élantec
HIGH PERFORMANCE ANALOG INTEGRATED CIRCUITS

EL5210C/EL5410C

30MHz Rail-to-Rail Input-Output Op Amps

Features

- 30MHz -3dB bandwidth
- Supply voltage = 4.5V to 16.5V
- Low supply current (per amplifier) = 2.5mA
- High slew rate = 33V/µs
- Unity-gain stable
- Beyond the rails input capability
- Rail-to-rail output swing
- Available in both standard and space-saving fine pitch packages

Applications

- Driver for A-to-D Converters
- Data Acquisition
- Video Processing
- Audio Processing
- Active Filters
- Test Equipment
- Battery Powered Applications
- Portable Equipment

Ordering Information

Part No.	Package	Tape & Reel	Outline #
EL5210CS	8-Pin SOIC	-	MDP0027
EL5210CS-T13	8-Pin SOIC	13”	MDP0027
EL5210CY	8-Pin MSOP	-	MDP0043
EL5210CY-T7	8-Pin MSOP	7”	MDP0043
EL5210CY-T13	8-Pin MSOP	13”	MDP0043
EL5410CS	14-Pin SOIC	-	MDP0027
EL5410CS-T13	14-Pin SOIC	13”	MDP0027
EL5410CR	14-Pin TSSOP	-	MDP0044
EL5410CR-T13	14-Pin TSSOP	13”	MDP0044

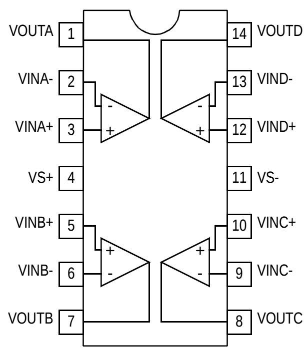
General Description

The EL5210C and EL5410C are low power, high voltage rail-to-rail input-output amplifiers. The EL5210C contains two amplifiers in one package and the EL5410C contains four amplifiers. Operating on supplies ranging from 5V to 15V, while consuming only 2.5mA per amplifier, the EL5410C and EL5210C have a bandwidth of 30MHz (-3dB). They also provide common mode input ability beyond the supply rails, as well as rail-to-rail output capability. This enables these amplifiers to offer maximum dynamic range at any supply voltage.

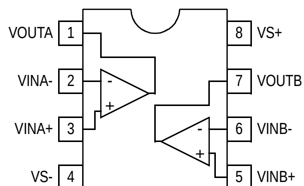
The EL5410C and EL5210C also feature fast slewing and settling times, as well as a high output drive capability of 30mA (sink and source). These features make these amplifiers ideal for high speed filtering and signal conditioning application. Other applications include battery power, portable devices, and anywhere low power consumption is important.

The EL5410C is available in a space-saving 14-Pin TSSOP package, as well as the industry-standard 14-Pin SOIC. The EL5210C is available in the 8-Pin MSOP and 8-Pin SOIC packages. Both feature a standard operational amplifier pin out. These amplifiers operate over a temperature range of -40°C to +85°C.

Connection Diagram



EL5410C (TSSOP-14, SOIC-14)



EL5210C (MSOP-8, SOIC-8)

Note: All information contained in this data sheet has been carefully checked and is believed to be accurate as of the date of publication; however, this data sheet cannot be a “controlled document”. Current revisions, if any, to these specifications are maintained at the factory and are available upon your request. We recommend checking the revision level before finalization of your design documentation.

EL5210C/EL5410C

30MHz Rail-to-Rail Input-Output Op Amps

Absolute Maximum Ratings (T_A = 25°C)

Values beyond absolute maximum ratings can cause the device to be prematurely damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

Supply Voltage between V_S+ and V_S-	+18V
Input Voltage	V_S- - 0.5V, V_S +0.5V
Maximum Continuous Output Current	30mA
Maximum Die Temperature	+125°C
Storage Temperature	-65°C to +150°C
Operating Temperature	-40°C to +85°C
Power Dissipation	See Curves
ESD Voltage	2kV

Important Note:

All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Characteristics

V_S+ = +5V, V_S - = -5V, R_L = 1kΩ and C_L = 12pF to 0V, T_A = 25°C unless otherwise specified.

Parameter	Description	Condition	Min	Typ	Max	Unit
Input Characteristics						
V_{OS}	Input Offset Voltage	V_{CM} = 0V		3	15	mV
TCV_{OS}	Average Offset Voltage Drift [1]			7		µV/°C
I_B	Input Bias Current	V_{CM} = 0V		2	60	nA
R_{IN}	Input Impedance			1		GΩ
C_{IN}	Input Capacitance			2		pF
CMIR	Common-Mode Input Range		-5.5		+5.5	V
CMRR	Common-Mode Rejection Ratio	for V_{IN} from -5.5V to 5.5V	50	70		dB
A_{VOL}	Open-Loop Gain	-4.5V ≤ V_{OUT} ≤ 4.5V	65	80		dB
Output Characteristics						
V_{OL}	Output Swing Low	I_L = -5mA		-4.9	-4.8	V
V_{OH}	Output Swing High	I_L = 5mA	4.8	4.9		V
I_{SC}	Short Circuit Current			±120		mA
I_{OUT}	Output Current			±30		mA
Power Supply Performance						
PSRR	Power Supply Rejection Ratio	V_S is moved from ±2.25V to ±7.75V	60	80		dB
I_S	Supply Current (Per Amplifier)	No Load		2.5	3.75	mA
Dynamic Performance						
SR	Slew Rate [2]	-4.0V ≤ V_{OUT} ≤ 4.0V, 20% o 80%		33		V/µs
t_S	Settling to +0.1% (A_V = +1)	(A_V = +1), V_O = 2V Step		140		ns
BW	-3dB Bandwidth			30		MHz
GBWP	Gain-Bandwidth Product			20		MHz
PM	Phase Margin			50		°
CS	Channel Separation	f = 5MHz		110		dB
d_G	Differential Gain [3]	$R_F = R_G$ = 1kΩ and V_{OUT} = 1.4V		0.12		%
d_P	Differential Phase[3]	$R_F = R_G$ = 1kΩ and V_{OUT} = 1.4V		0.17		°

1. Measured over operating temperature range
2. Slew rate is measured on rising and falling edges
3. NTSC signal generator used

Electrical Characteristics

V_S+ = 5V, V_S- = 0V, R_L = 1kΩ and C_L = 12pF to 2.5V, T_A = 25°C unless otherwise specified.

Parameter	Description	Condition	Min	Typ	Max	Unit
Input Characteristics						
V_{OS}	Input Offset Voltage	V_{CM} = 2.5V		3	15	mV
TCV_{OS}	Average Offset Voltage Drift [1]			7		µV/°C
I_B	Input Bias Current	V_{CM} = 2.5V		2	60	nA
R_{IN}	Input Impedance			1		GΩ
C_{IN}	Input Capacitance			2		pF
CMIR	Common-Mode Input Range		-0.5		+5.5	V
CMRR	Common-Mode Rejection Ratio	for V_{IN} from -0.5V to 5.5V	45	66		dB
A_{VOL}	Open-Loop Gain	0.5V ≤ V_{OUT} ≤ 4.5V	65	80		dB
Output Characteristics						
V_{OL}	Output Swing Low	I_L = -5mA		100	200	mV
V_{OH}	Output Swing High	I_L = 5mA	4.8	4.9		V
I_{SC}	Short Circuit Current			±120		mA
I_{OUT}	Output Current			±30		mA
Power Supply Performance						
PSRR	Power Supply Rejection Ratio	V_S is moved from 4.5V to 15.5V	60	80		dB
I_S	Supply Current (Per Amplifier)	No Load		2.5	3.75	mA
Dynamic Performance						
SR	Slew Rate [2]	1V ≤ V_{OUT} ≤ 4V, 20% o 80%		33		V/µs
t_S	Settling to +0.1% (A_V = +1)	(A_V = +1), V_O = 2V Step		140		ns
BW	-3dB Bandwidth			30		MHz
GBWP	Gain-Bandwidth Product			20		MHz
PM	Phase Margin			50		°
CS	Channel Separation	f = 5MHz		110		dB
d_G	Differential Gain [3]	R_F = R_G = 1kΩ and V_{OUT} = 1.4V		0.30		%
d_P	Differential Phase[3]	R_F = R_G = 1kΩ and V_{OUT} = 1.4V		0.66		°

1. Measured over operating temperature range
2. Slew rate is measured on rising and falling edges
3. NTSC signal generator used

EL5210C/EL5410C

30MHz Rail-to-Rail Input-Output Op Amps

Electrical Characteristics

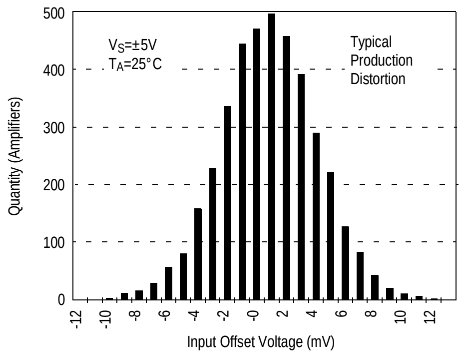
V_S+ = 15V, V_S- = 0V, R_L = 1kΩ and C_L = 12pF to 7.5V, T_A = 25°C unless otherwise specified.

Parameter	Description	Condition	Min	Typ	Max	Unit
Input Characteristics						
V_{OS}	Input Offset Voltage	V_{CM} = 7.5V		3	15	mV
TCV_{OS}	Average Offset Voltage Drift [1]			7		µV/°C
I_B	Input Bias Current	V_{CM} = 7.5V		2	60	nA
R_{IN}	Input Impedance			1		GΩ
C_{IN}	Input Capacitance			2		pF
CMIR	Common-Mode Input Range		-0.5		+15.5	V
CMRR	Common-Mode Rejection Ratio	for V_{IN} from -0.5V to 15.5V	53	72		dB
A_{VOL}	Open-Loop Gain	0.5V ≤ V_{OUT} ≤ 14.5V	65	80		dB
Output Characteristics						
V_{OL}	Output Swing Low	I_L = -7.5mA		170	350	mV
V_{OH}	Output Swing High	I_L = 7.5mA	14.65	14.83		V
I_{SC}	Short Circuit Current			±120		mA
I_{OUT}	Output Current			±30		mA
Power Supply Performance						
PSRR	Power Supply Rejection Ratio	V_S is moved from 4.5V to 15.5V	60	80		dB
I_S	Supply Current (Per Amplifier)	No Load		2.5	3.75	mA
Dynamic Performance						
SR	Slew Rate [2]	1V ≤ V_{OUT} ≤ 14V, 20% o 80%		33		V/µs
t_S	Settling to +0.1% (A_V = +1)	(A_V = +1), V_O = 2V Step		140		ns
BW	-3dB Bandwidth			30		MHz
GBWP	Gain-Bandwidth Product			20		MHz
PM	Phase Margin			50		°
CS	Channel Separation	f = 5MHz		110		dB
d_G	Differential Gain [3]	R_F = R_G = 1kΩ and V_{OUT} = 1.4V		0.10		%
d_P	Differential Phase[3]	R_F = R_G = 1kΩ and V_{OUT} = 1.4V		0.11		°

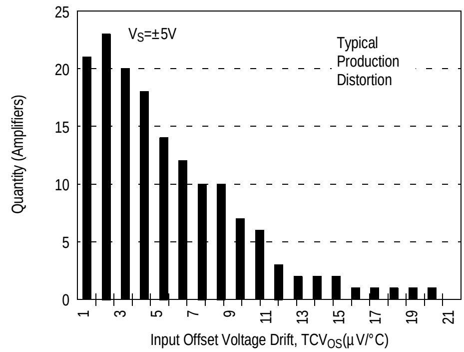
1. Measured over operating temperature range
2. Slew rate is measured on rising and falling edges
3. NTSC signal generator used

Typical Performance Curves

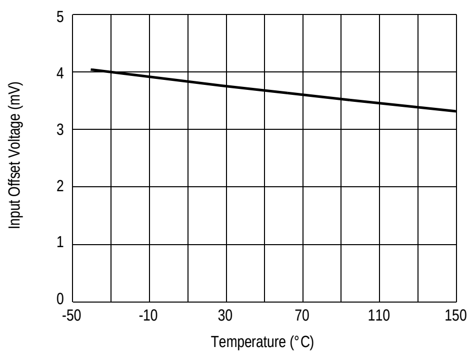
EL5410C Input Offset Voltage Distribution



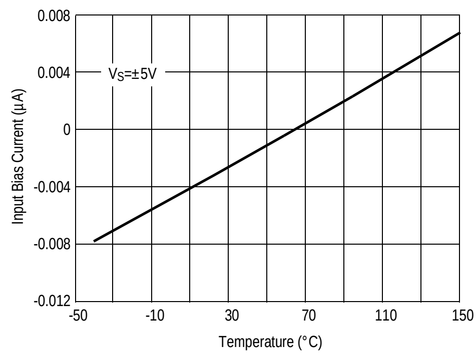
EL5410C Input Offset Voltage Drift



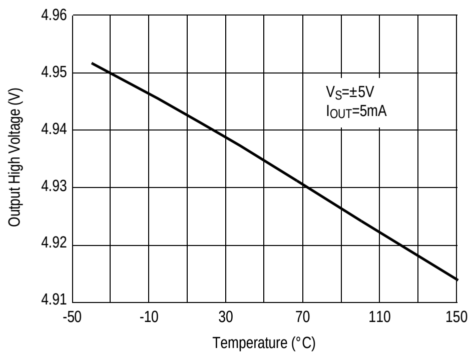
Input Offset Voltage vs Temperature



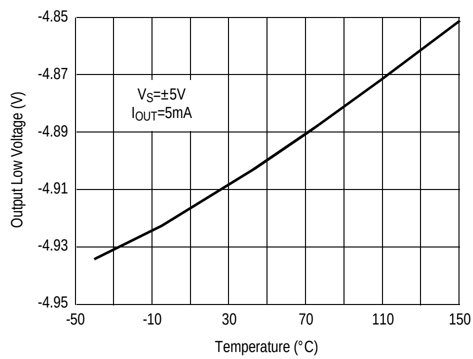
Input Bias Current vs Temperature



Output High Voltage vs Temperature

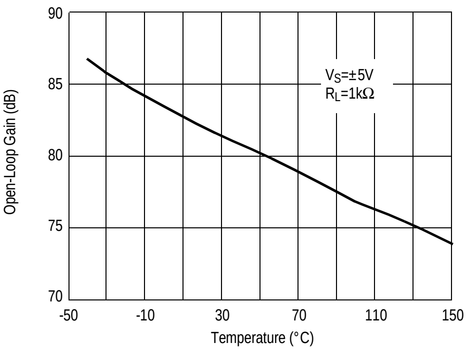


Output Low Voltage vs Temperature

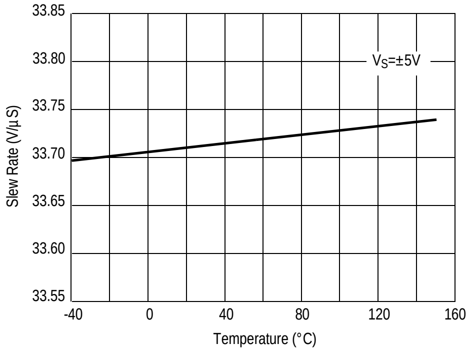


Typical Performance Curves

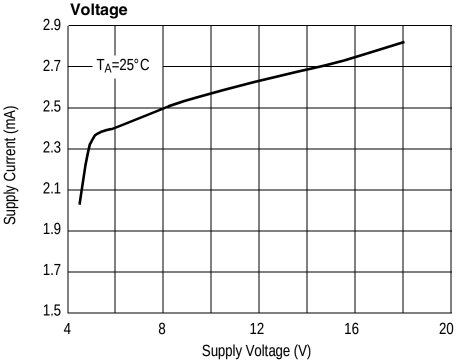
Open-Loop Gain vs Temperature



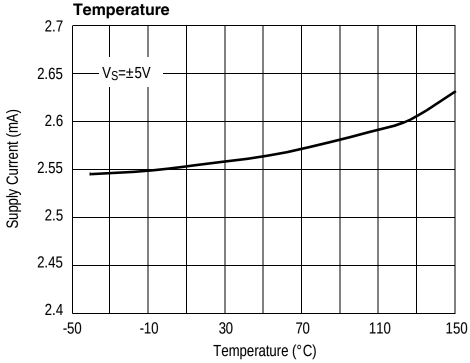
Slew Rate vs Temperature



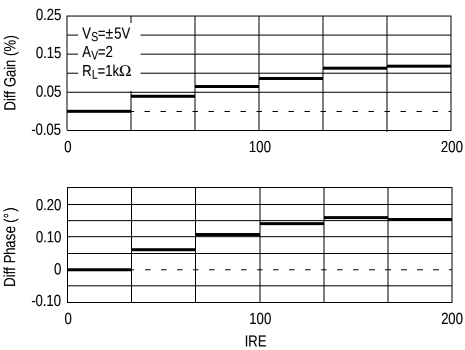
EL5410C Supply Current per Amplifier vs Supply Voltage



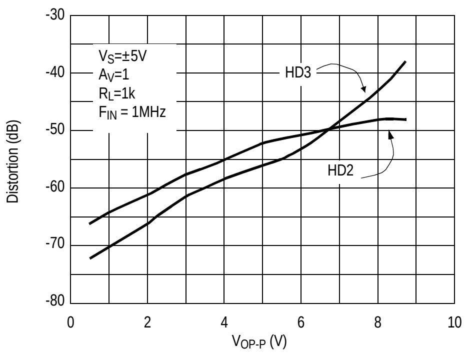
EL5410C Supply Current per Amplifier vs Temperature



Differential Gain and Phase



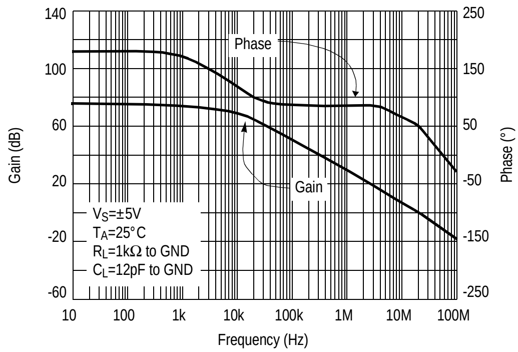
Harmonic Distortion vs V_{OP-P}



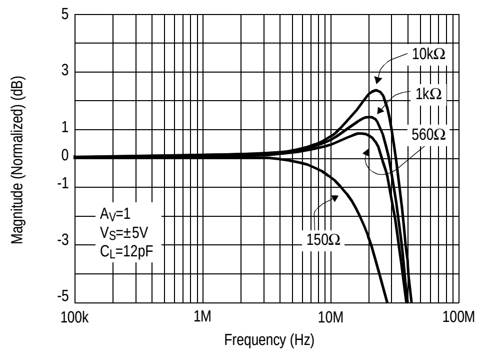
EL5210C/EL5410C

Typical Performance Curves

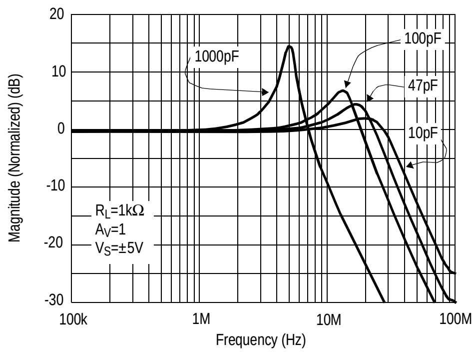
Open Loop Gain and Phase vs Frequency



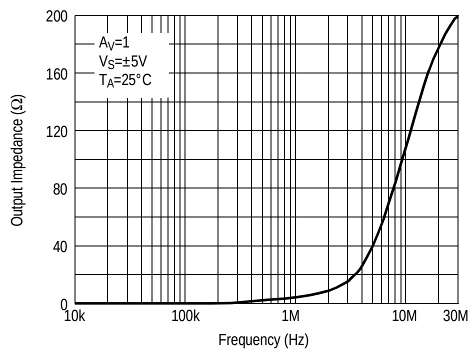
Frequency Response for Various R_L



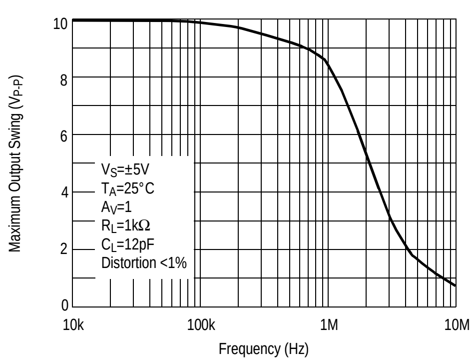
Frequency Response for Various C_L



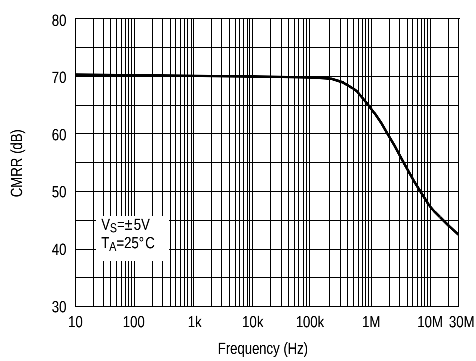
Closed Loop Output Impedance vs Frequency



Maximum Output Swing vs Frequency

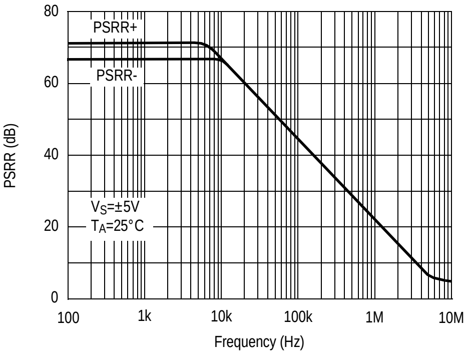


CMRR vs Frequency

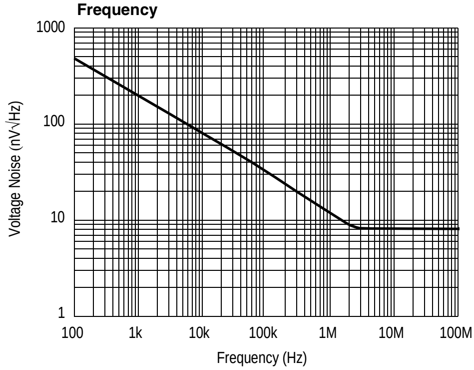


Typical Performance Curves

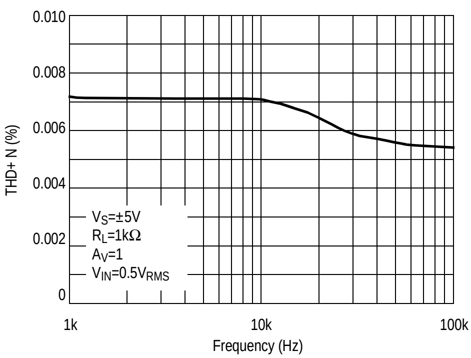
PSRR vs Frequency



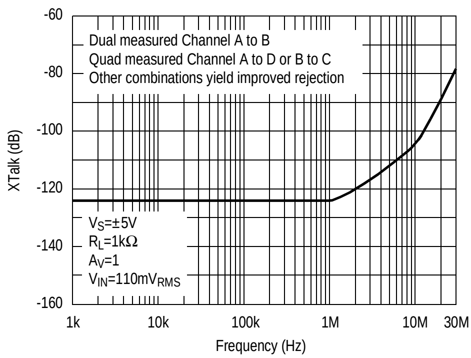
Input Voltage Noise Spectral Density vs Frequency



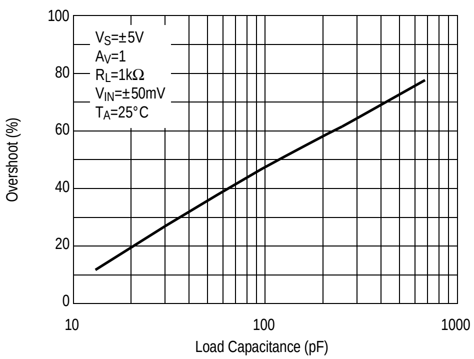
Total Harmonic Distortion + Noise vs Frequency



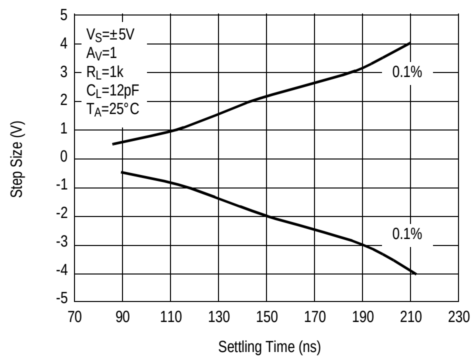
Channel Separation vs Frequency Response



Small-Signal Overshoot vs Load Capacitance

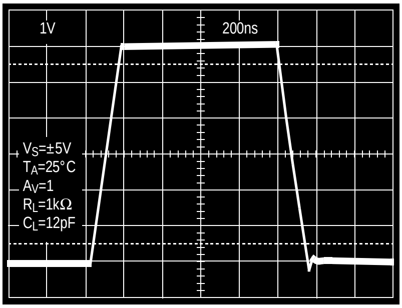


Settling Time vs Step Size

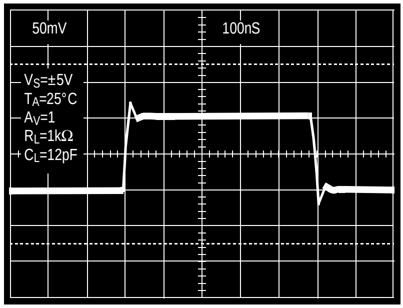


Typical Performance Curves

Large Signal Transient Response



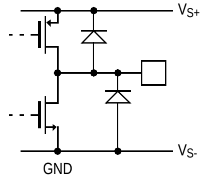
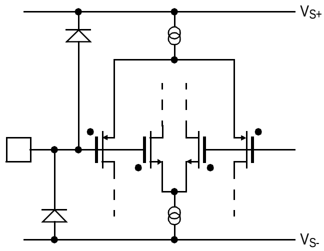
Small Signal Transient Response



EL5210C/EL5410C

30MHz Rail-to-Rail Input-Output Op Amps

Pin Descriptions

EL5210C	EL5410C	Name	Function	Equivalent Circuit
1	1	V_{OUTA}	Amplifier A Output	 **Circuit 1**
2	2	V_{INA-}	Amplifier A Inverting Input	 **Circuit 2**
3	3	V_{INA+}	Amplifier A Non-Inverting Input	(Reference Circuit 2)
8	4	V_{S+}	Positive Power Supply	
5	5	V_{INB+}	Amplifier B Non-Inverting Input	(Reference Circuit 2)
6	6	V_{INB-}	Amplifier B Inverting Input	(Reference Circuit 2)
7	7	V_{OUTB}	Amplifier B Output	(Reference Circuit 1)
	8	V_{OUTC}	Amplifier C Output	(Reference Circuit 1)
	9	V_{INC-}	Amplifier C Inverting Input	(Reference Circuit 2)
	10	V_{INC+}	Amplifier C Non-Inverting Input	(Reference Circuit 2)
4	11	V_{S-}	Negative Power Supply	
	12	V_{IND+}	Amplifier D Non-Inverting Input	(Reference Circuit 2)
	13	V_{IND-}	Amplifier D Inverting Input	(Reference Circuit 2)
	14	V_{OUTD}	Amplifier D Output	(Reference Circuit 1)

Applications Information

Product Description

The EL5210C and EL5410C voltage feedback amplifiers are fabricated using a high voltage CMOS process. They exhibit Rail-to-Rail input and output capability, are unity gain stable and have low power consumption (2.5mA per amplifier). These features make the EL5210C and EL5410C ideal for a wide range of general-purpose applications. Connected in voltage follower mode and driving a load of 1kΩ and 12pF, the EL5210C and EL5410C have a -3dB bandwidth of 30MHz while maintaining a 33V/µS slew rate. The EL5210C is a dual amplifier while the EL5410C is a quad amplifier.

Operating Voltage, Input, and Output

The EL5210C and EL5410C are specified with a single nominal supply voltage from 5V to 15V or a split supply with its total range from 5V to 15V. Correct operation is guaranteed for a supply range of 4.5V to 16.5V. Most EL5210C and EL5410C specifications are stable over both the full supply range and operating temperatures of -40 °C to +85 °C. Parameter variations with operating voltage and/or temperature are shown in the typical performance curves.

The input common-mode voltage range of the EL5210C and EL5410C extends 500mV beyond the supply rails. The output swings of the EL5210C and EL5410C typically extend to within 100mV of positive and negative supply rails with load currents of 5mA. Decreasing load currents will extend the output voltage range even closer to the supply rails. Figure 1 shows the input and output waveforms for the device in the unity-gain configuration. Operation is from +/-5V supply with a 1kΩ load connected to GND. The input is a 10Vp-p sinusoid. The output voltage is approximately $9.8V_{P-P}$.

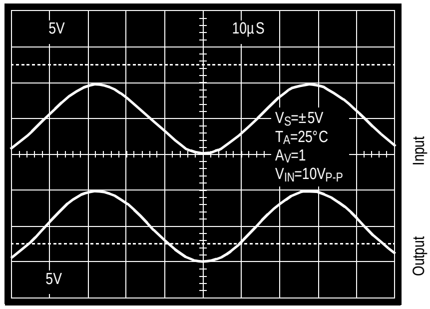


Figure 1. Operation with Rail-to-Rail Input and Output

Short Circuit Current Limit

The EL5210C and EL5410C will limit the short circuit current to +/-120mA if the output is directly shorted to the positive or the negative supply. If an output is shorted indefinitely, the power dissipation could easily increase such that the device may be damaged. Maximum reliability is maintained if the output continuous current never exceeds +/-30mA. This limit is set by the design of the internal metal interconnects.

Output Phase Reversal

The EL5210C and EL5410C are immune to phase reversal as long as the input voltage is limited from V_S- -0.5V to V_S+ +0.5V. Figure 2 shows a photo of the output of the device with the input voltage driven beyond the supply rails. Although the device's output will not change phase, the input's overvoltage should be avoided. If an input voltage exceeds supply voltage by more than 0.6V, electrostatic protection diodes placed in the input stage of the device begin to conduct and overvoltage damage could occur.

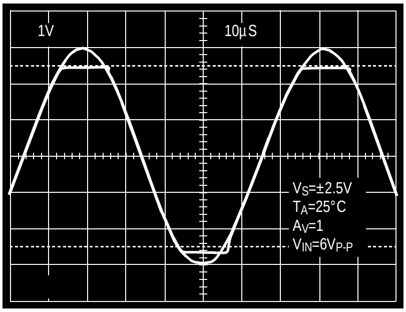


Figure 2. Operation with Beyond-the-Rails Input

Power Dissipation

With the high-output drive capability of the EL5210C and EL5410C amplifiers, it is possible to exceed the 125°C 'absolute-maximum junction temperature' under certain load current conditions. Therefore, it is important to calculate the maximum junction temperature for the application to determine if load conditions need to be modified for the amplifier to remain in the safe operating area.

The maximum power dissipation allowed in a package is determined according to:

$$P_{DMAX} = \frac{T_{JMAX} - T_{AMAX}}{\Theta_{JA}}$$

Where:

T_{JMAX} = Maximum Junction Temperature

T_{AMAX}= Maximum Ambient Temperature

Θ_{JA} = Thermal Resistance of the Package

P_{DMAX} = Maximum Power Dissipation in the Package.

The maximum power dissipation actually produced by an IC is the total quiescent supply current times the total power supply voltage, plus the power in the IC due to the loads, or:

$$P_{DMAX} = \Sigma i[V_S \times I_{SMAX} + (V_S + - V_{OUT} i) \times I_{LOAD} i]$$

when sourcing, and

$$P_{DMAX} = \Sigma i[V_S \times I_{SMAX} + (V_{OUT} i - V_S -) \times I_{LOAD} i]$$

when sinking.

Where:

i = 1 to 2 for Dual and 1 to 4 for Quad

V_S = Total Supply Voltage

I_{SMAX} = Maximum Supply Current Per Amplifier

$V_{OUT}i$ = Maximum Output Voltage of the Application

$I_{LOAD}i$ = Load current

If we set the two P_{DMAX} equations equal to each other, we can solve for $R_{LOAD}i$ to avoid device overheat. Figure 3 and Figure 4 provide a convenient way to see if the device will overheat. The maximum safe power dissipation can be found graphically, based on the package type and the ambient temperature. By using the previous equation, it is a simple matter to see if P_{DMAX} exceeds the device's power derating curves. To ensure proper operation, it is important to observe the recommended derating curves shown in Figure 3 and Figure 4.

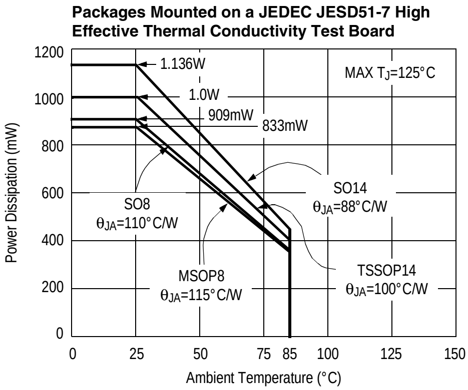


Figure 3. Package Power Dissipation vs Ambient Temperature

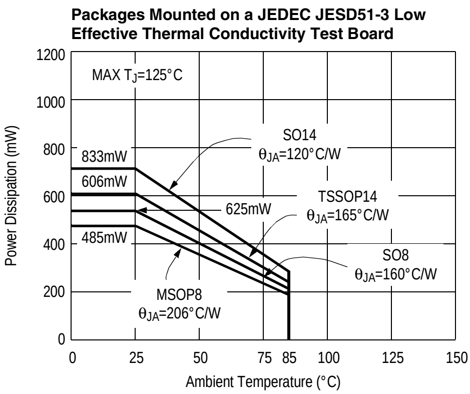


Figure 4. Package Power Dissipation vs Ambient Temperature

Unused Amplifiers

It is recommended that any unused amplifiers in a dual and a quad package be configured as a unity gain follower. The inverting input should be directly connected to the output and the non-inverting input tied to the ground plane.

Driving Capacitive Loads

The EL5210C and EL5410C can drive a wide range of capacitive loads. As load capacitance increases, however, the -3dB bandwidth of the device will decrease and the peaking increase. The amplifiers drive 10pF loads in parallel with 1kΩ with just 1.2dB of peaking, and 100pF with 6.5dB of peaking. If less peaking is desired in these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output. However, this will obviously reduce the gain slightly. Another method of reducing peaking is to add a "snubber" circuit at the output. A snubber is a shunt load consisting of a resistor in series with a capacitor. Values of 150Ω and 10nF are typical. The advantage of a snubber is that it does not draw any DC load current or reduce the gain

Power Supply Bypassing and Printed Circuit Board Layout

The EL5210C and EL5410C can provide gain at high frequency. As with any high-frequency device, good printed circuit board layout is necessary for optimum performance. Ground plane construction is highly recommended, lead lengths should be as short as possible and the power supply pins must be well bypassed to reduce the risk of oscillation. For normal single supply operation, where the V_S- pin is connected to ground, a 0.1µF ceramic capacitor should be placed from V_S+ to pin to V_S- pin. A 4.7µF tantalum capacitor should then be connected in parallel, placed in the region of the amplifier. One 4.7µF capacitor may be used for multiple devices. This same capacitor combination should be placed at each supply pin to ground if split supplies are to be used.

EL5210C/EL5410C

30MHz Rail-to-Rail Input-Output Op Amps

General Disclaimer

Specifications contained in this data sheet are in effect as of the publication date shown. Elantec, Inc. reserves the right to make changes in the circuitry or specifications contained herein at any time without notice. Elantec, Inc. assumes no responsibility for the use of any circuits described herein and makes no representations that they are free from patent infringement.

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HIGH PERFORMANCE ANALOG INTEGRATED CIRCUITS

Elantec Semiconductor, Inc.
675 Trade Zone Blvd.
Milpitas, CA 95035
Telephone: (408) 945-1323
(888) ELANTEC
Fax: (408) 945-9305
European Office: +44-118-977-6080
Japan Technical Center: +81-45-682-5820

WARNING - Life Support Policy

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