



TS486 TS487

100mW STEREO HEADPHONE AMPLIFIER WITH STANDBY MODE

- OPERATING FROM $V_{CC}=2V$ to 5.5V
- **STANDBY MODE ACTIVE LOW** (TS486) or HIGH (TS487)
- **OUTPUT POWER:** 102mW @5V, 38mW @3.3V into 16Ω with 0.1% THD+N max (1kHz)
- **LOW CURRENT CONSUMPTION:** 2.5mA max
- High Signal-to-Noise ratio: 103dB(A) at 5V
- High Crosstalk immunity: 83dB (F=1kHz)
- PSRR: 58 dB (F=1kHz), inputs grounded
- ON/OFF click reduction circuitry
- Unity-Gain Stable
- **SHORT CIRCUIT LIMITATION**
- Available in SO8, MiniSO8 & QFN 3x3mm

DESCRIPTION

The TS486/7 is a dual audio power amplifier capable of driving, in single-ended mode, either a 16 or a 32Ω stereo headset.

Capable of descending to low voltages, it delivers up to 90mW per channel (into 16Ω loads) of continuous average power with 0.3% THD+N in the audio bandwidth from a 5V power supply.

An externally-controlled standby mode reduces the supply current to 10nA (typ.). The unity gain stable TS486/7 can be configured by external gain-setting resistors or used in a fixed gain version.

APPLICATIONS

- Headphone Amplifier
- Mobile phone, PDA, computer motherboard
- High end TV, portable audio player

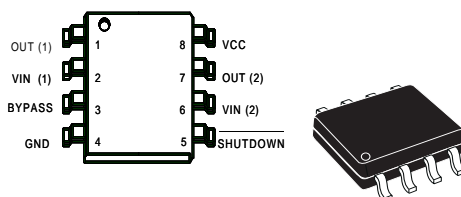
ORDER CODE

Part Number	Temperature Range: I	Package			Gain	Marking
		D	S	Q		
TS486	-40, +85°C	•			external	TS486I
TS487		•			external	TS487I
TS486			•	tba	external	K86A
TS486-1			tba	tba	x1/0dB	K86B
TS486-2			tba	tba	x2/6dB	K86C
TS486-4			tba	tba	x4/12dB	K86D
TS487			•	tba	external	K87A
TS487-1			tba	tba	x1/0dB	K87B
TS487-2			tba	tba	x2/6dB	K87C
TS487-4			tba	tba	x4/12dB	K87D

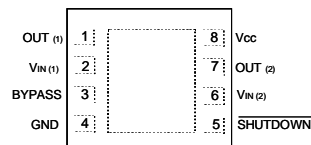
MiniSO & QFN only available in Tape & Reel with T suffix, SO is available in Tube (D) and in Tape & Reel (DT)

PIN CONNECTIONS (top view)

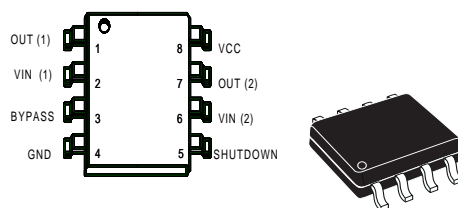
TS486IDT: SO8, TS486IST, TS486-1IST, TS486-2IST, TS486-4IST: MiniSO8



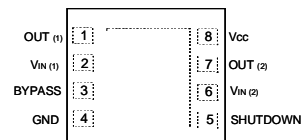
TS486-IQT, TS486-1IQT, TS486-2IQT, TS486-4IQT: QFN8



TS487IDT: SO8, TS487IST, TS487-1IST, TS487-2IST, TS487-4IST: MiniSO8



TS487-IQT, TS487-1IQT, TS487-2IQT, TS487-4IQT: QFN8



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{CC}	Supply voltage ¹⁾	6	V
V _i	Input Voltage	-0.3v to V _{CC} +0.3v	V
T _{stg}	Storage Temperature	-65 to +150	°C
T _j	Maximum Junction Temperature	150	°C
R _{thja}	Thermal Resistance Junction to Ambient SO8 MiniSO8 QFN8	175 215 70	°C/W
Pd	Power Dissipation ²⁾ SO8 MiniSO8 QFN8	0.71 0.58 1.79	W
ESD	Human Body Model (pin to pin): TS486, TS487 ³⁾	1.5	kV
ESD	Machine Model - 220pF - 240pF (pin to pin)	100	V
Latch-up	Latch-up Immunity (All pins)	200	mA
	Lead Temperature (soldering, 10sec)	250	°C
	Output Short-Circuit to V _{CC} or GND	continuous ⁴⁾	

1. All voltage values are measured with respect to the ground pin.
2. Pd has been calculated with Tamb = 25°C, Tjunction = 150°C.
3. TS487 stands 1.5KV on all pins except standby pin which stands 1KV.
4. Attention must be paid to continuous power dissipation (V_{DD} x 300mA). Exposure of the IC to a short circuit for an extended time period is dramatically reducing product life expectancy.

OPERATING CONDITIONS

Symbol	Parameter	Value	Unit
V _{CC}	Supply Voltage	2 to 5.5	V
R _L	Load Resistor	≥ 16	Ω
T _{oper}	Operating Free Air Temperature Range	-40 to + 85	°C
C _L	Load Capacitor R _L = 16 to 100Ω R _L > 100Ω	400 100	pF
V _{STB}	Standby Voltage Input TS486 ACTIVE / TS487 in STANDBY TS486 in STANDBY / TS487 ACTIVE	1.5 ≤ V _{STB} ≤ V _{CC} GND ≤ V _{STB} ≤ 0.4 ¹⁾	V
R _{THJA}	Thermal Resistance Junction to Ambient SO8 miniSO8 QFN8 ²⁾	150 190 41	°C/W

1. The minimum current consumption (I_{STANDBY}) is guaranteed at GND (TS486) or V_{CC} (TS487) for the whole temperature range.
2. When mounted on a 4-layer PCB.

FIXED GAIN VERSION SPECIFIC ELECTRICAL CHARACTERISTICS

V_{CC} from +5V to +2V, GND = 0V, T_{amb} = 25°C (unless otherwise specified)

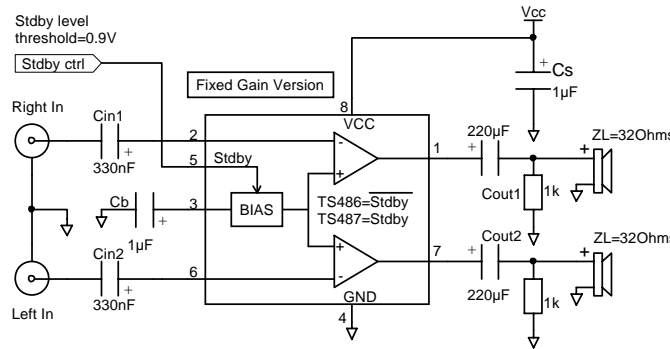
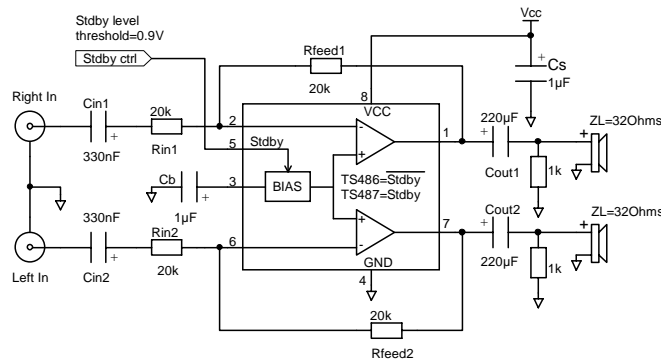
Symbol	Parameter	Min.	Typ.	Max.	Unit
R _{IN 1,2}	Input Resistance ¹⁾		20		kΩ
G	Gain value for Gain TS486/TS487-1		0dB		dB
	Gain value for Gain TS486/TS487-2		6dB		
	Gain value for Gain TS486/TS487-4		12dB		

1. See figure 30 to establish the value of C_{in} vs. -3dB cut off frequency.

APPLICATION COMPONENTS INFORMATION

Components	Functional Description
R _{IN1,2}	Inverting input resistor which sets the closed loop gain in conjunction with R _{FEED} . This resistor also forms a high pass filter with C _{IN} (fc = 1 / (2 x Pi x R _{IN} x C _{IN})). Not needed in fixed gain versions.
C _{IN1,2}	Input coupling capacitor which blocks the DC voltage at the amplifier's input terminal.
R _{FEED1,2}	Feedback resistor which sets the closed loop gain in conjunction with R _{IN} . A _V = Closed Loop Gain= -R _{FEED} /R _{IN} . Not needed in fixed gain versions.
C _S	Supply Bypass capacitor which provides power supply filtering.
C _B	Bypass capacitor which provides half supply filtering.
C _{OUT1,2}	Output coupling capacitor which blocks the DC voltage at the load input terminal. This capacitor also forms a high pass filter with R _L (fc = 1 / (2 x Pi x R _L x C _{OUT})).

TYPICAL APPLICATION SCHEMATICS



ELECTRICAL CHARACTERISTICS

$V_{CC} = +5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.8	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ¹⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	60 95	64 65 102 108		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 60mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 90mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ²⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, F = 1kHz, $V_{ripple} = 200mV_{pp}$	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	106	115		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	4.45 4.2	0.45 4.52 0.6 4.35	0.5 0.7	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ²⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	103		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz		83 79 80 72		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. Only for external gain version.

2. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS

$V_{CC} = +3.3V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified) ¹⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.8	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ²⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	23 36	26 28 38 42		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 35mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ³⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, F = 1kHz, $V_{ripple} = 200mV_{pp}$	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	64	75		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	2.85 2.68	0.3 3 0.45 2.85	0.38 0.52	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ³⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	98		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. All electrical values are guaranteed with correlation measurements at 2V and 5V.

2. Only for external gain version.

3. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS

$V_{CC} = +2.5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)¹⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.7	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ²⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	12.5 17.5	13 14 21 22		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 10mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ³⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, F = 1kHz, Vripple = 200mVpp	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	45	56		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	2.14 1.97	0.25 2.25 0.35 2.15	0.32 0.45	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ³⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	95		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. All electrical values are guaranteed with correlation measurements at 2V and 5V.

2. Only for external gain version.

3. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS $V_{CC} = +2V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.7	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ¹⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.3% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	7 9.5	8 9 12 13		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 6.5mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 8mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ²⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, F = 1kHz, Vripple = 200mVpp	52	57		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	33	41		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	1.67 1.53	0.24 1.73 0.33 1.63	0.29 0.41	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ²⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	93		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$, $A_v=-1$ F = 1kHz F = 20Hz to 20kHz		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. Only for external gain version.

2. Guaranteed by design and evaluation.

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Crosstalk vs Frequency	92 to 95	24
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Fig. 1: Open Loop Gain and Phase vs Frequency

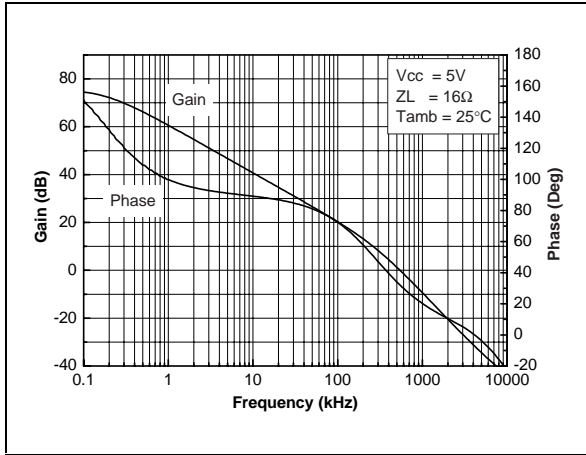


Fig. 2: Open Loop Gain and Phase vs Frequency

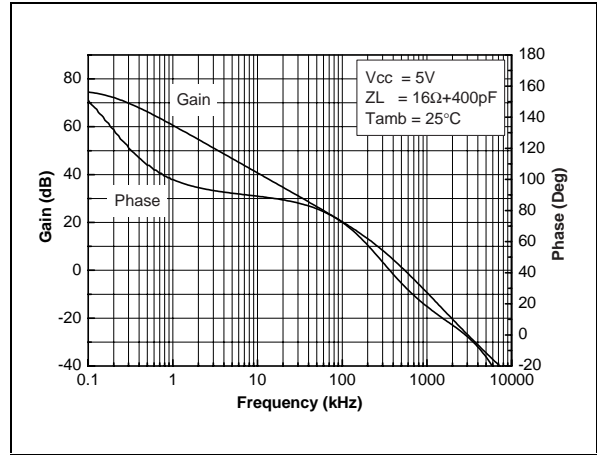


Fig. 3: Open Loop Gain and Phase vs Frequency

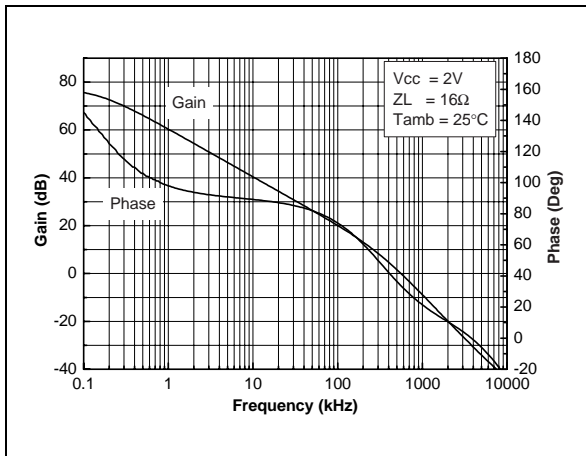


Fig. 4: Open Loop Gain and Phase vs Frequency

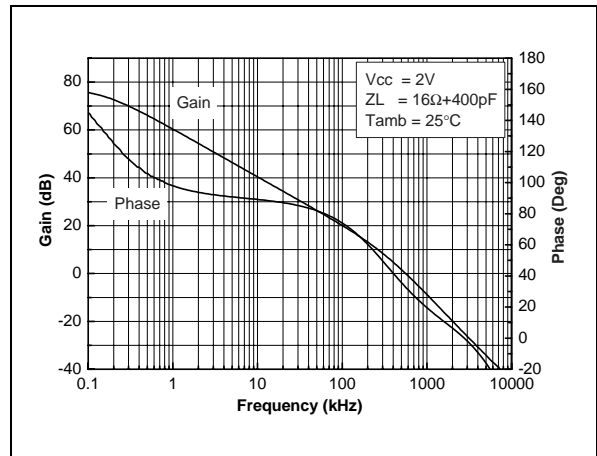


Fig. 5: Open Loop Gain and Phase vs Frequency

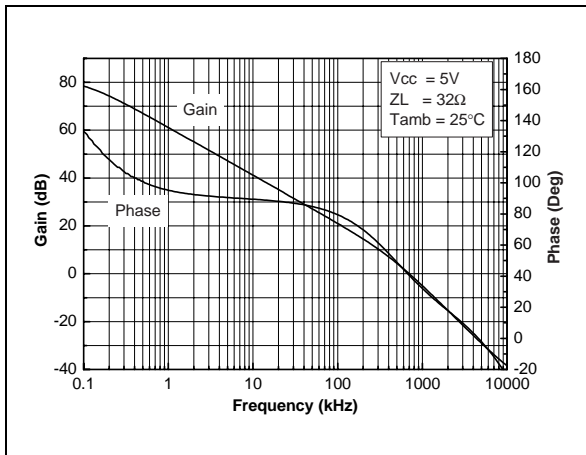


Fig. 6: Open Loop Gain and Phase vs Frequency

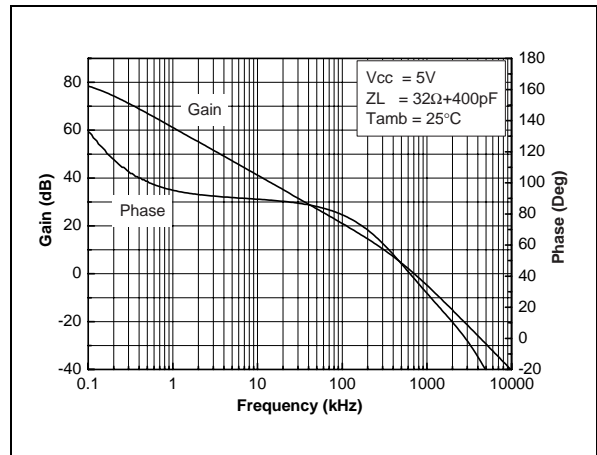


Fig. 7: Open Loop Gain and Phase vs Frequency

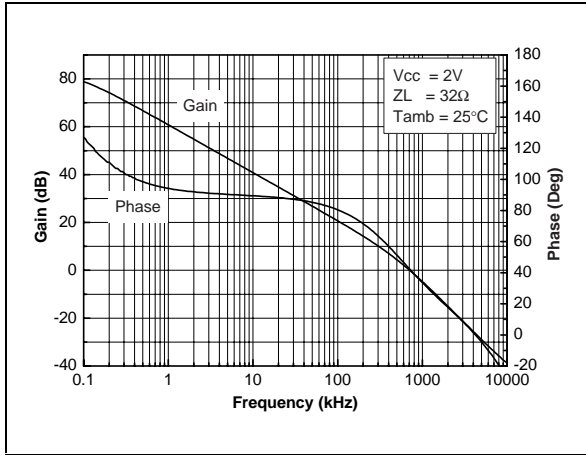


Fig. 8: Open Loop Gain and Phase vs Frequency

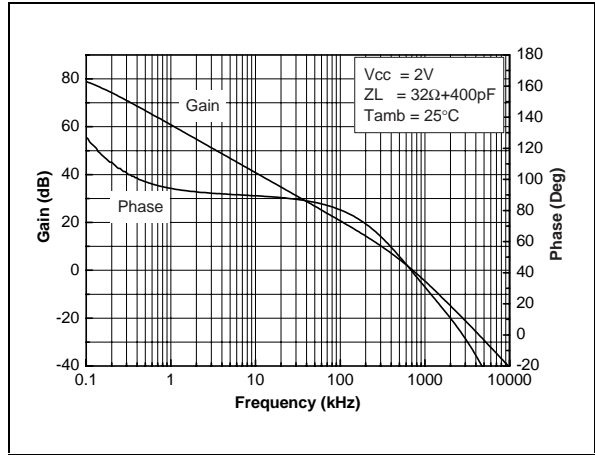


Fig. 9: Open Loop Gain and Phase vs Frequency

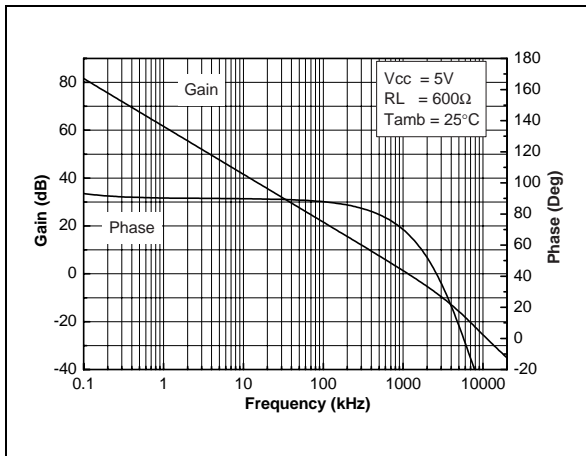


Fig. 10: Open Loop Gain and Phase vs Frequency

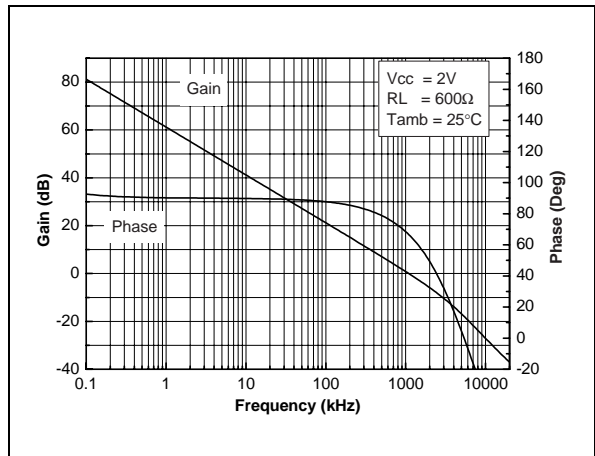


Fig. 11: Current Consumption vs Power Supply Voltage

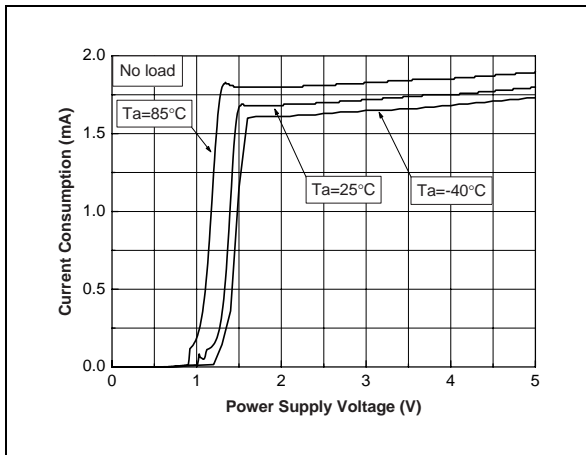


Fig. 12: Current Consumption vs Standby Voltage

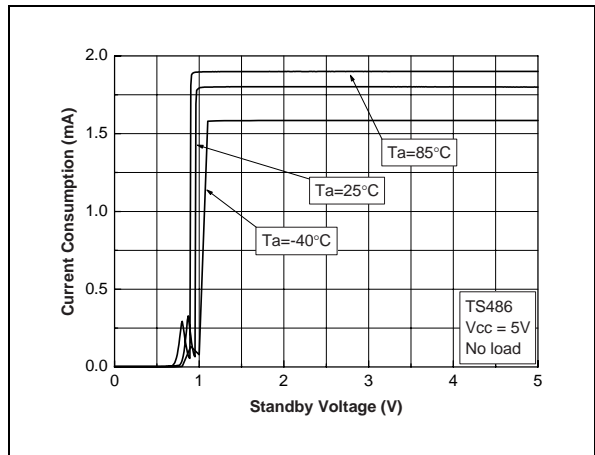


Fig. 13: Current Consumption vs Standby Voltage

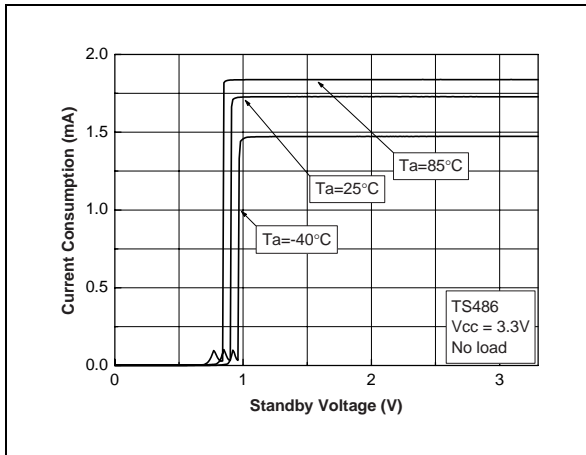


Fig. 14: Current Consumption vs Standby Voltage

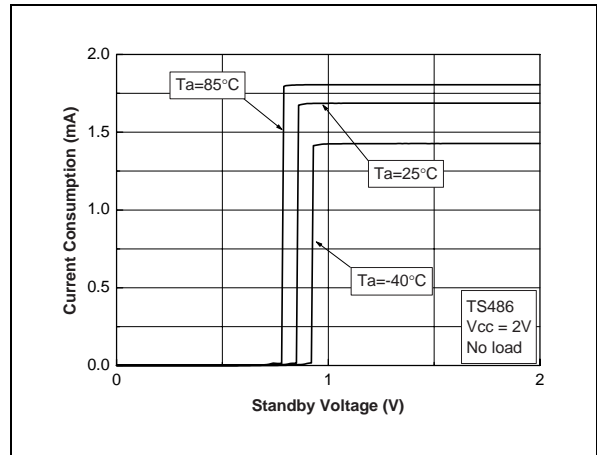


Fig. 15: Current Consumption vs Standby Voltage

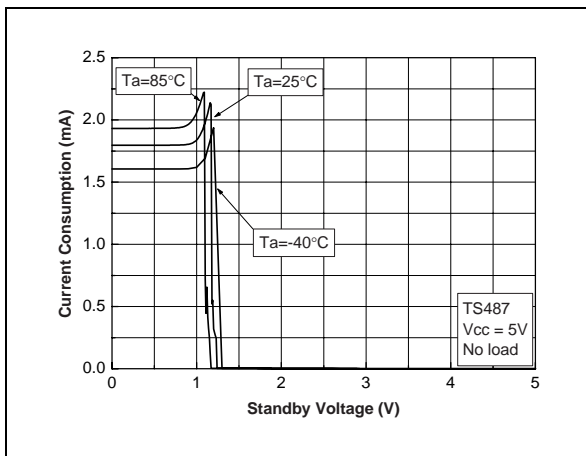


Fig. 16: Current Consumption vs Standby Voltage

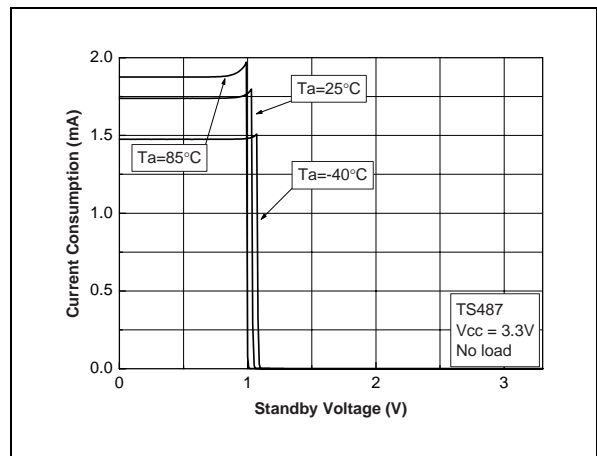


Fig. 17: Current Consumption vs Standby Voltage

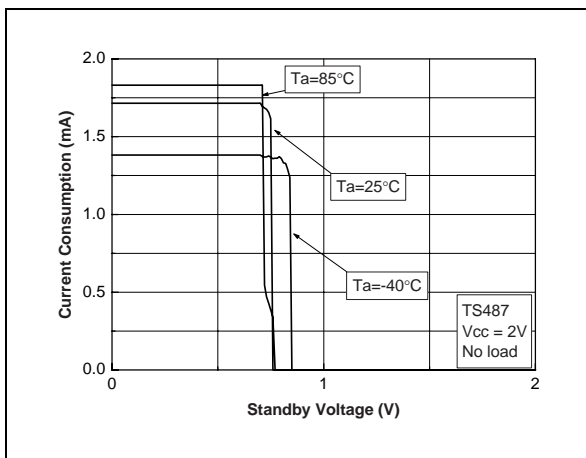


Fig. 18: Output Power vs Power Supply Voltage

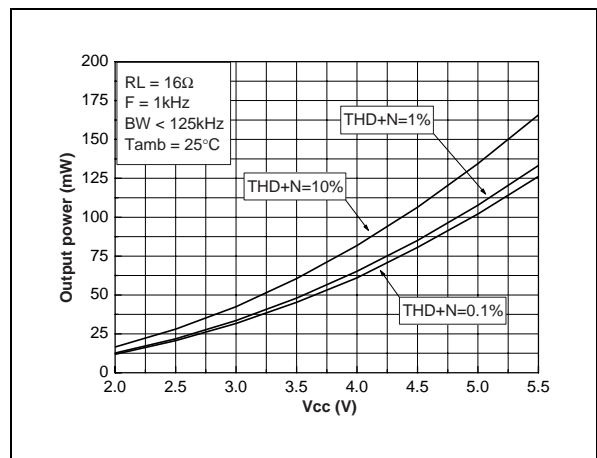


Fig. 19: Output Power vs Power Supply Voltage

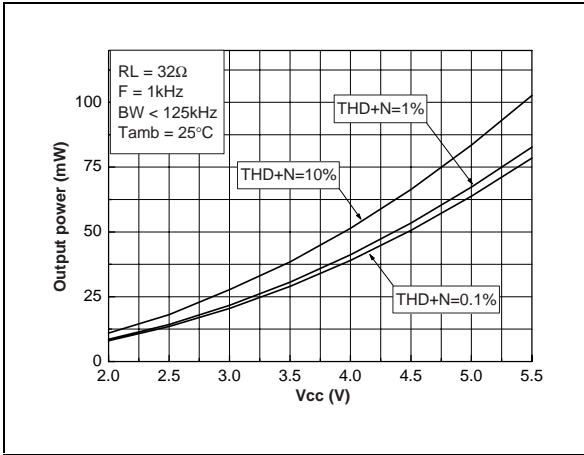


Fig. 20: Output Power vs Load Resistor

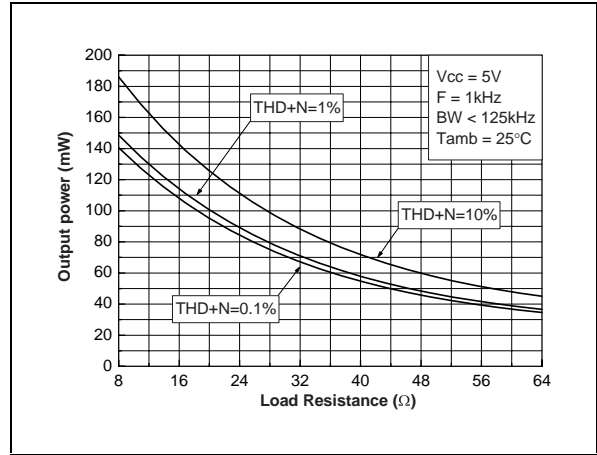


Fig. 21: Output Power vs Load Resistor

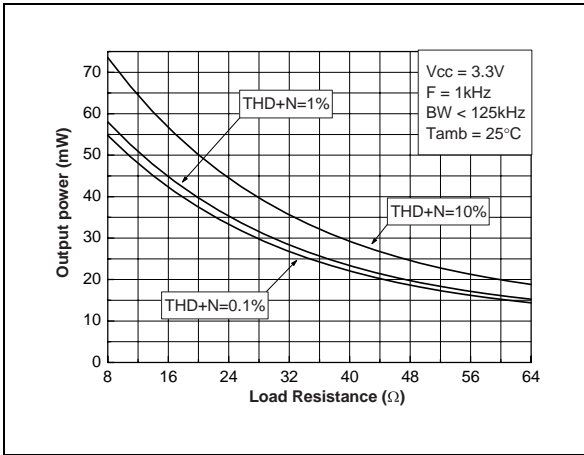


Fig. 22: Output Power vs Load Resistor

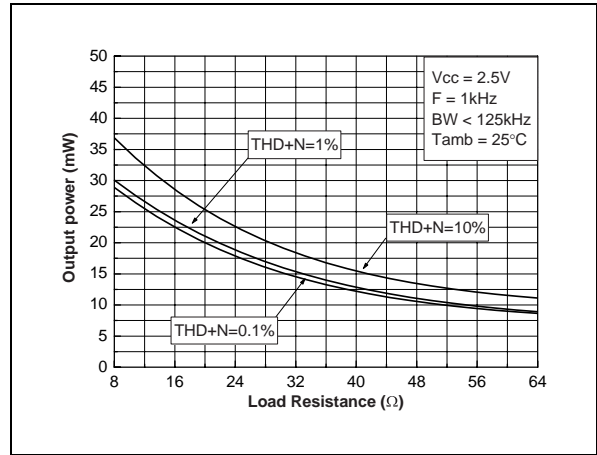


Fig. 23: Output Power vs Load Resistor

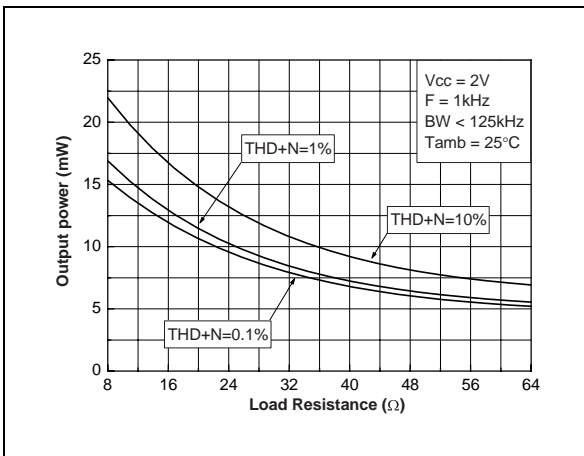


Fig. 24: Power Dissipation vs Output Power

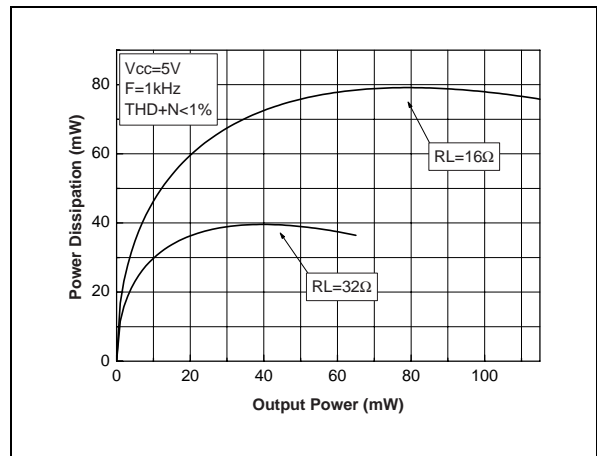


Fig. 25: Power Dissipation vs Output Power

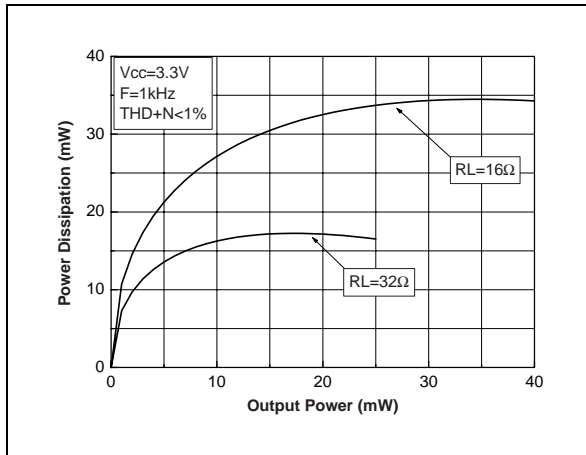


Fig. 26: Power Dissipation vs Output Power

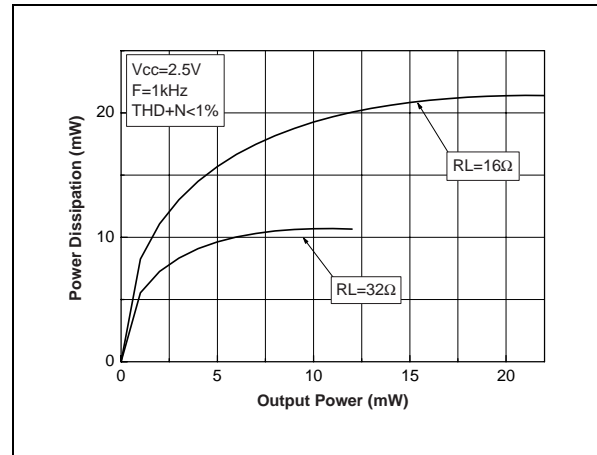


Fig. 27: Power Dissipation vs Output Power

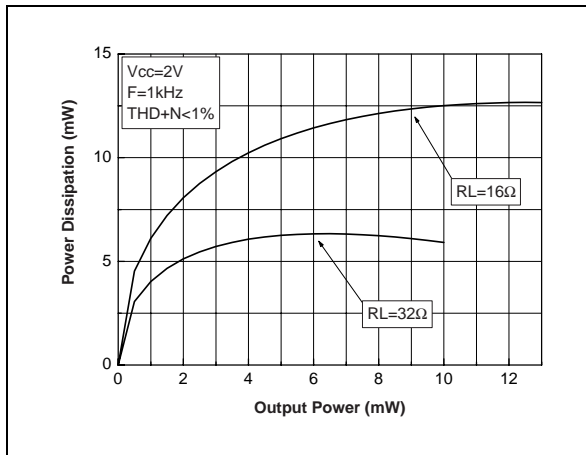


Fig. 28: Power Derating vs Ambient Temperature

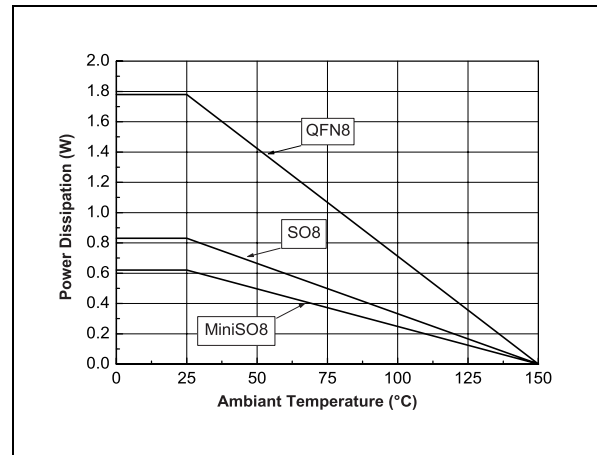


Fig. 29: Output Voltage Swing vs Power Supply Voltage

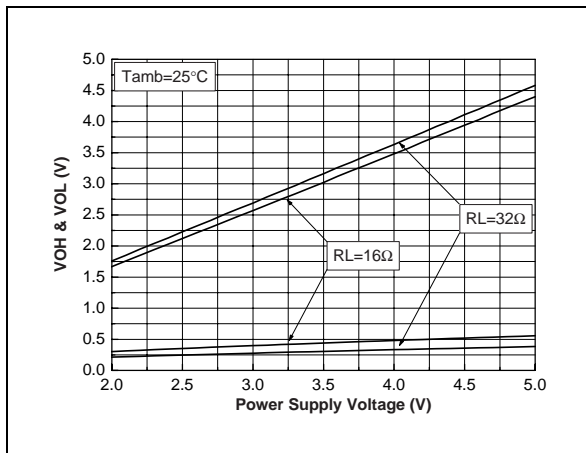


Fig. 30: Low Frequency Cut Off vs Input Capacitor for fixed gain versions.

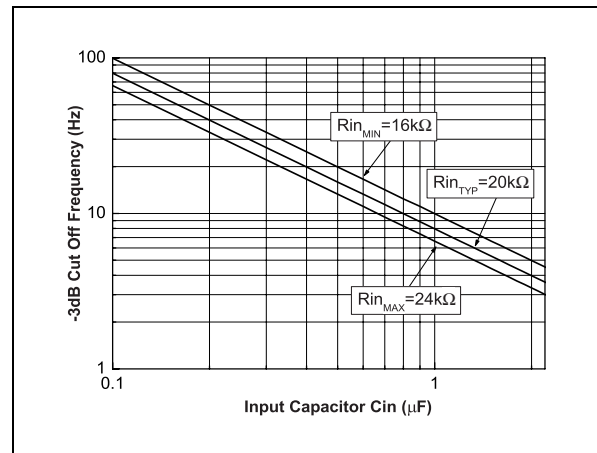


Fig. 31: THD + N vs Output Power

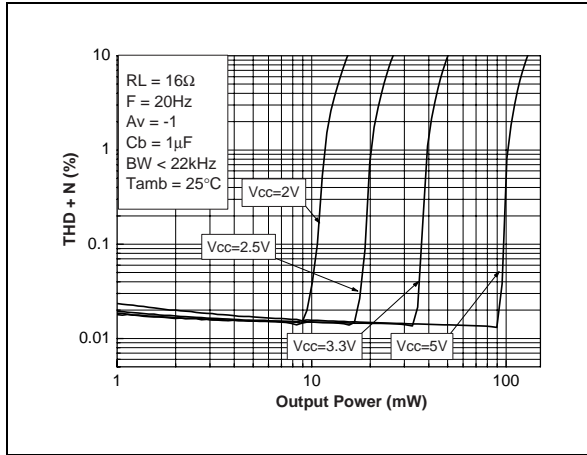


Fig. 32: THD + N vs Output Power

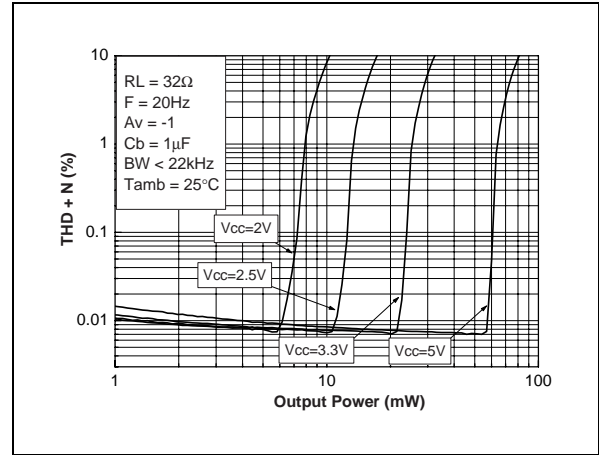


Fig. 33: THD + N vs Output Power

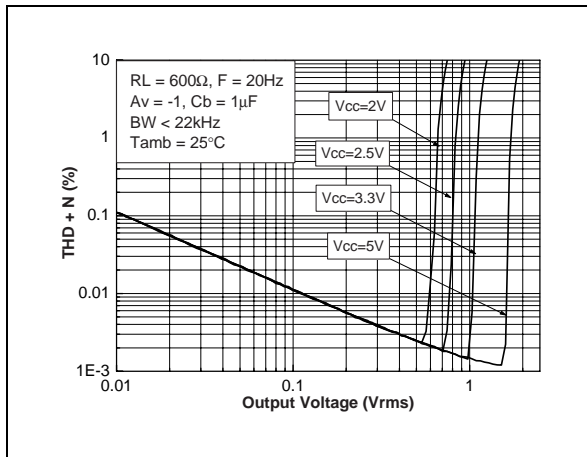


Fig. 34: THD + N vs Output Power

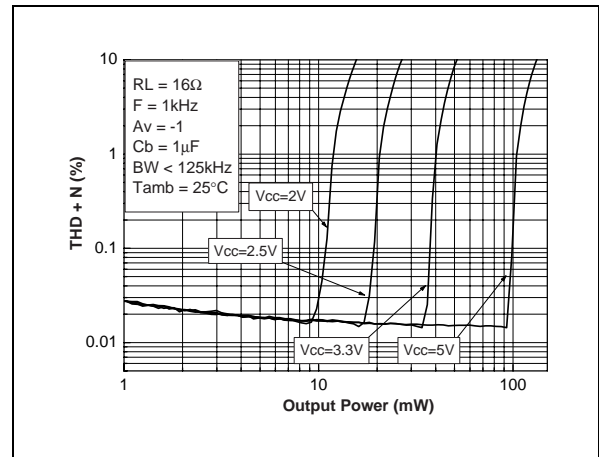


Fig. 35: THD + N vs Output Power

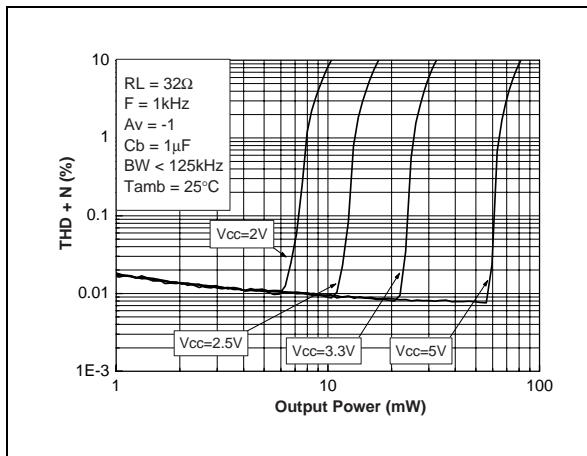


Fig. 36: THD + N vs Output Power

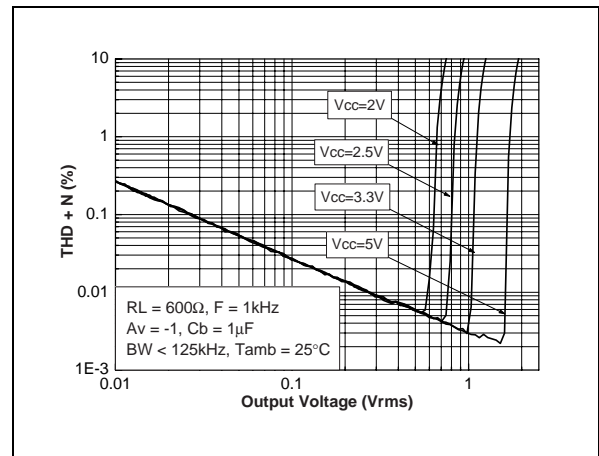


Fig. 37: THD + N vs Output Power

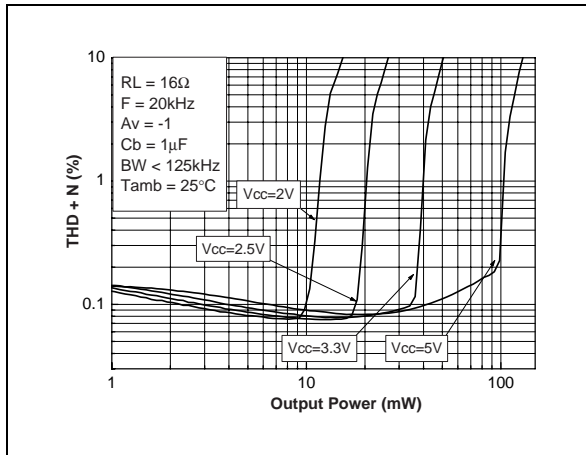


Fig. 38: THD + N vs Output Power

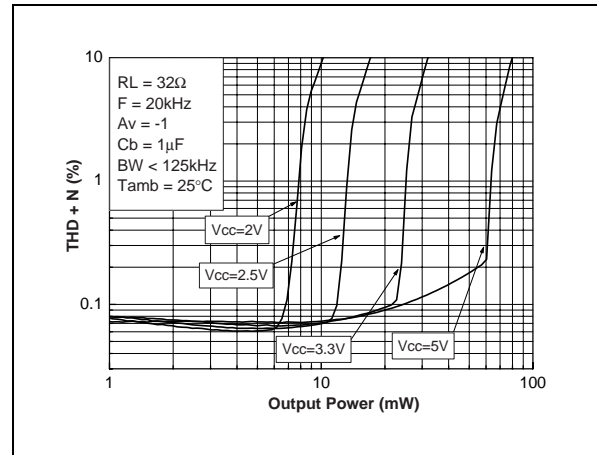


Fig. 39: THD + N vs Output Power

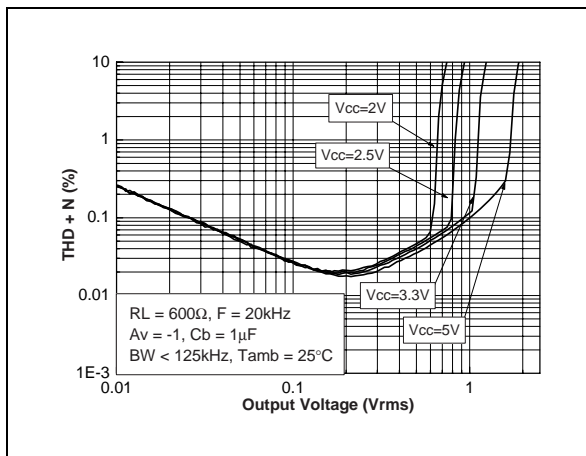


Fig. 40: THD + N vs Frequency

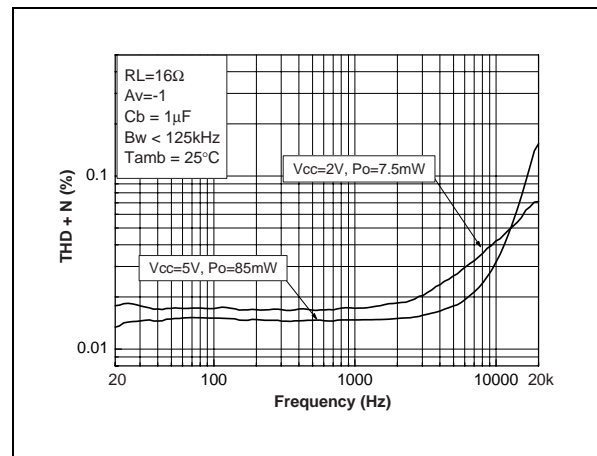


Fig. 41: THD + N vs Frequency

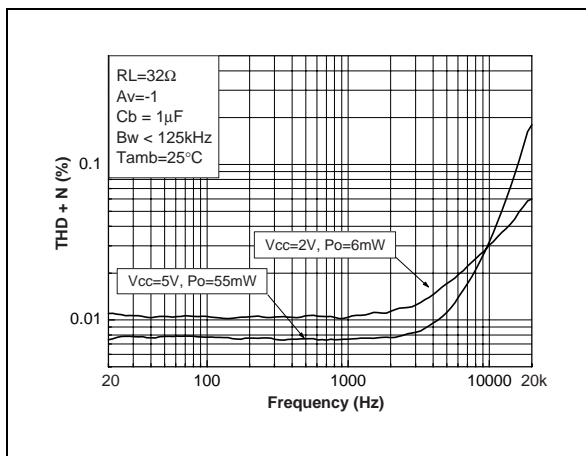


Fig. 42: THD + N vs Frequency

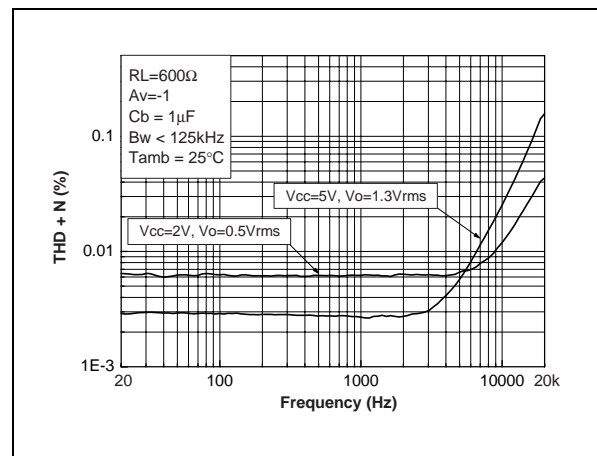


Fig. 43: Crosstalk vs Frequency

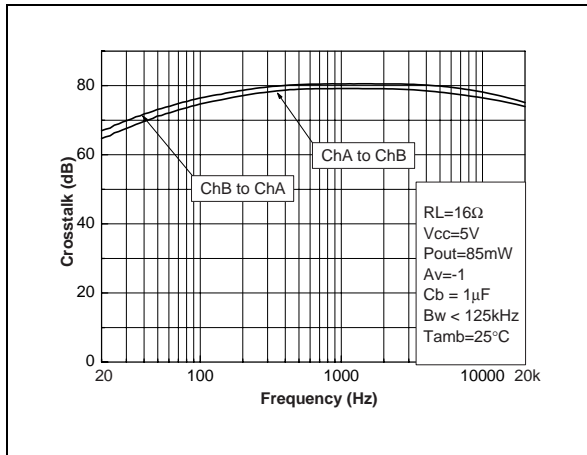


Fig. 44: Crosstalk vs Frequency

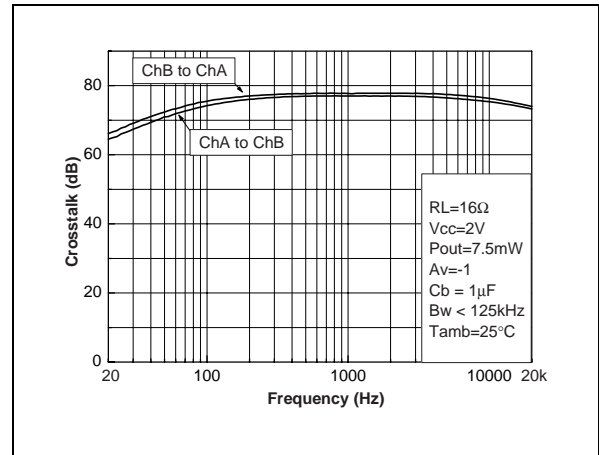


Fig. 45: Crosstalk vs Frequency

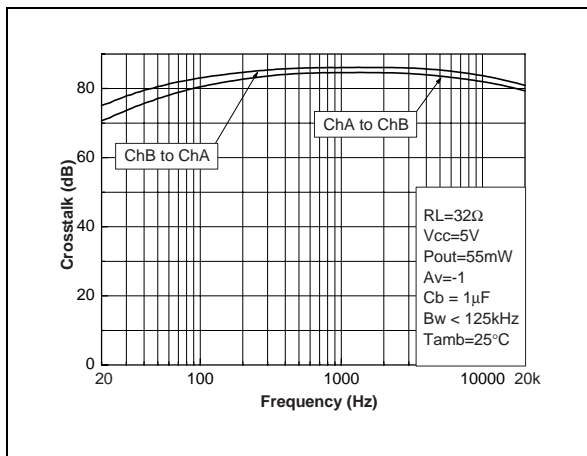


Fig. 46: Crosstalk vs Frequency

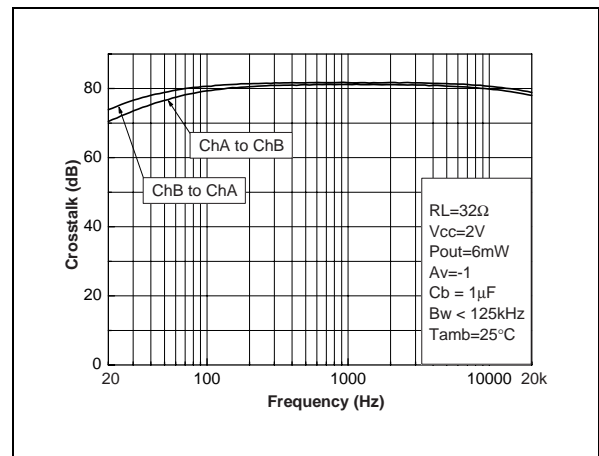


Fig. 47: Crosstalk vs Frequency

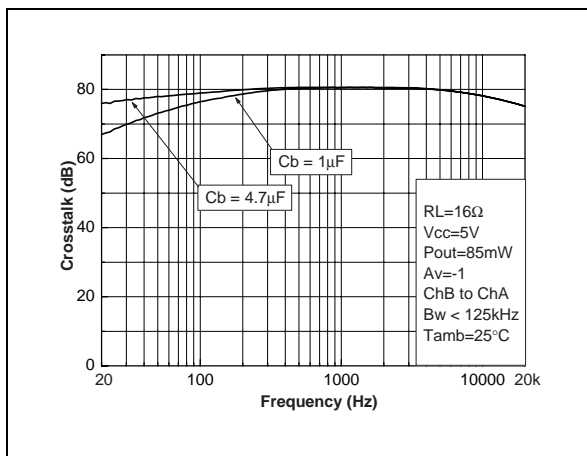


Fig. 48: Crosstalk vs Frequency

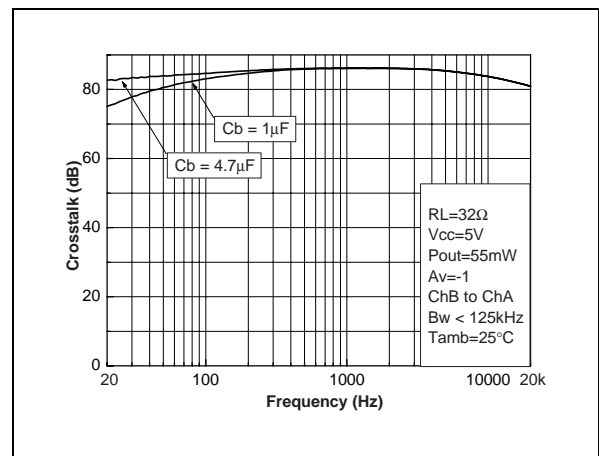


Fig. 49: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

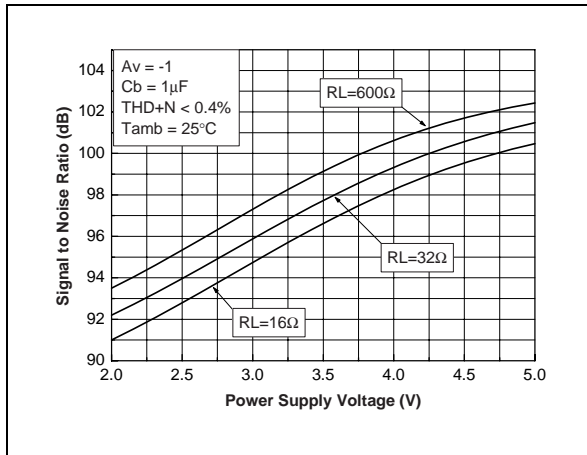


Fig. 50: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

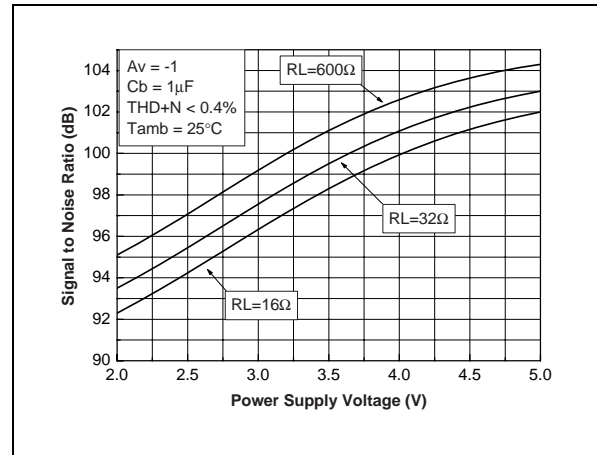


Fig. 51: PSRR vs Power Supply Voltage

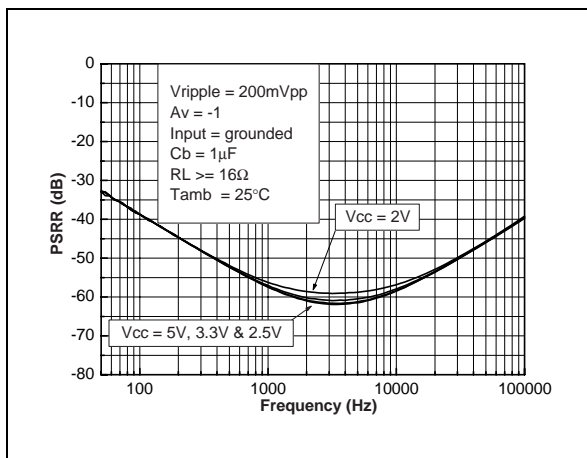


Fig. 52: PSRR vs Bypass Capacitor

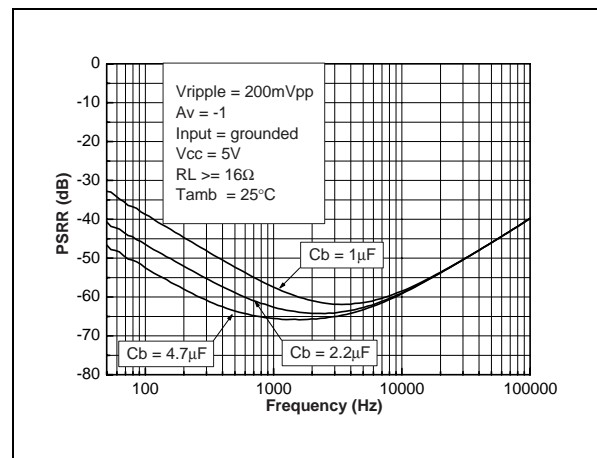


Fig. 53: PSRR vs Input Capacitor

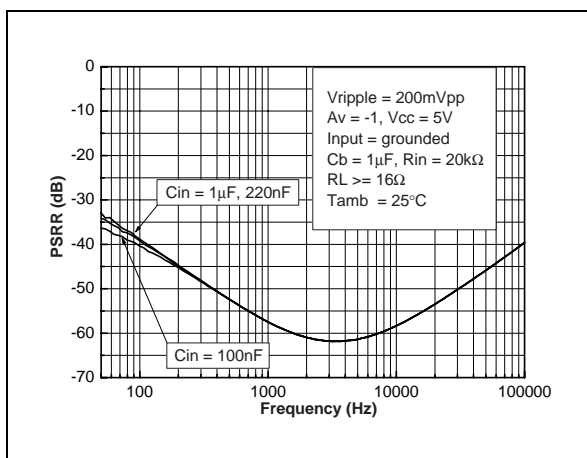


Fig. 54: PSRR vs Output Capacitor

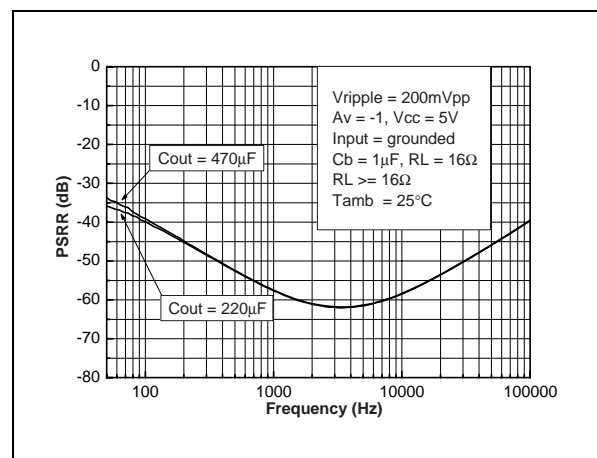


Fig. 55: PSRR vs Output Capacitor

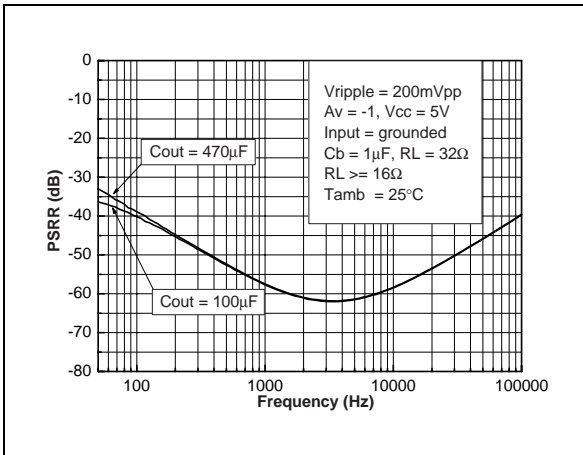


Fig. 56: PSRR vs Power Supply Voltage

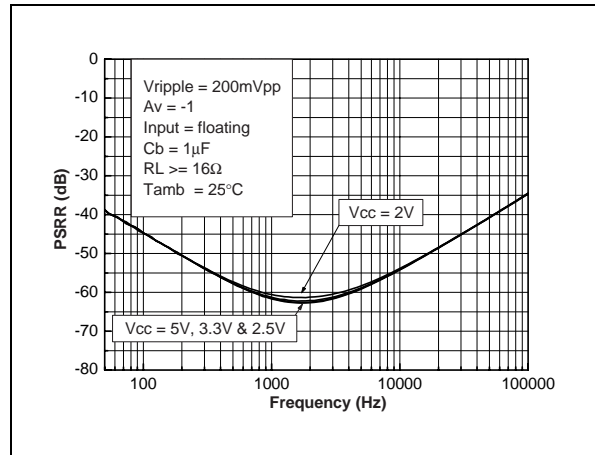


Fig. 57: THD + N vs Output Power

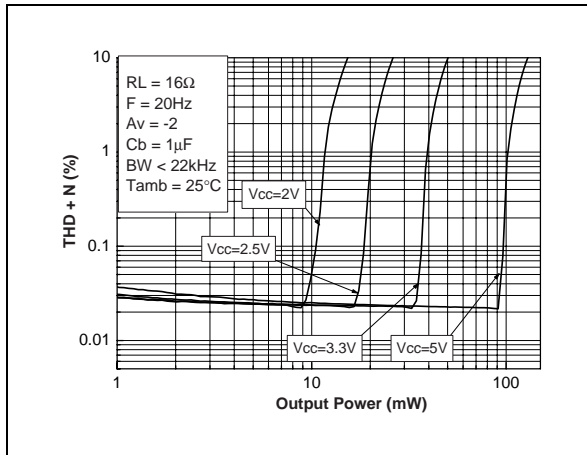


Fig. 58: THD + N vs Output Power

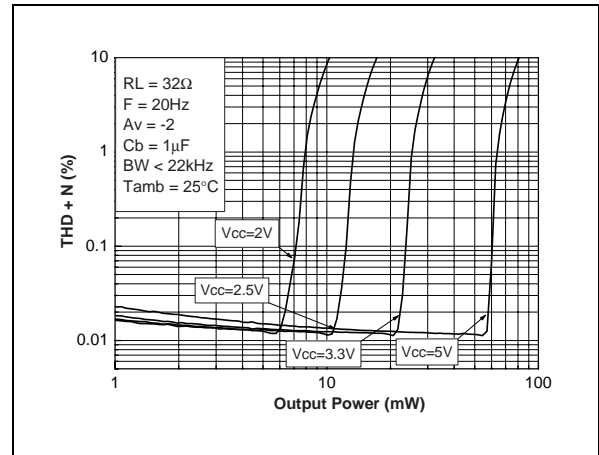


Fig. 59: THD + N vs Output Power

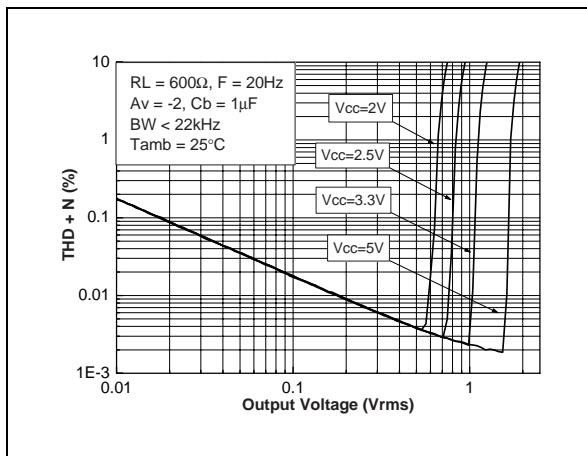


Fig. 60: THD + N vs Output Power

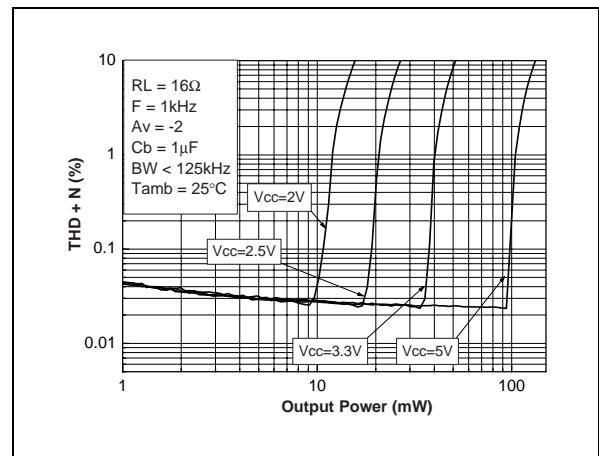


Fig. 61: THD + N vs Output Power

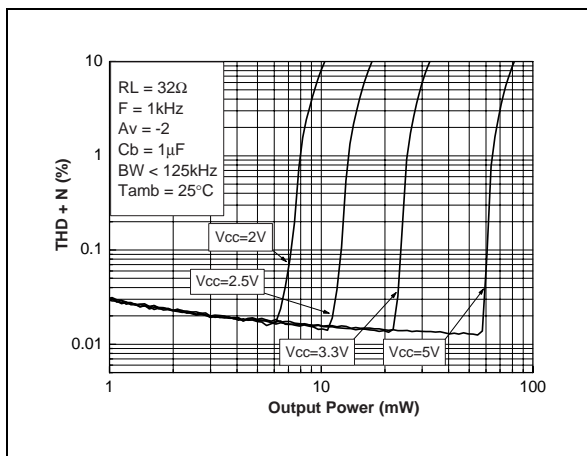


Fig. 62: THD + N vs Output Power

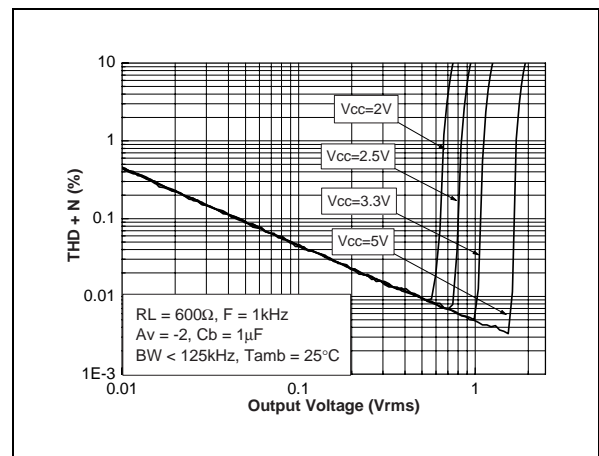


Fig. 63: THD + N vs Output Power

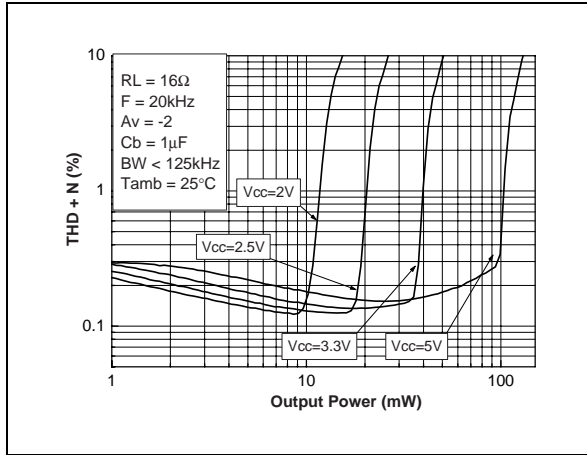


Fig. 64: THD + N vs Output Power

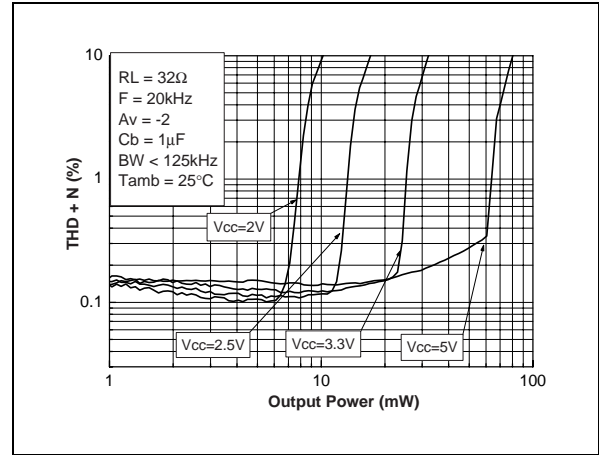


Fig. 65: THD + N vs Output Power

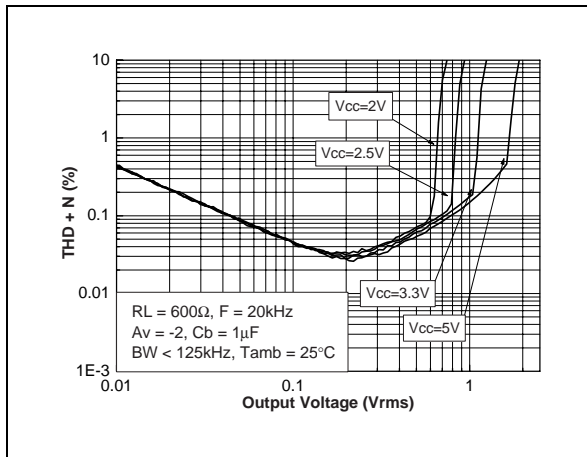


Fig. 66: THD + N vs Frequency

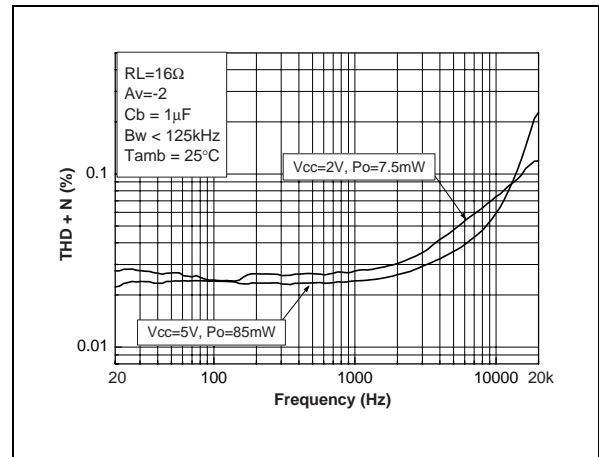


Fig. 67: THD + N vs Frequency

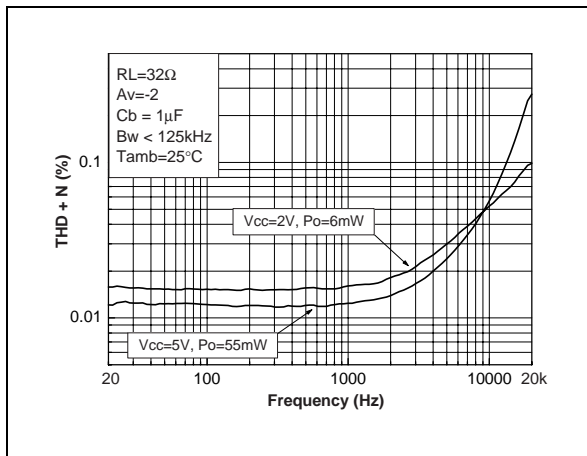


Fig. 68: THD + N vs Frequency

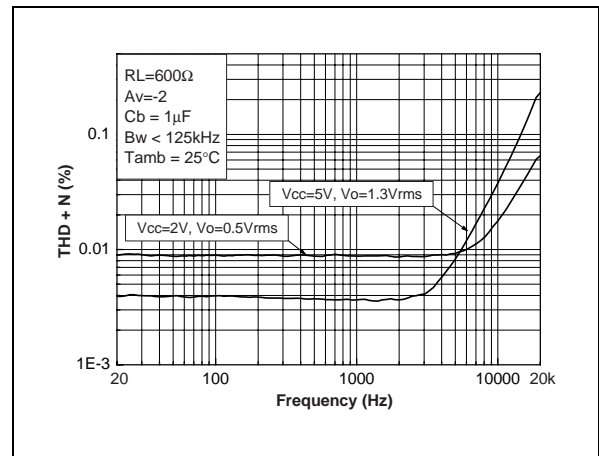


Fig. 69: Crosstalk vs Frequency

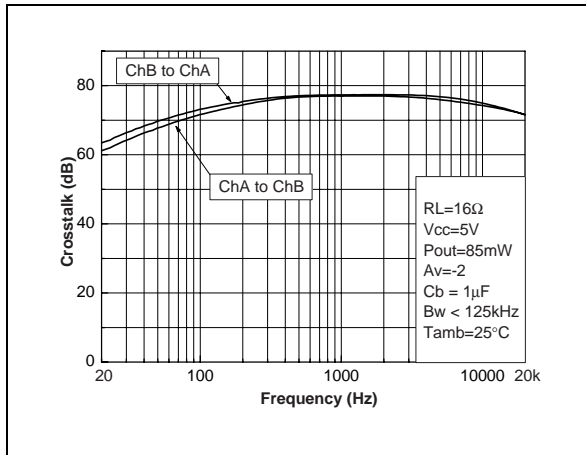


Fig. 70: Crosstalk vs Frequency

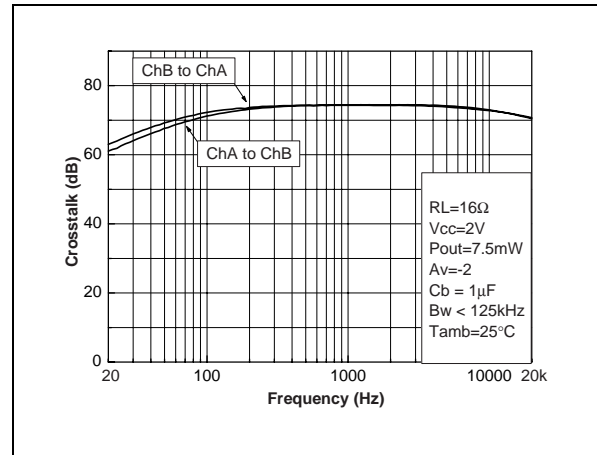


Fig. 71: Crosstalk vs Frequency

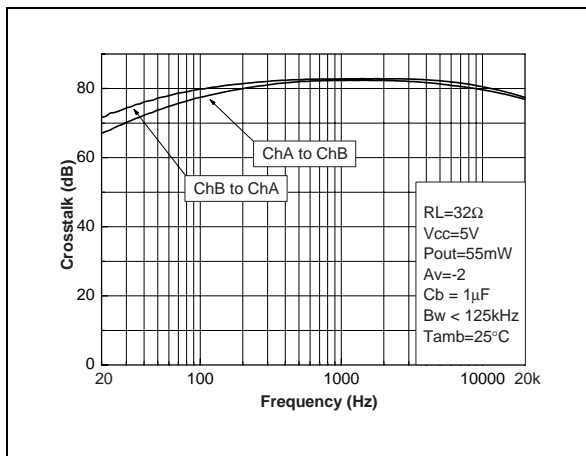


Fig. 72: Crosstalk vs Frequency

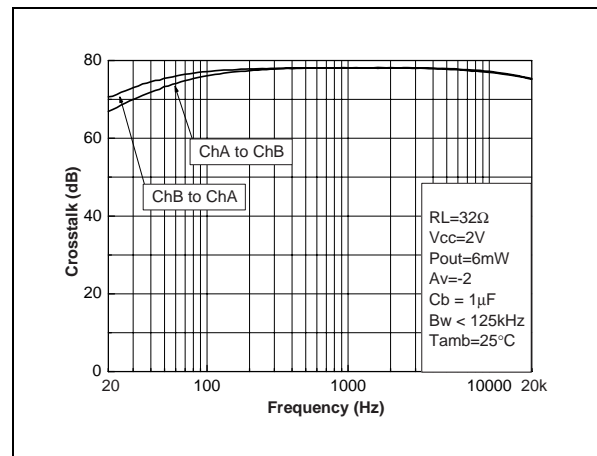


Fig. 73: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

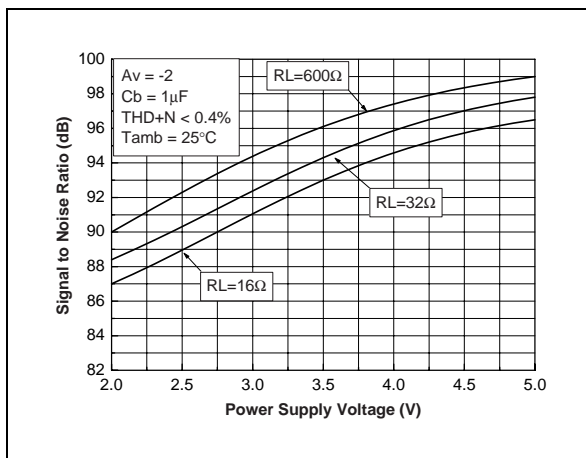


Fig. 74: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

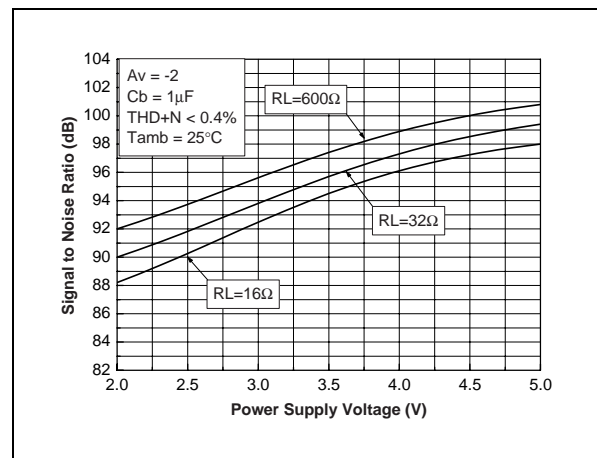


Fig. 75: PSRR vs Power Supply Voltage

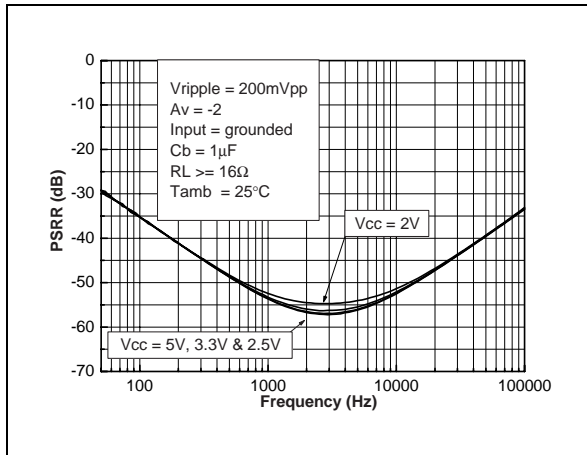


Fig. 76: PSRR vs Bypass Capacitor

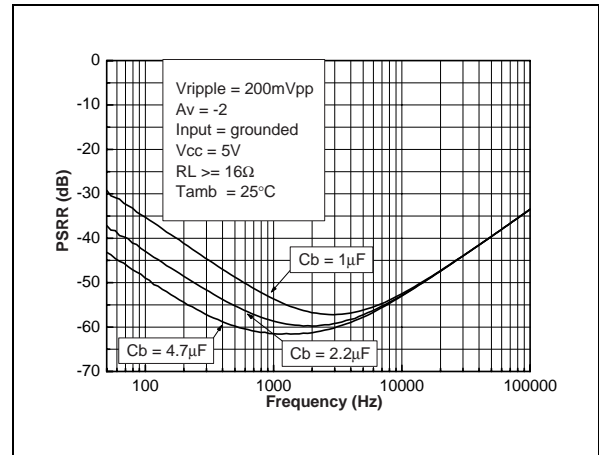


Fig. 77: PSRR vs Input Capacitor

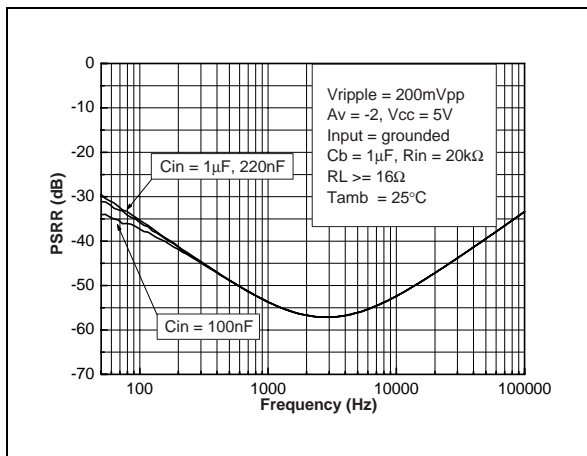


Fig. 78: PSRR vs Output Capacitor

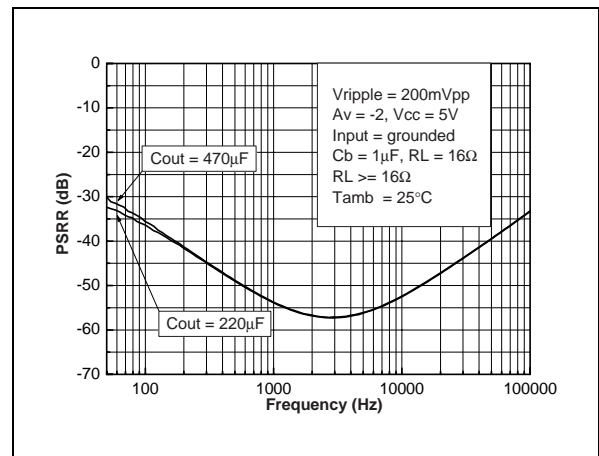


Fig. 79: PSRR vs Output Capacitor

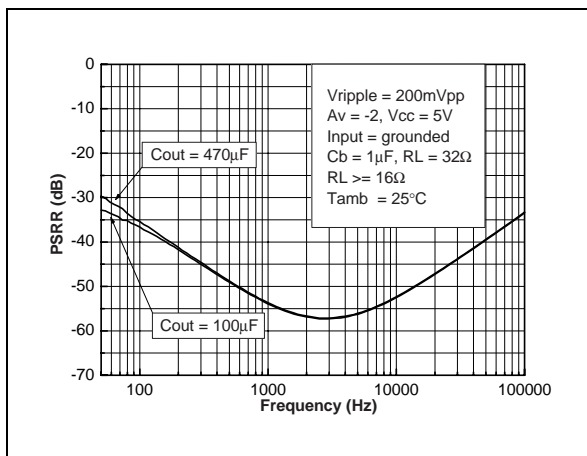


Fig. 80: THD + N vs Output Power

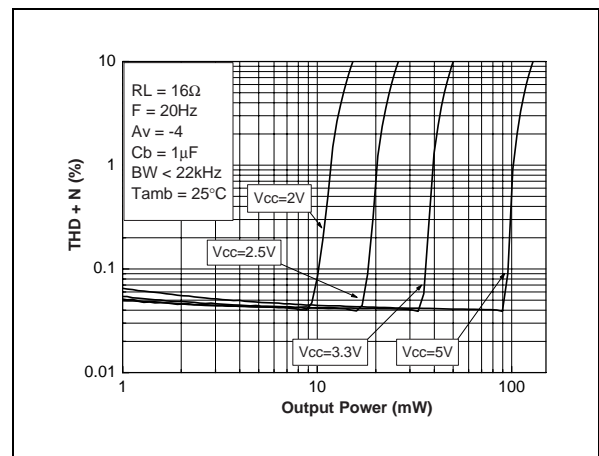


Fig. 81: THD + N vs Output Power

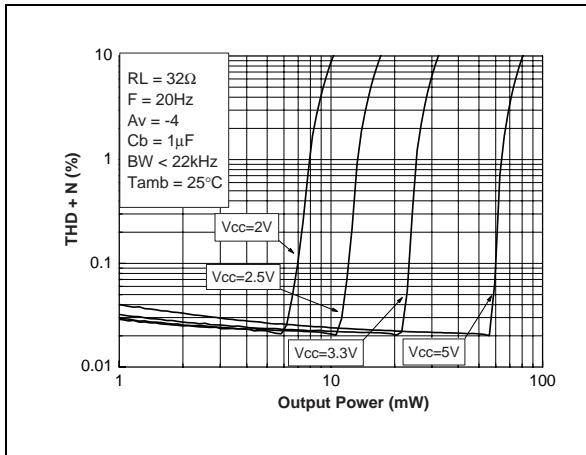


Fig. 82: THD + N vs Output Power

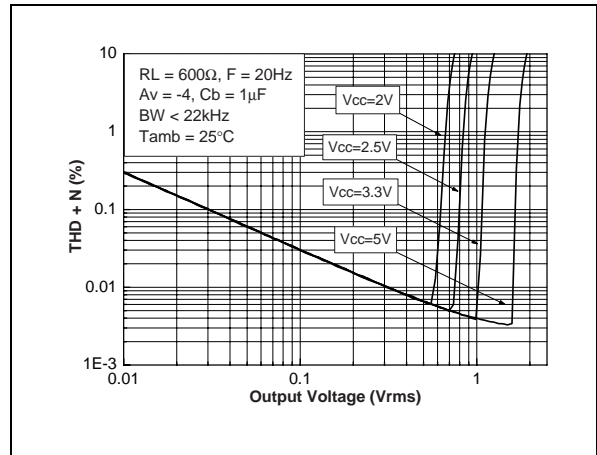


Fig. 83: THD + N vs Output Power

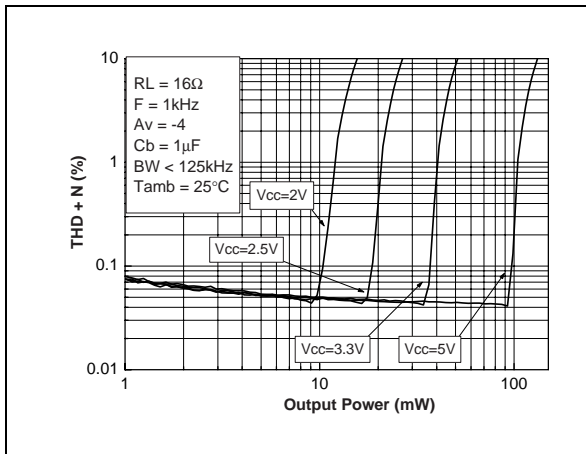


Fig. 84: THD + N vs Output Power

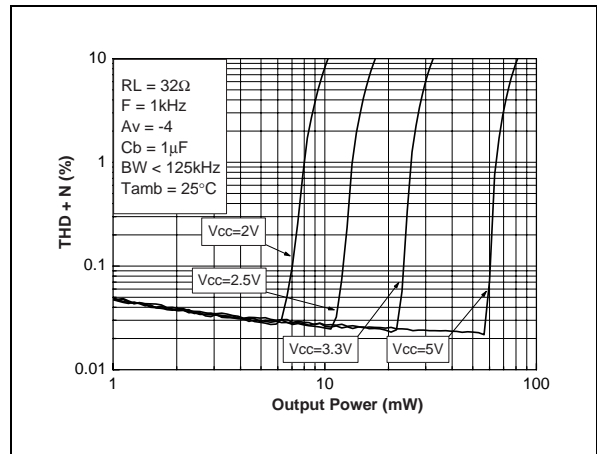


Fig. 85: THD + N vs Output Power

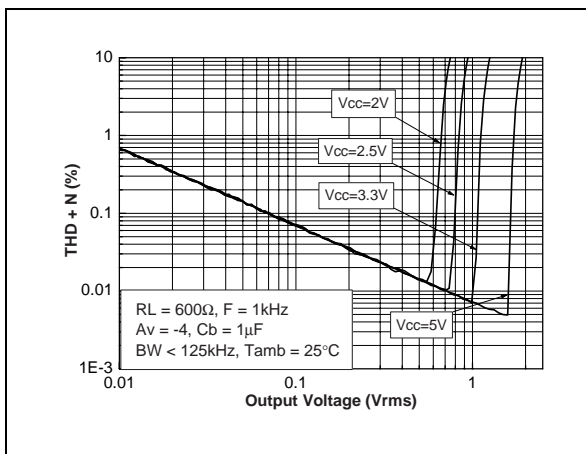


Fig. 86: THD + N vs Output Power

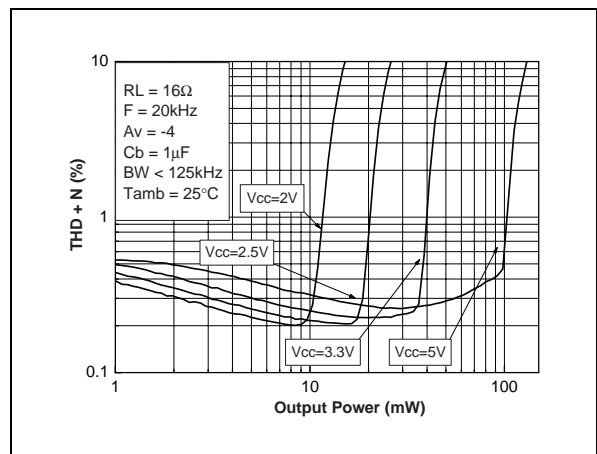


Fig. 87: THD + N vs Output Power

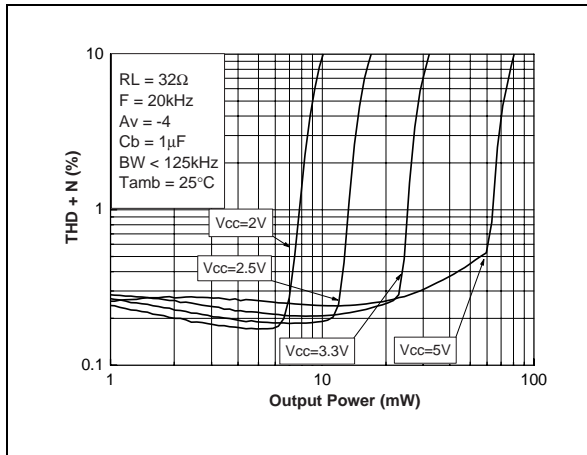


Fig. 88: THD + N vs Output Power

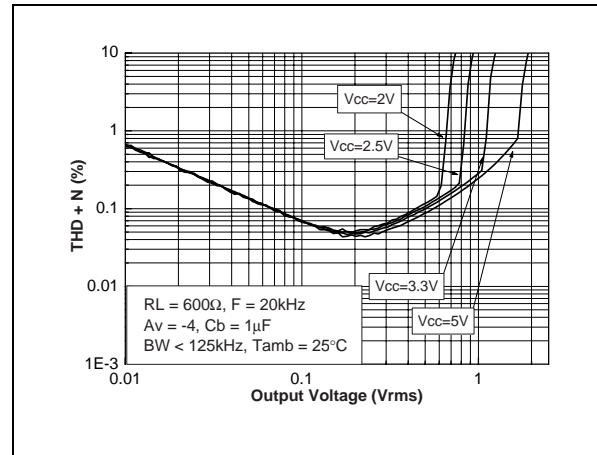


Fig. 89: THD + N vs Frequency

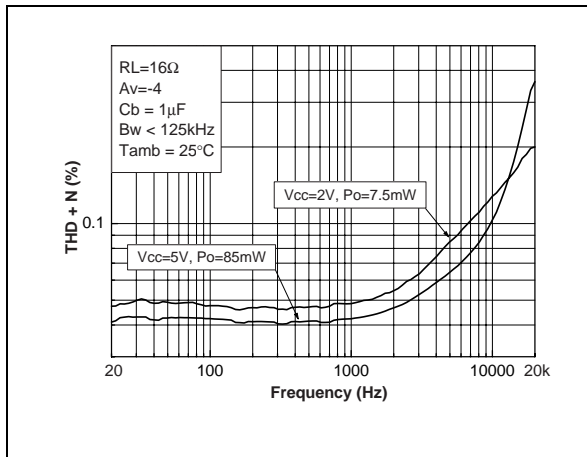


Fig. 90: THD + N vs Frequency

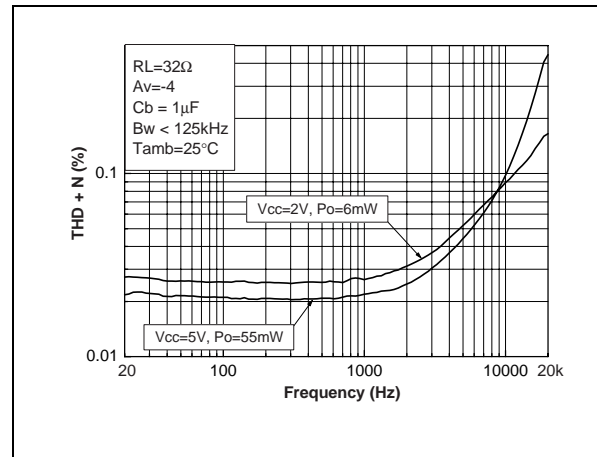


Fig. 91: THD + N vs Frequency

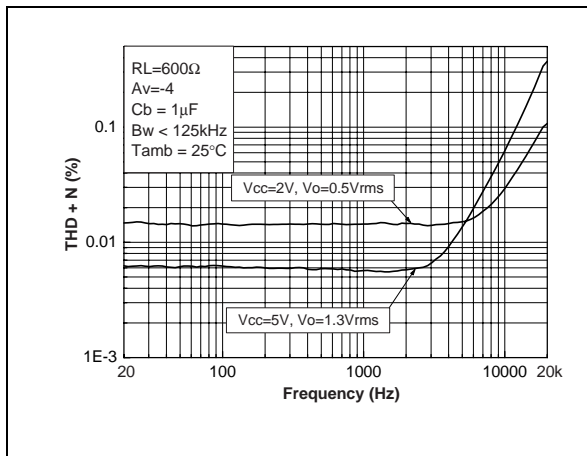


Fig. 92: Crosstalk vs Frequency

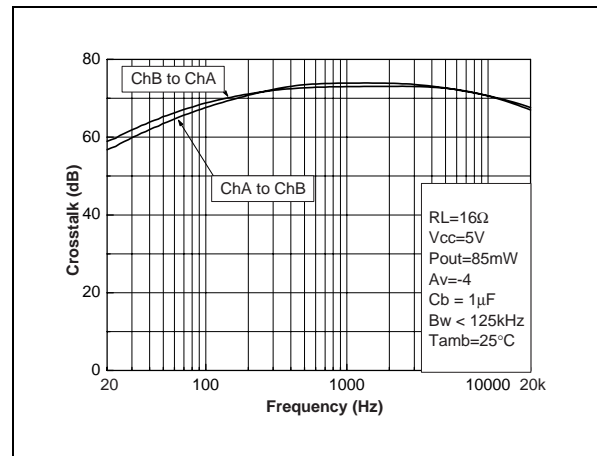


Fig. 93: Crosstalk vs Frequency

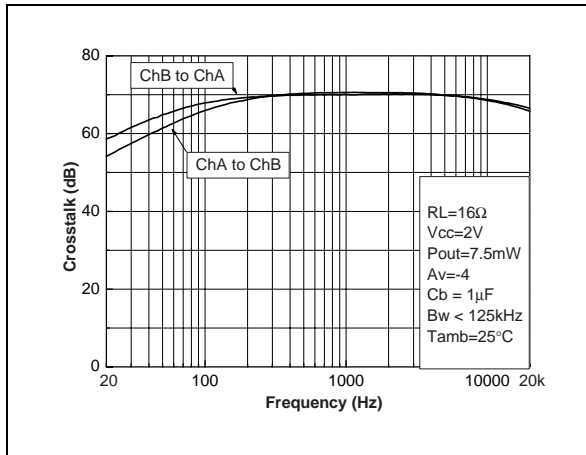


Fig. 94: Crosstalk vs Frequency

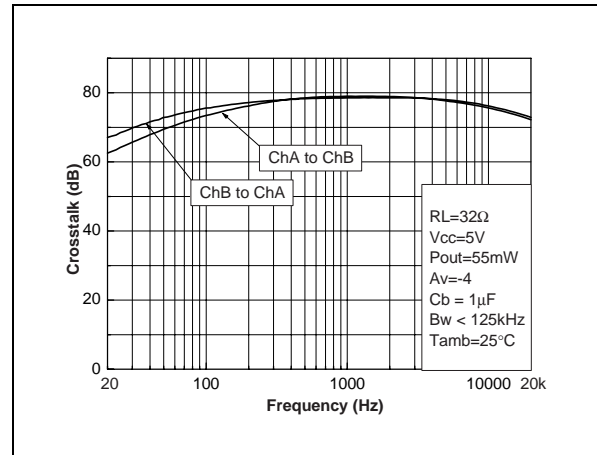


Fig. 95: Crosstalk vs Frequency

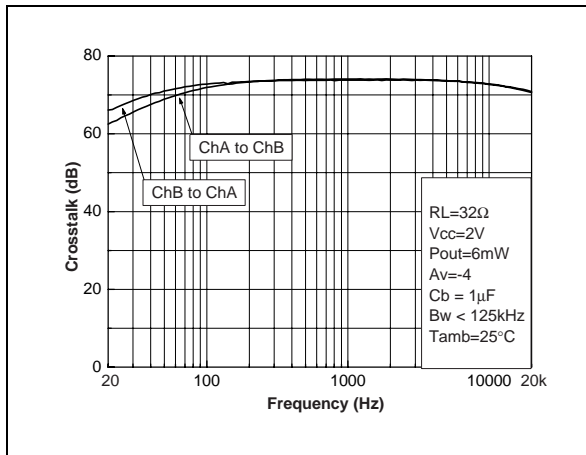


Fig. 96: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

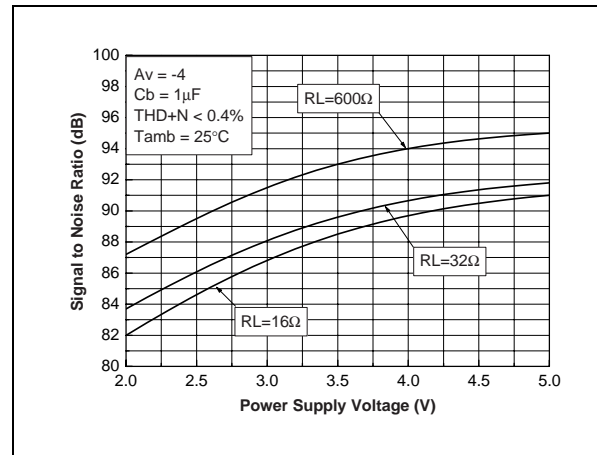


Fig. 97: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

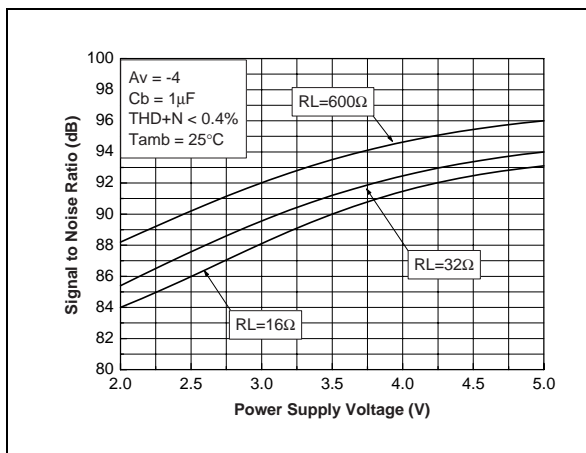


Fig. 98: PSRR vs Power Supply Voltage

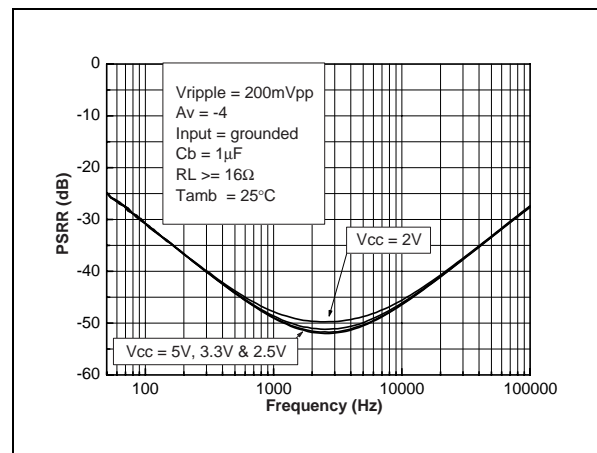


Fig. 99: PSRR vs Input Capacitor

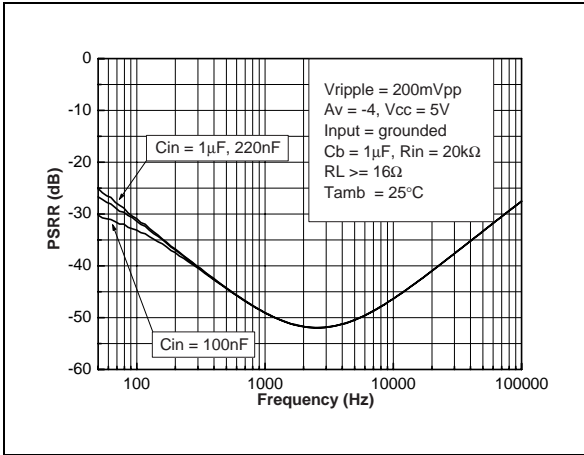


Fig. 100: PSRR vs Bypass Capacitor

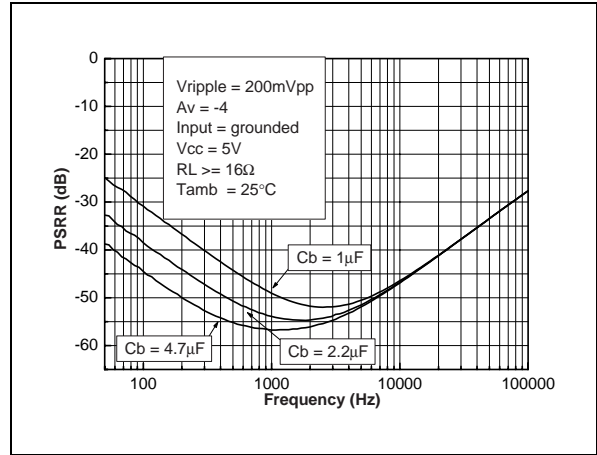


Fig. 101: PSRR vs Output Capacitor

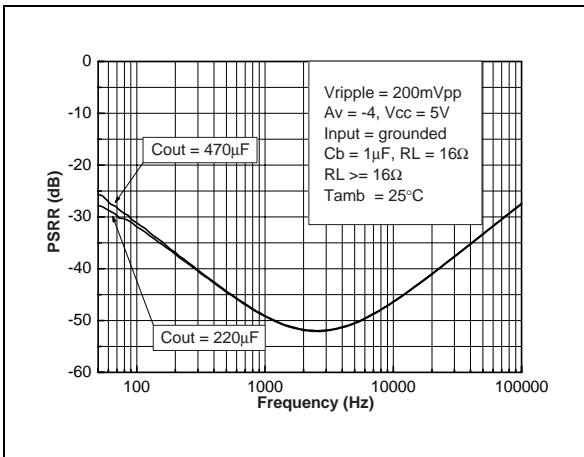
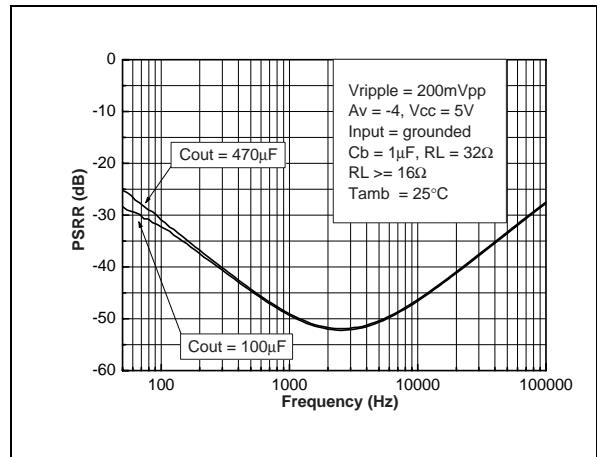
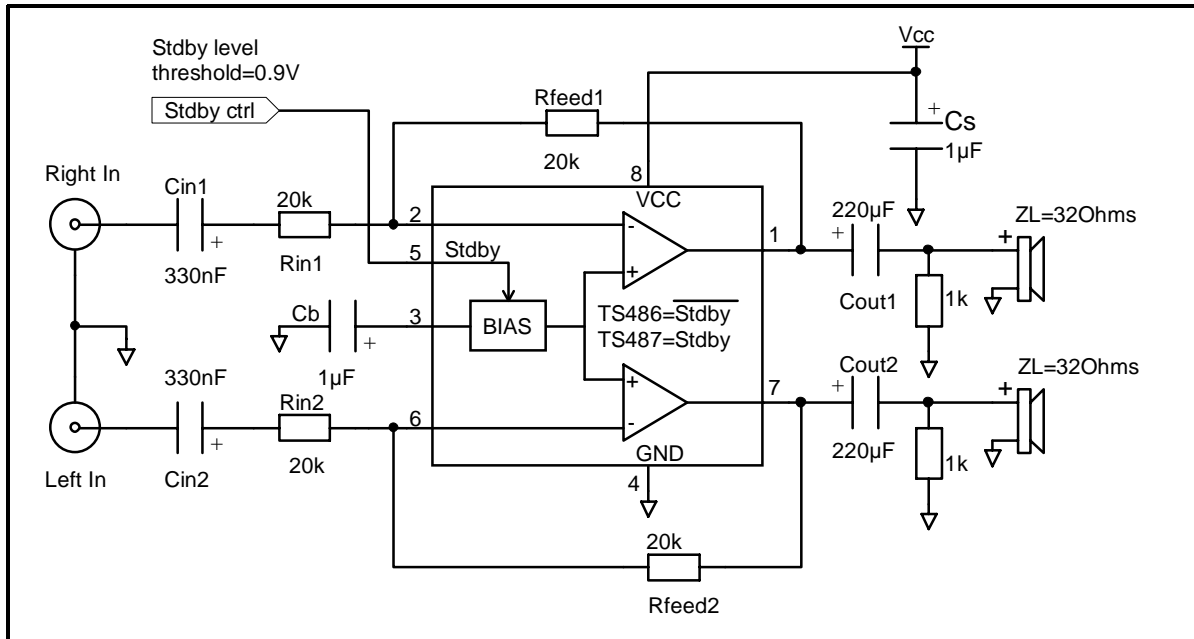


Fig. 102: PSRR vs Output Capacitor



APPLICATION NOTE:**TS486/487 GENERAL DESCRIPTION**

TS486/487 is a family of dual audio amplifiers able to drive 16Ω or 32Ω headsets. Working in the 2V to 5.5V supply voltage range, they deliver 100mW at 5V and 12mW at 2V in a 16Ω load. An internal output current limitation, offers protection against short-circuits at the output over a limited time period.

Fixed gain versions of the TS486 and TS487 including the feedback resistor and the input resistors are also proposed to reduce the number of external parts.

The TS486 and TS487 exhibit a low quiescent current of typically 1.8mA, allowing usage in portable applications.

The standby mode is selected using the SHUTDOWN input. For TS486 (respectively TS487), the device is in sleep mode when PIN 5 is connected at GND (resp. V_{CC}).

GAIN SETTING

The gain of each inverter amplifier of the TS486 and TS487 is set by the resistors R_{IN} and R_{FEED}.

$$\text{Gain}_{\text{LINEAR}} = -(R_{\text{FEED}}/R_{\text{IN}})$$

$$\text{Gain}_{\text{dB}} = 20 \text{ Log}(R_{\text{FEED}}/R_{\text{IN}})$$

Fixed gain versions TS486-n and TS487-n including R_{IN} and R_{FEED} are proposed to reduce external parts.

LOW FREQUENCY ROLL-OFF WITH INPUT CAPACITORS

The low roll-off frequency of the headphone amplifiers depends on the input capacitors C_{IN1} and C_{IN2} and the input resistors R_{IN1} and R_{IN2}.

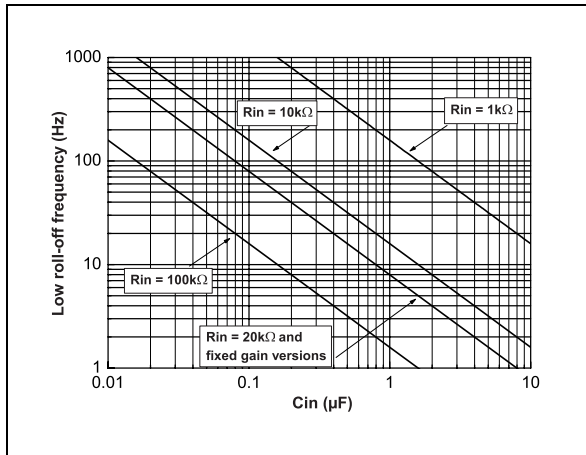
The C_{IN} capacitor in series with the input resistor R_{IN} of the amplifier is equivalent to a first order high pass filter.

Assuming that F_{min} is the lowest frequency to be amplified (with a 3dB attenuation), the minimum value of C_{IN} is:

$$C_{\text{IN}} > 1 / (2 * \pi * F_{\text{min}} * R_{\text{IN}})$$

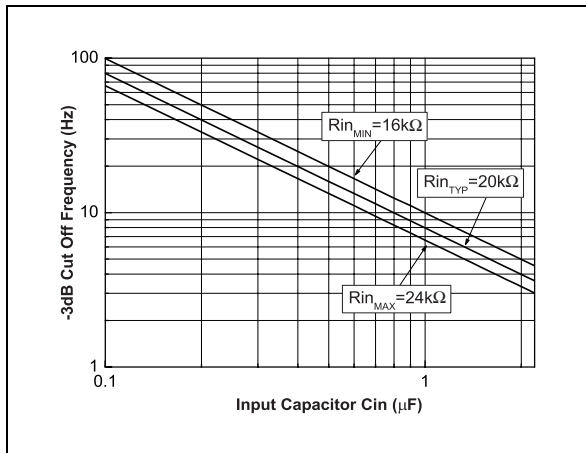
The following curve gives directly the low frequency roll-off versus the input capacitor C_{IN}

and for various values of the input resistor R_{IN} .



The input resistance of the fixed gain version is typically $20k\Omega$.

The following curve shows the limits of the roll off frequency depending on the min. and max. values of R_{in} :



LOW FREQUENCY ROLL OFF WITH OUTPUT CAPACITORS

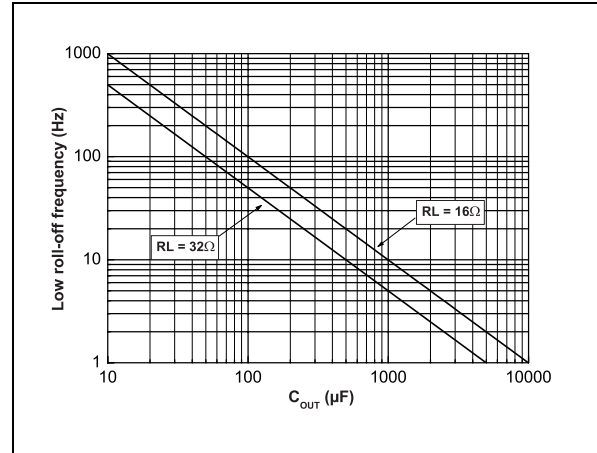
The DC voltage on the outputs of the TS486/487 is blocked by the output capacitors C_{OUT1} and C_{OUT2} . Each output capacitor C_{OUT} in series with the resistance of the load R_L is equivalent to a first order high pass filter.

Assuming that F_{min} is the lowest frequency to be amplified (with a 3dB attenuation), the minimum value of C_{OUT} is:

$$C_{OUT} > 1 / (2 * \pi * F_{min} * R_L)$$

The following curve gives directly the low roll-off

frequency versus the output capacitor C_{OUT} in μF and for the two typical 16Ω and 32Ω impedances:



DECOUPLING CAPACITOR C_B

The internal bias voltage at $V_{cc}/2$ is decoupled with the external capacitor C_B .

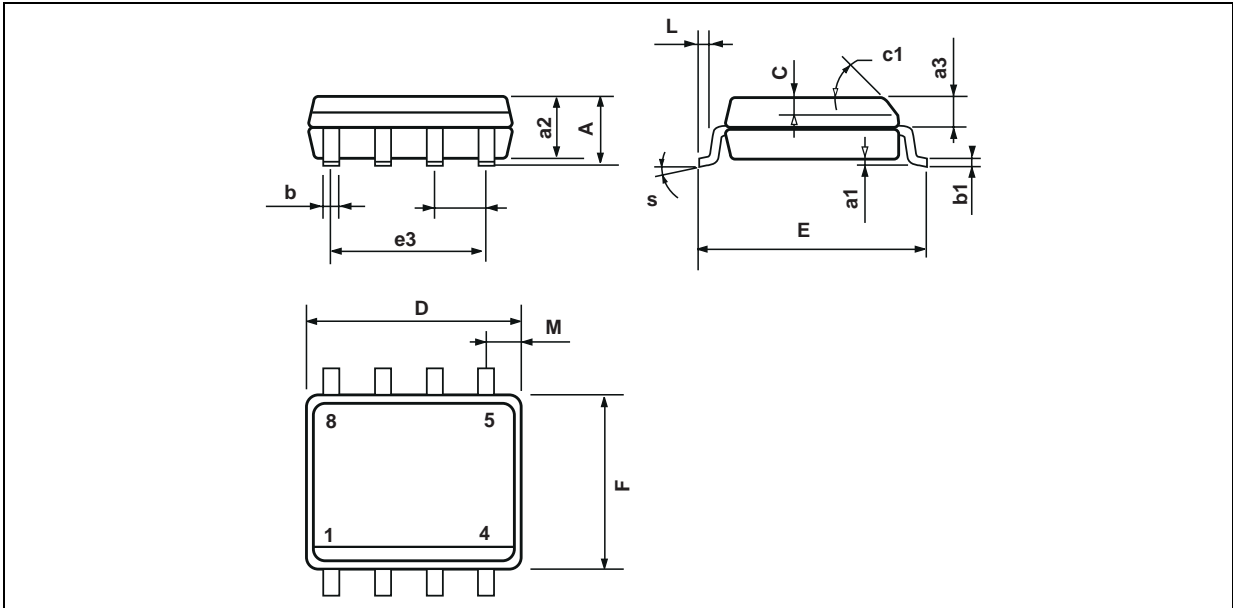
The TS486 and TS487 have a specified Power Supply Rejection Ratio parameter with $C_B = 1\mu F$. A higher value of C_B improves the PSRR, for example, a $4.7\mu F$ improves the PSRR by 15dB at 200Hz (please, refer to fig. 76 "PSRR vs Bypass Capacitor").

POP PRECAUTIONS

Generally headphones are connected using a connector as a jack. To prevent a pop in the headphones when plugged in the jack, a resistor should be connected in parallel with each headphone output. This allows the capacitors C_{out} to be charged even when no headphone is plugged.

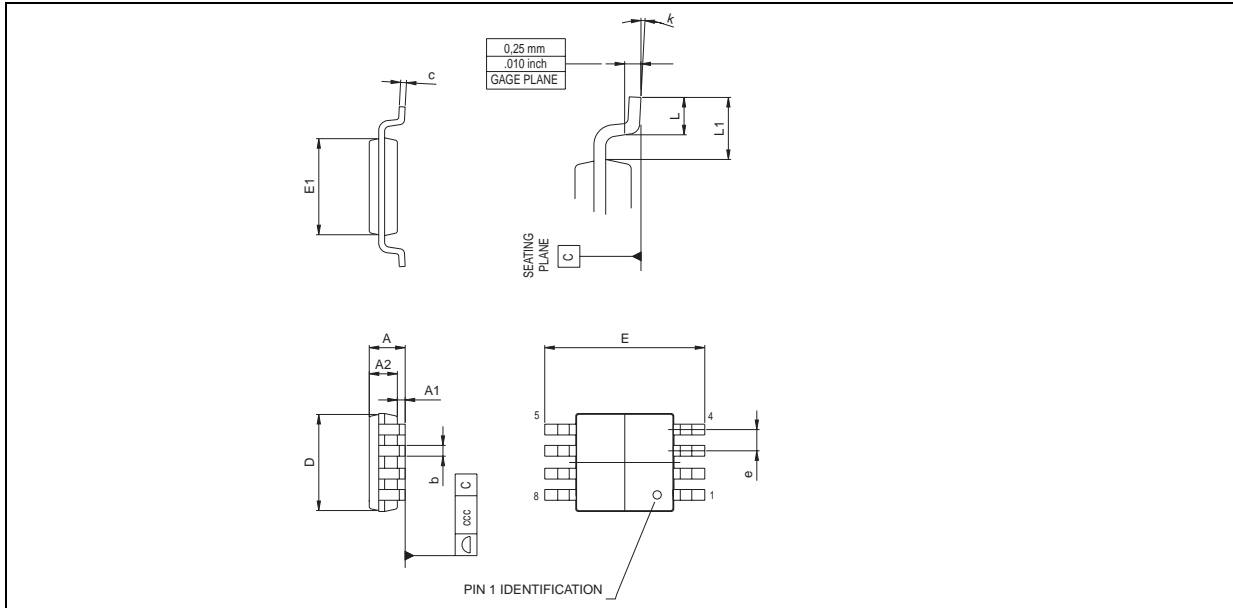
A resistor of $1k\Omega$ is high enough to be a negligible load, and low enough to charge the capacitors C_{out} in less than one second.

PACKAGE MECHANICAL DATA
8 PINS - PLASTIC MICROPACKAGE (SO)

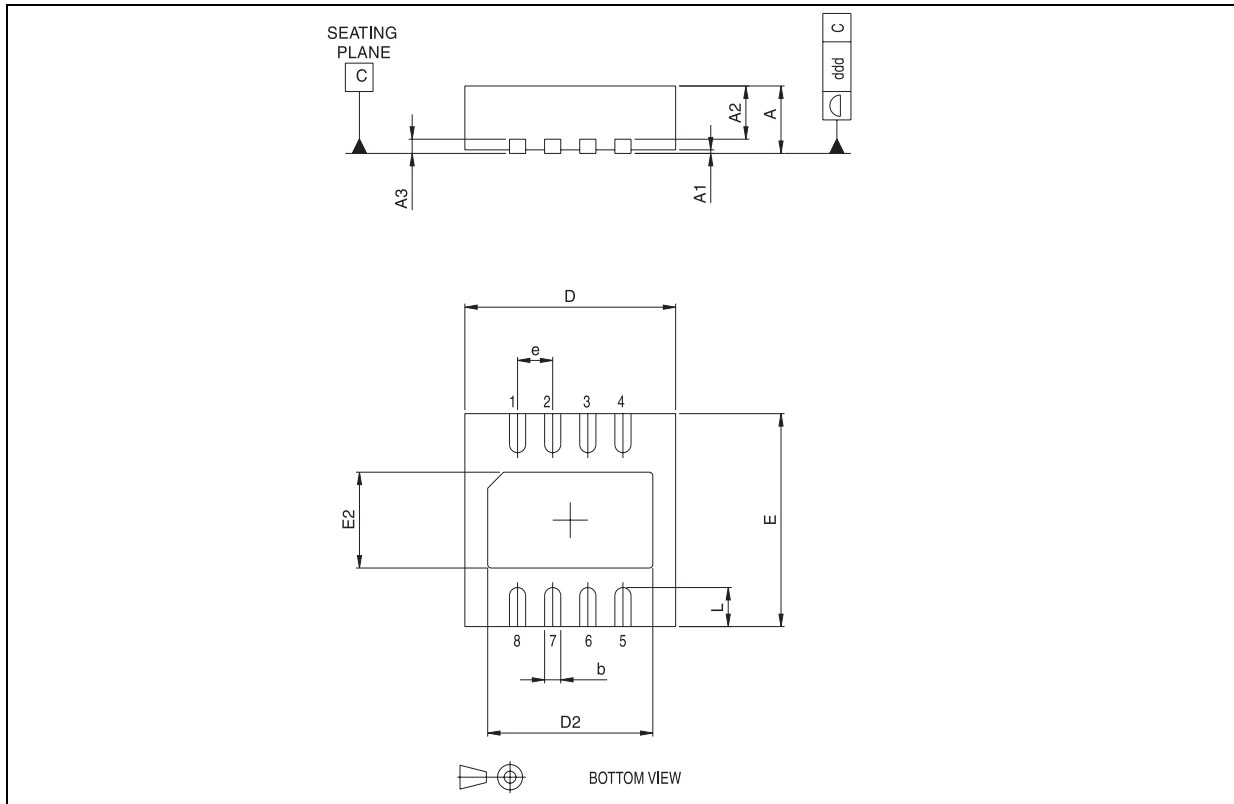


Dim.	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.069
a1	0.1		0.25	0.004		0.010
a2			1.65			0.065
a3	0.65		0.85	0.026		0.033
b	0.35		0.48	0.014		0.019
b1	0.19		0.25	0.007		0.010
C	0.25		0.5	0.010		0.020
c1	45° (typ.)					
D	4.8		5.0	0.189		0.197
E	5.8		6.2	0.228		0.244
e		1.27			0.050	
e3		3.81			0.150	
F	3.8		4.0	0.150		0.157
L	0.4		1.27	0.016		0.050
M			0.6			0.024
S	8° (max.)					

PACKAGE MECHANICAL DATA
8 PINS - PLASTIC MICROPACKAGE (miniSO)



Dim.	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.100			0.043
A1	0.050	0.100	0.150	0.002	0.004	0.006
A2	0.780	0.860	0.940	0.031	0.034	0.037
b	0.250	0.330	0.400	0.010	0.013	0.016
c	0.130	0.180	0.230	0.005	0.007	0.009
D	2.900	3.000	3.100	0.114	0.118	0.122
E	4.750	4.900	5.050	0.187	0.193	0.199
E1	2.900	3.000	3.100	0.114	0.118	0.122
e		0.650			0.026	
L	0.400	0.550	0.700	0.016	0.022	0.028
L1		0.950			0.037	
k	0d	3d	6d	0d	3d	6d
aaa			0.100			0.004

PACKAGE MECHANICAL DATA**8 CONNECTIONS - Dual Micro Leadframe Package (QFN)**

Dimensions	Millimeters		
	Min.	Typ.	Max.
A	0.80	0.90	1.00
A1		0.02	0.05
A2		0.70	
A3		0.20	
b	0.18	0.23	0.30
D		3.00	
D2	2.20	2.35	2.45
E		3.00	
E2	1.20	1.35	1.45
e		0.50	
L	0.45	0.55	0.65
ddd			0.08

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