



# BUH615D

## HIGH VOLTAGE FAST-SWITCHING NPN POWER TRANSISTOR

- STMicroelectronics PREFERRED SALESTYPE
- HIGH VOLTAGE CAPABILITY
- NPN TRANSISTOR WITH INTEGRATED FREEWHEELING DIODE
- U.L. RECOGNISED ISOWATT218 PACKAGE (U.L. FILE # E81734 (N))

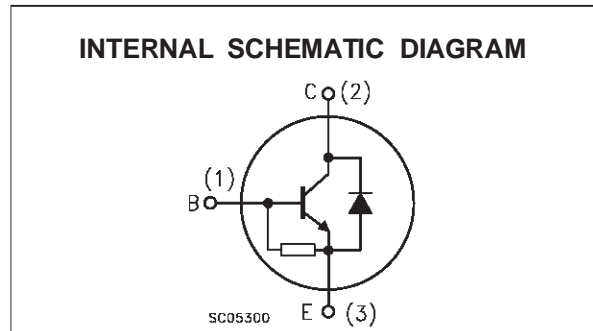
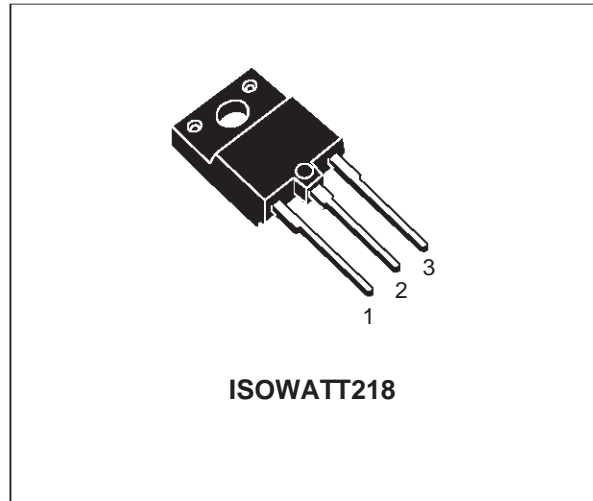
### APPLICATIONS:

- HORIZONTAL DEFLECTION FOR COLOUR TV

### DESCRIPTION

The BUH615D is manufactured using Multiepitaxial Mesa technology for cost-effective high performance and uses a Hollow Emitter structure to enhance switching speeds.

The BUH series is designed for use in horizontal deflection circuits in televisions and monitors.



### ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
$V_{CBO}$	Collector-Base Voltage ( $I_E = 0$ )	1500	V
$V_{CEO}$	Collector-Emitter Voltage ( $I_B = 0$ )	700	V
$V_{EBO}$	Emitter-Base Voltage ( $I_C = 0$ )	5	V
$I_C$	Collector Current	8	A
$I_{CM}$	Collector Peak Current ( $t_p < 5$ ms)	12	A
$I_B$	Base Current	5	A
$I_{BM}$	Base Peak Current ( $t_p < 5$ ms)	8	A
$P_{tot}$	Total Dissipation at $T_C = 25$ °C	55	W
$T_{stg}$	Storage Temperature	-65 to 150	°C
$T_j$	Max. Operating Junction Temperature	150	°C

# BUH615D

## THERMAL DATA

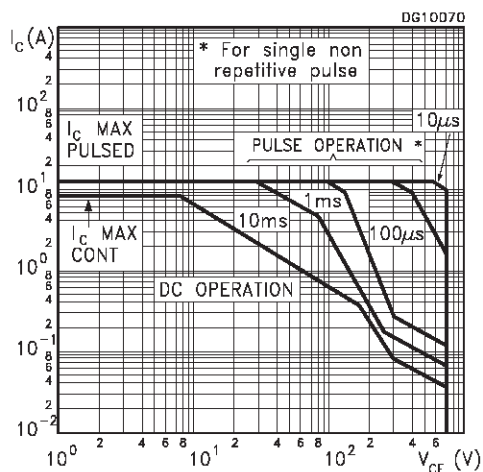
R <sub>thj-case</sub>	Thermal Resistance Junction-case	Max	2.3	°C/W
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## ELECTRICAL CHARACTERISTICS (T<sub>case</sub> = 25 °C unless otherwise specified)

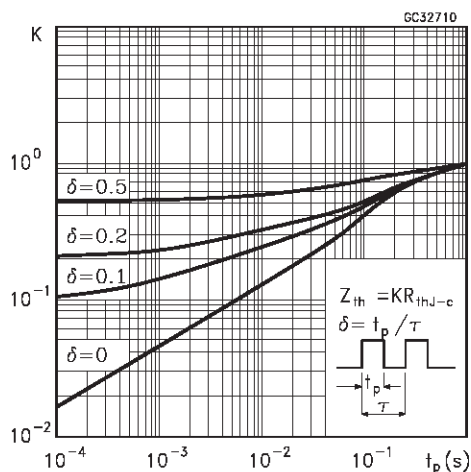
Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
I <sub>CES</sub>	Collector Cut-off Current (V <sub>BE</sub> = 0)	V <sub>CE</sub> = 1500 V V <sub>CE</sub> = 1500 V T <sub>j</sub> = 125 °C			0.2 2	mA mA
I <sub>EBO</sub>	Emitter Cut-off Current (I <sub>C</sub> = 0)	V <sub>EB</sub> = 5 V			300	mA
V <sub>CE(sat)*</sub>	Collector-Emitter Saturation Voltage	I <sub>C</sub> = 6 A I <sub>B</sub> = 2.5 A			1.5	V
V <sub>BE(sat)*</sub>	Base-Emitter Saturation Voltage	I <sub>C</sub> = 6 A I <sub>B</sub> = 2.5 A			1.3	V
h <sub>FE*</sub>	DC Current Gain	I <sub>C</sub> = 6 A V <sub>CE</sub> = 5 V	4		9	
t <sub>s</sub> t <sub>f</sub>	RESISTIVE LOAD Storage Time Fall Time	V <sub>CC</sub> = 400 V I <sub>C</sub> = 6 A I <sub>B1</sub> = 1.5 A I <sub>B2</sub> = -3 A		2.7 190	3.9 280	μs ns
t <sub>s</sub> t <sub>f</sub>	INDUCTIVE LOAD Storage Time Fall Time	I <sub>C</sub> = 6 A f = 15625 Hz I <sub>B1</sub> = 1.25 A I <sub>B2</sub> = -3 A V <sub>ceflyback</sub> = 1050 sin(π/10 10 <sup>6</sup> ) t V		2.3 350		μs ns
t <sub>s</sub> t <sub>f</sub>	INDUCTIVE LOAD Storage Time Fall Time	I <sub>C</sub> = 6 A f = 31250 Hz I <sub>B1</sub> = 1.5 A I <sub>B2</sub> = -3 A V <sub>ceflyback</sub> = 1200 sin(π/5 10 <sup>6</sup> ) t V		2.3 200		μs ns
V <sub>f</sub>	Diode Forward Voltage	I <sub>F</sub> = 5 A			2	V

\* Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

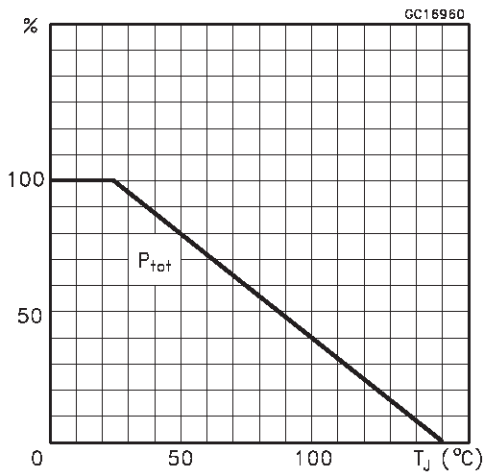
## Safe Operating Area



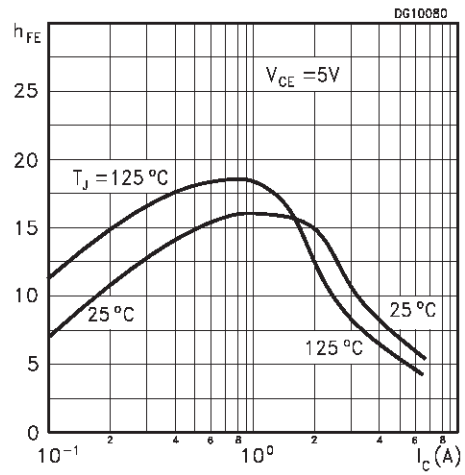
## Thermal Impedance



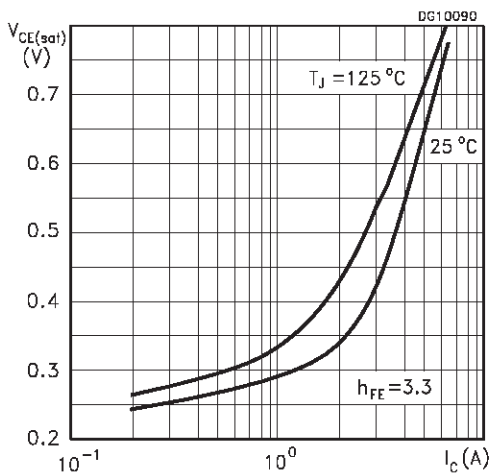
Derating Curve



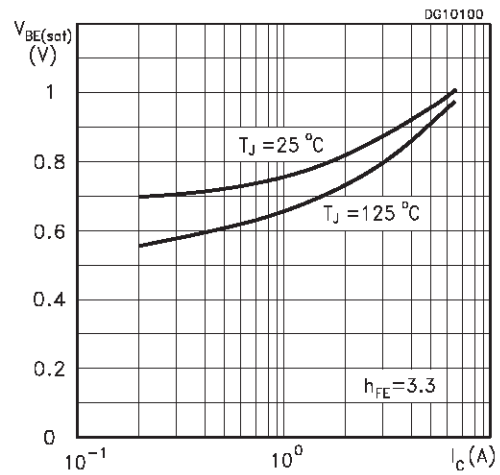
DC Current Gain



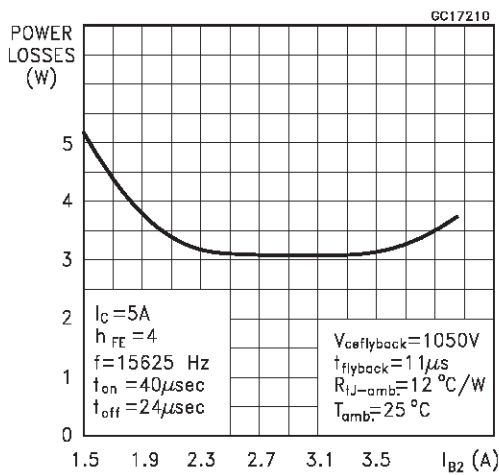
Collector Emitter Saturation Voltage



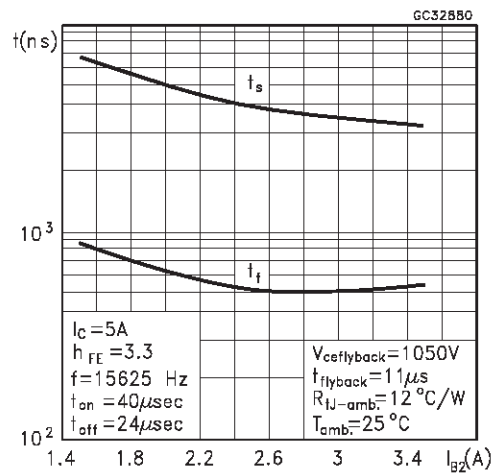
Base Emitter Saturation Voltage



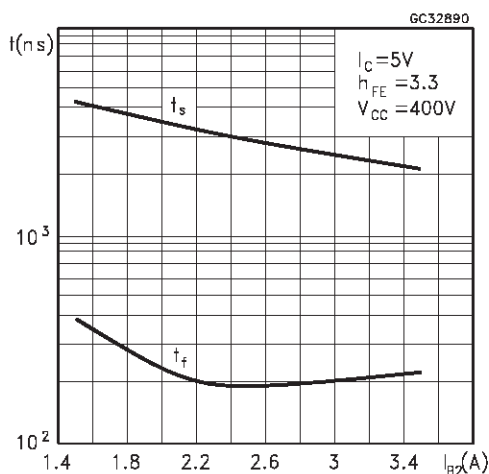
Power Losses at 16KHz



Switching Time Inductive Load at 16KHz



Switching Time Resistive Load



BASE DRIVE INFORMATION

A fundamental parameter of high voltage power transistors like those used in the horizontal deflection stage is their junction temperature  $T_j$ , which, in turn, depends on the power dissipation. This parameter turns out to influence the system reliability under normal operation. Based on that, STMicroelectronics has introduced a new dynamic, application-oriented characterization differing from the traditional data given in most datasheets.

In order to saturate the power switch and reduce conduction losses, adequate direct base current  $I_{B1}$  has to be provided for the lowest gain  $h_{FE}$  at  $T_j = 100\text{ }^\circ\text{C}$  (line scan phase). On the other hand, negative base current  $I_{B2}$  must be provided for the transistor to be turned off (retrace phase). Most of the dissipation, especially in the deflection application, occurs at switch-off so it is essential to determine the value of  $I_{B2}$  which minimizes power losses, fall time  $t_f$  and, consequently,  $T_j$ . A new set of curves have been defined to give total power losses,  $t_s$  and  $t_f$  as a function of  $I_{B2}$  at both

16 KHz and 32 KHz scanning frequencies for choosing the optimum negative drive. The test circuit is illustrated in fig. 1.

Inductance  $L_1$  serves to control the slope of the negative base current  $I_{B2}$  in order that excess carriers in the collector recombine when base current is still present, thus avoiding any tailing phenomenon in the collector current. This effect is, in any case, markedly reduced intrinsically by adopting the hollow emitter technology.

The values of  $L$  and  $C$  are calculated from the following equations:

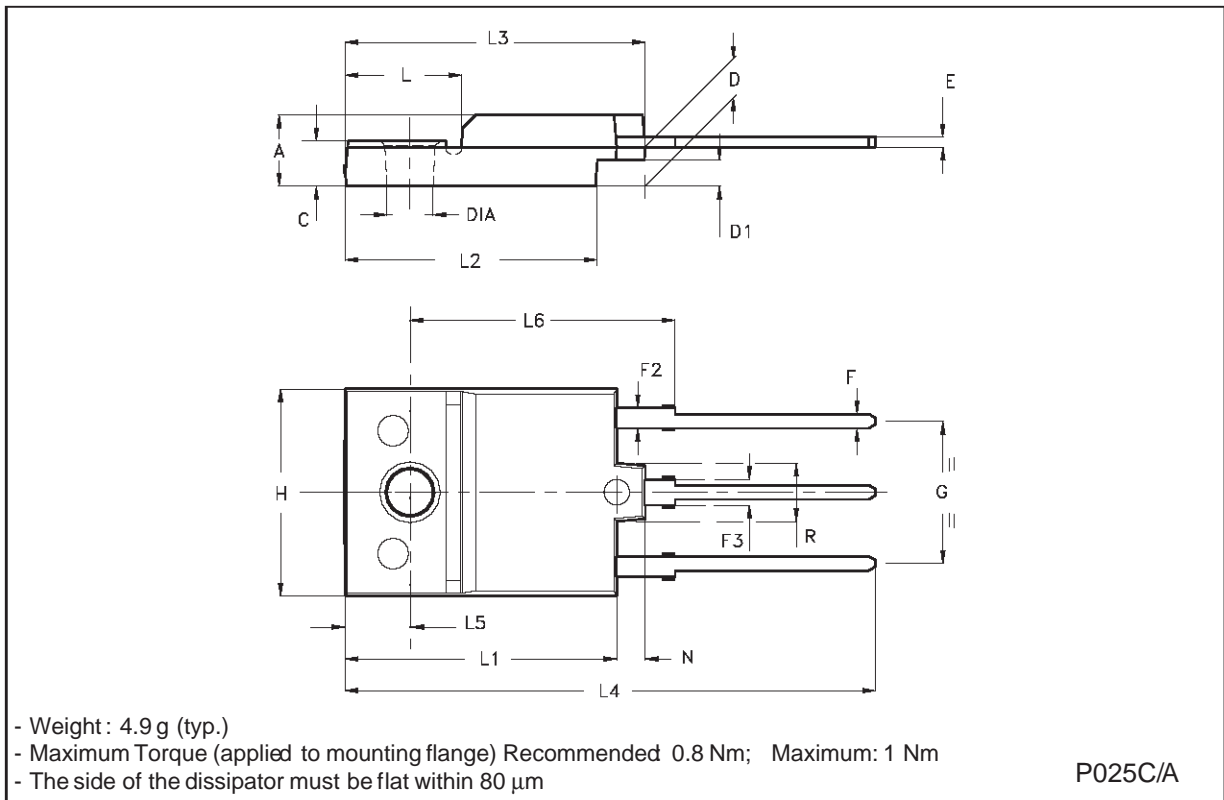
$$\frac{1}{2} L (I_C)^2 = \frac{1}{2} C (V_{CEfly})^2$$

$$\omega = 2 \pi f = \frac{1}{\sqrt{L C}}$$

Where  $I_C$ = operating collector current,  $V_{CEfly}$ = flyback voltage,  $f$ = frequency of oscillation during retrace.

**ISOWATT218 MECHANICAL DATA**

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	5.35		5.65	0.211		0.222
C	3.30		3.80	0.130		0.150
D	2.90		3.10	0.114		0.122
D1	1.88		2.08	0.074		0.082
E	0.75		0.95	0.030		0.037
F	1.05		1.25	0.041		0.049
F2	1.50		1.70	0.059		0.067
F3	1.90		2.10	0.075		0.083
G	10.80		11.20	0.425		0.441
H	15.80		16.20	0.622		0.638
L		9			0.354	
L1	20.80		21.20	0.819		0.835
L2	19.10		19.90	0.752		0.783
L3	22.80		23.60	0.898		0.929
L4	40.50		42.50	1.594		1.673
L5	4.85		5.25	0.191		0.207
L6	20.25		20.75	0.797		0.817
N	2.1		2.3	0.083		0.091
R		4.6			0.181	
DIA	3.5		3.7	0.138		0.146



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