \frac{10 \text{ mA}}{450 \text{ V}/\mu\text{s}} = 22 \text{ pF}$$

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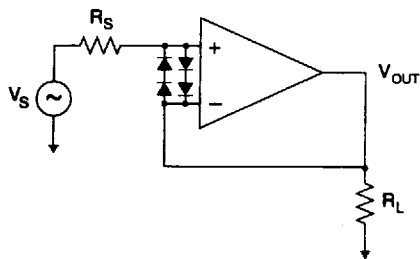
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Application Hints — Contd.

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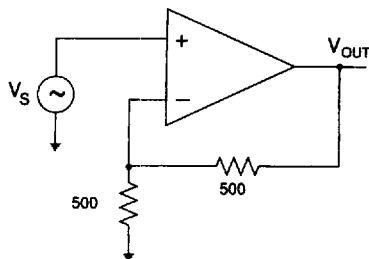
This value of R and C will give a -3 dB bandwidth of 8 MHz through the op amp. This technique should be avoided if the intended use is a small signal sinewave application.

Clamp diodes across the inputs: To obtain full slew rate at elevated temperature requires a V_{BE} of overdrive across the inputs. To insure adequate protection and slew rate requires two diodes in each direction across the inputs. A small series resistance in the input will limit the current through the diodes.



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Take voltage gain in the op amp: By taking voltage gain, the input stage does not have to handle as large a signal swing for a given output swing. For a voltage gain of 2, remember that the closed loop bandwidth will go to 45 MHz.

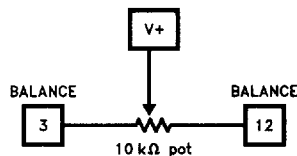


2041-20

Using the BALANCE Pins on the EL2041 Operational Amplifier

The BALANCE pins on the EL2041 can be used to tune out or adjust the input offset voltage of the op amp. To use these pins, ignore the connections shown on the simplified schematic on page 1-171. The adjust current is mirrored up to pnp current sources near $V+$ and the BALANCE adjustment pot goes between the collectors of two same-sex transistors. Take a 10 k Ω potentiometer (or lower) and connect the ends of the pot to the BALANCE pins and then connect the wiper to $+V$ as shown (see Figure below). Moving the wiper between the two values should zero out the offset voltage.

In hooking up the example test circuit and measuring the voltage between pins 4 and 5 of an EL2041J, with supply voltages of ± 15 V and a 10 k Ω pot between pins 3 and 12, an example adjustment range of $+10.3$ mV to -13.4 mV for V_{OS} was measured. The adjust range can be increased by lowering the size of the trim pot and decreased by increasing it. Adjustment range will vary slightly from part to part.



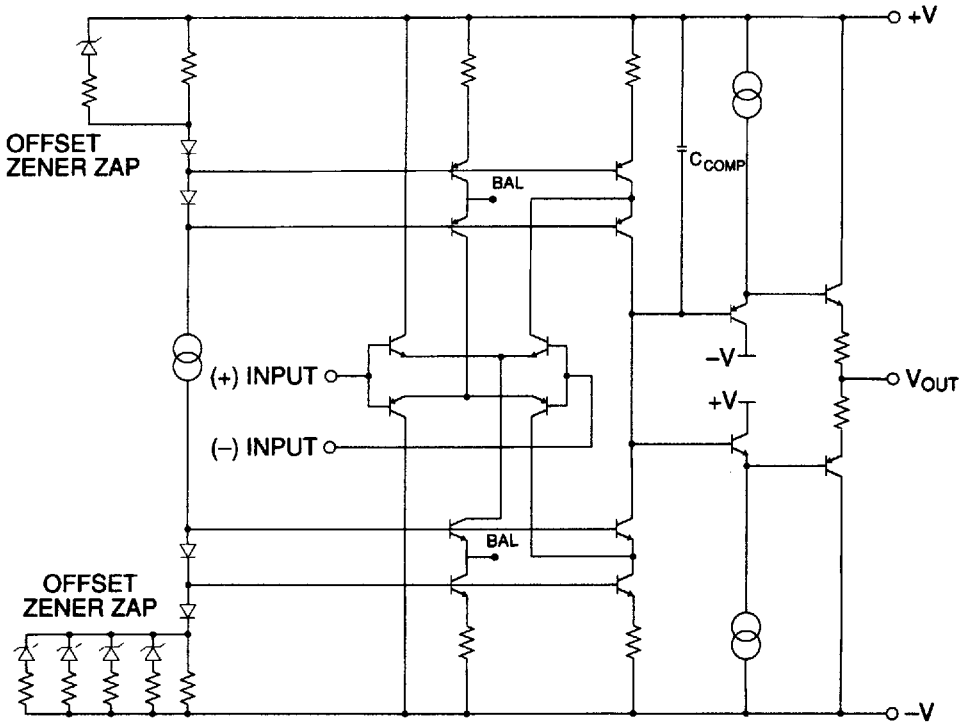
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Simplified Schematic



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EL2041/EL2041C*Wideband, Fast Settling, Unity Gain Stable, Operational Amplifier*

EL2041/EL2041C

EL2041 Macromodel

```

* Connections:  + input
*               |      - input
*               |      |      + Vsupply
*               |      |      |      - Vsupply
*               |      |      |      |      output
*               |      |      |      |      |
.subckt M2041  5  4  11  6  10

```

* Input stage

```

ie 37 6 3.7mA
r6 36 37 75
r7 38 37 75
rc1 11 30 75
rc2 11 39 75
q1 30 5 36 qn
q2 39 4 38 qna
ediff 33 0 39 30 3
rdiff 33 0 1Meg

```

* Compensation Section

```

ga 0 34 33 0 5.2m
rh 34 0 1Meg
ch 34 0 16pF
rc 34 40 300
cc 40 0 1.5pF

```

* Poles

```

ep 41 0 40 0 1
rpa 41 42 75
cpa 42 0 2pF
rpb 42 43 50
cpb 43 0 3pF

```

* Output Stage

```

ios1 11 50 1.25mA
ios2 51 6 1.25mA
q3 6 43 50 qp
q4 11 43 51 qn
q5 11 50 52 qn
q6 6 51 53 qp
ros1 52 10 25
ros2 10 53 25

```

* Power Supply Current

```

ips 11 6 6.5mA

```

* Models

```

.model qn npn(is=800.0E-18 bf=340 tf=0.2nS)
.model qna npn(is=864E-18 bf=400 tf=0.2nS)
.model qp pnp(is=800E-18 bf=60 tf=0.2nS)
.ends

```

1

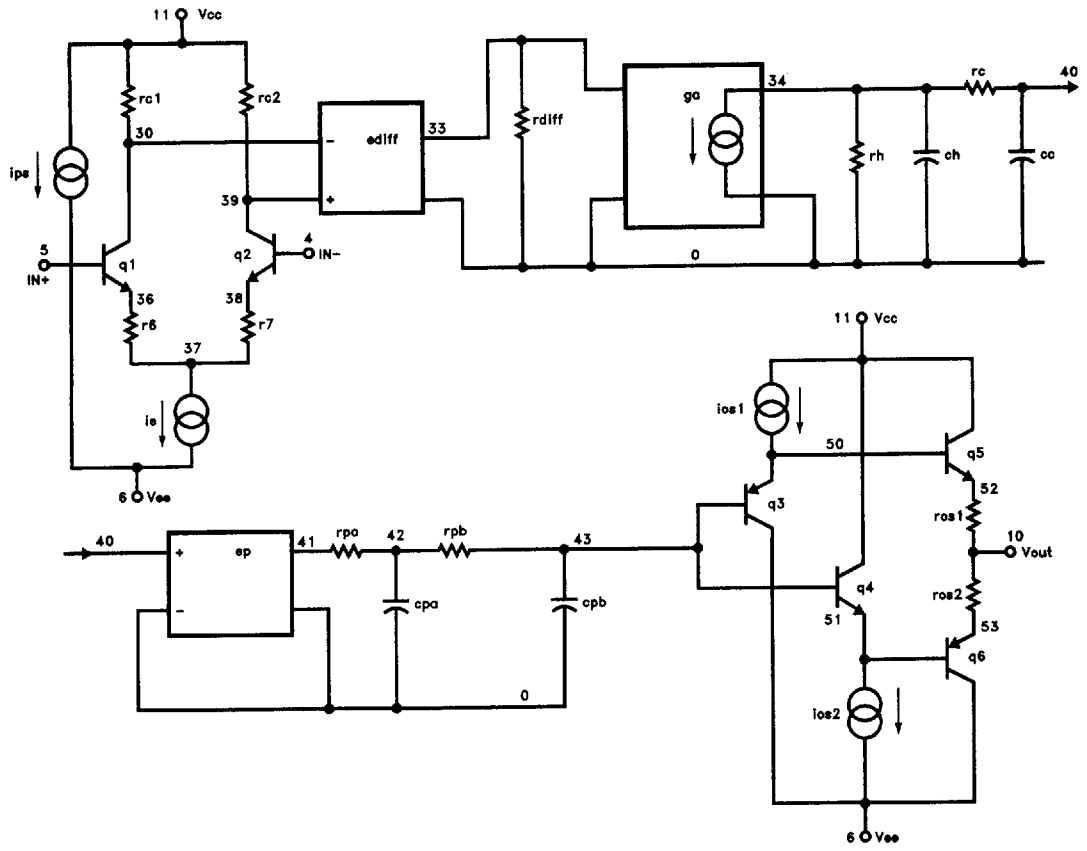
EL2041/EL2041C

EL2041/EL2041C

ELANTEC INC

Wideband, Fast Settling, Unity Gain Stable, Operational Amplifier

EL2041 Macromodel — Contd.



2041-22