



155Mbps/622Mbps Clock Recovery and Data Retiming IC with Fully Integrated Phase/Frequency Detector

MAX3270

General Description

The MAX3270 is a complete Clock Recovery and Data Retiming IC for 155Mbps and 622Mbps SDH/SONET and ATM applications. The MAX3270 meets Bellcore and CCITT jitter tolerance specifications ensuring error-free data recovery. Recovered clock and data are phase aligned using a fully integrated phase-locked loop (PLL). An output frequency monitor (FM) is included to detect loss of PLL acquisition or a loss of input data.

The MAX3270 has differential ECL input and output interfaces, so it is less susceptible to noise in a high-frequency environment. The fully integrated PLL includes an integrated phase-frequency detector that eliminates the need for external references.

Applications

155Mbps (STM-1/OC-3)/622Mbps (STM-4/OC-12) SDH/SONET Transmission Systems
155Mbps/622Mbps ATM/SONET Access Nodes
Add/Drop Multiplexers
Cross-Connects

Features

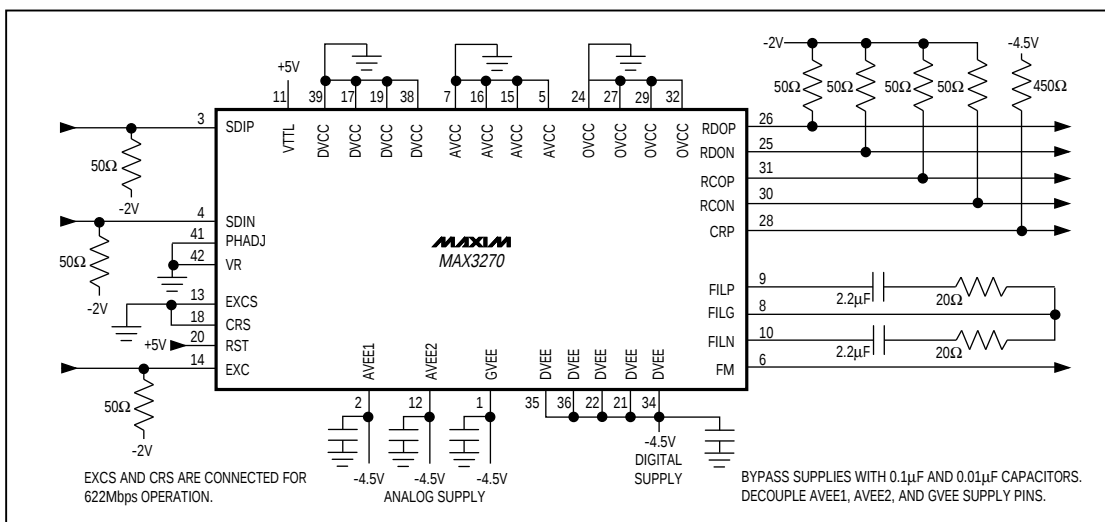
- ♦ Supports Both 155Mbps and 622Mbps Clock Recovery and Data Retiming
- ♦ Fully Integrated Phase/Frequency Detector
- ♦ Capable of Switching to an External Clock
- ♦ Differential 100K ECL Data and Clock I/Os
- ♦ Output Monitor Provides Lock Detection
- ♦ No External Reference Clock Required

Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
MAX3270EMH	-40°C to +85°C	44 MQFP

Pin Configuration appears at end of data sheet.

Typical Operating Circuit



Maxim Integrated Products 1

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ABSOLUTE MAXIMUM RATINGS

Supply Voltages		
VTTL to GND	-0.5V to +8.0V	PHADJ, VR
VCC to GND	-0.5V to +8.0V	FM
VEE to GND	-8.0V to +0.5V	Input Differential Voltage Level, SDIP, SDIN
SDIP, SDIN, EXC	-8.0V to +0.5V	Continuous Power Dissipation (T _A = +85°C)
RDOP, RDON, RCOP, RCON, CRP	-8.0V to +0.5V	Operating Temperature Range
EXCS, RST, CRS	-0.5V to +8.0V	Storage Temperature Range
FILP, FILG, FILN	-8.0V to +0.5V	Lead Temperature (soldering, 10sec)

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{EE} = -4.5V ±5%, VTTL = 5V ±5%, T_A = -40°C to +85°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Positive Voltage Supply (with respect to ground)	V _{TTL}		4.75	5.00	5.25	V
Negative Voltage Supply (with respect to ground)	V _{EE}		-4.725	-4.50	-4.275	V
Static Supply Current from VTTL	I _{TTL}			2.4	5	mA
Static Supply Current from V _{EE}	I _{VEE}			150	210	mA
ECL INPUTS: EXC, SDIP, SDIN						
Input High Voltage	V _{IH}		-1165		-870	mV
Input Low Voltage	V _{IL}		-1830		-1475	mV
Input High Current	I _{IH}	V _{IN} = V _{OH} (typ)	0		100	μA
Input Low Current	I _{IL}	V _{IN} = V _{OL} (typ)	-100		100	nA
ECL OUTPUTS: RCOP, RCON, RDOP, RDON						
Output High Voltage	V _{OH}	Loaded with 50Ω to -2V	-1025	-955	-870	mV
Output Low Voltage	V _{OL}	Loaded with 50Ω to -2V	-1830	-1705	-1550	mV
LOW-POWER ECL OUTPUT: CRP						
Output High Voltage	V _{OH}	Loaded with 470Ω to V _{EE}	-1025	-955	-870	mV
Output Low Voltage	V _{OL}	Loaded with 470Ω to V _{EE}	-1830	-1705	-1620	mV
TTL INPUTS: CRS, RST, EXCS						
Input High Voltage	V _{IH}		2			V
Input Low Voltage	V _{IL}			0.8		V
Input High Current	I _{IH}	VTTL = 5.00V, V _{IN} = 2V	0		40	μA
Input Low Current	I _{IL}	VTTL = 5.00V, V _{IN} = 0.8V	0		40	μA
PHASE ADJUST INPUTS: PHADJ, VR						
Input Bias Current	I _{BIAS}	VR = PHADJ = 0, T _A = +25°C	0		10	μA

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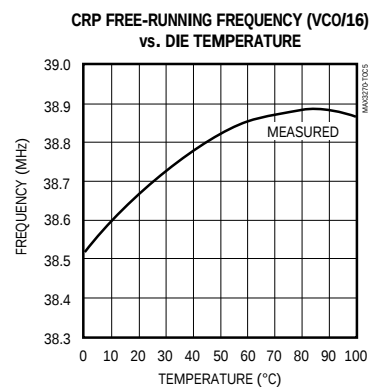
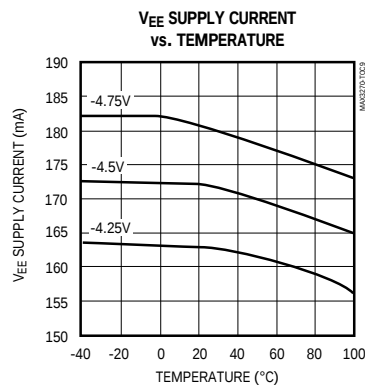
MAX3270

AC ELECTRICAL CHARACTERISTICS (continued)

(V_{EE} = -4.5V, V_{TTL} = 5V, T_A = 25°C, unless otherwise noted.)

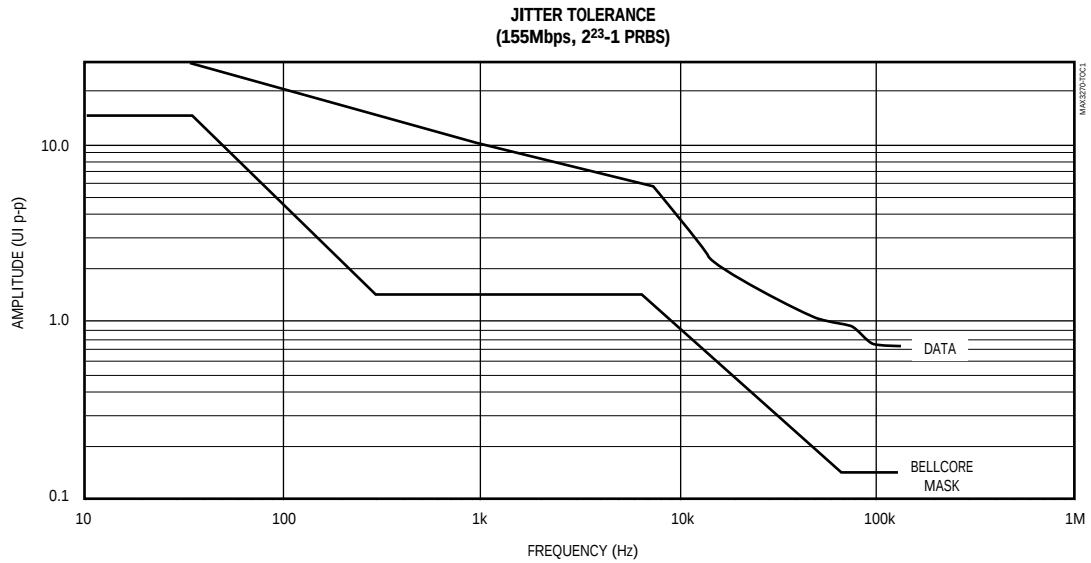
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
ECL OUTPUTS: RDOP, RDON, RCOP, RCON						
Transition Time 20% to 80%	t _r , t _f	Loaded with 50Ω to -2V and 5pF to GND		600		ps
Time Difference between RDO and RCO	TD	Loaded with 50Ω to -2V and 5pF to GND		100		ps
PFD AND FILTER AMPLIFIER TEST LEVELS						
Output Offset Voltage of the Monitor Amplifier	V _O	PHADJ = 0, FILP and FILN shorted	-35		35	mV
Gain of the Monitor Amplifier	GFM	PHADJ = 0	0.95		1.05	V/V
Filter Amplifier Open-Loop Voltage Gain	GOL	FILP and FILN open	21	26		dB
VCO TEST PARAMETERS; CPR OUTPUT						
Center Frequency	F _O	FILP and FILN shorted, PFD = neutral state	38.00		39.50	MHz
Frequency Range	DF _O	FILP - FILN = 1.6V	6		10	MHz
Mean Frequency Sensitivity	K _O	FILP - FILN = 1.6V	3.75		6	MHz/V
Frequency Sensitivity to Power-Supply Voltage	K _{OV}	FILP and FILN shorted			550	kHz/V
PLL ELECTRICAL SPECIFICATIONS						
Frequency of VCO	F _O			622.08		MHz
Incremental Tuning Sensitivity (Incremental Slope, Δf/ΔV _t)	K _O	f _t = 622.08MHz		75		MHz/V
Phase-Detector Gain	K _D			192		mV/rad
Transconductance Gain of Filter Amplifier	G _m			1.25		mA/V
Phase Offset Sensitivity, ΔΦ/ΔPHADJ	K _{PHADJ}			2		rad/V

Typical Operating Characteristics



155Mbps/622Mbps Clock Recovery and Data Retiming IC with Fully Integrated Phase/Frequency Detector

Typical Operating Characteristics

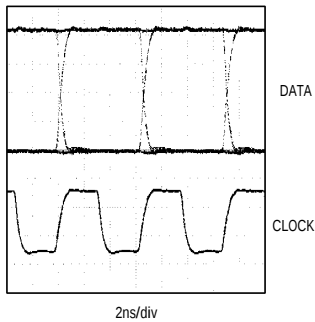


155Mbps/622Mbps Clock Recovery and Data Retiming IC with Fully Integrated Phase/Frequency Detector

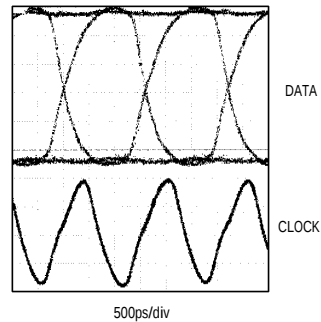
Typical Operating Characteristics

MAX3270

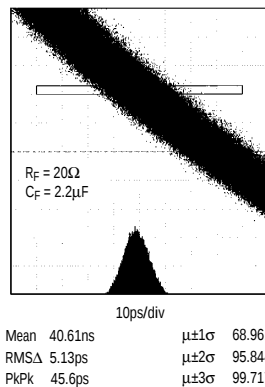
155Mbps RECOVERED CLOCK AND RETIMED DATA (SINGLE ENDED)



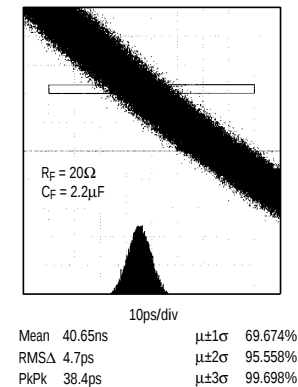
622Mbps RECOVERED CLOCK AND RETIMED DATA (SINGLE ENDED)



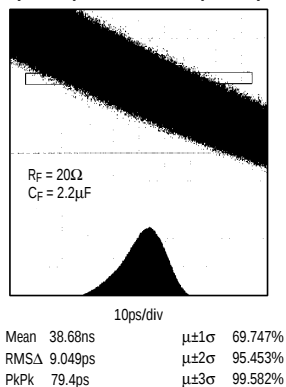
RECOVERED CLOCK JITTER
(155Mbps, 2⁷-1 PRBS, 5.1ps RMS)



RECOVERED CLOCK JITTER
(155Mbps, 1-0 PATTERN, 4.7ps RMS)



RECOVERED CLOCK JITTER
(622Mbps 2⁷-1 PRBS 9.0ps RMS)



155Mbps/622Mbps Clock Recovery and Data Retiming IC with Fully Integrated Phase/Frequency Detector

Pin Description

PIN	NAME	FUNCTION
1	GVEE	Guard-Ring Negative Supply to Substrate: -4.5V
2	AVEE1	Negative Supply for Input Buffers: -4.5V
3	SDIP	Serial Data Input: 155Mbps or 622Mbps. Differential ECL Positive.
4	SDIN	Serial Data Input: 155Mbps or 622Mbps. Differential ECL Negative.
5	AVCC	Ground for Input Buffers: 0V
6	FM	Frequency Monitor Output. This pin monitors the input voltage to the VCO. When the PLL is locked, the pin will be $\approx 0V$.
7	AVCC	Guard-Ring Positive Supply to Epi: 0V
8	FILG	Loop Filter Ground. This pin connects to an external filter.
9	FILP	Loop Filter Positive. This pin connects to an external filter.
10	FILN	Loop Filter Negative. This pin connects to an external filter.
11	VTTL	TTL Positive Supply: +5.0V
12	AVEE2	Negative Supply for VCO: -4.5V
13	EXCS	External Clock-Select TTL Input. A logical high selects the external clock.
14	EXC	External Clock. Single-ended ECL input.
15, 16	AVCC	Ground for VCO: 0V
17, 19, 38, 39	DVCC	Digital Ground for Mux: 0V
18	CRS	Clock-Rate Select TTL Input. This selects the clock rate to be either 155Mbps or 622Mbps. A logic-high level selects the 622Mbps mode.
20	RST	Resets all digital flip-flops, TTL input. Reset is assert when low.
21, 22, 34, 35, 36	DVEE	Digital Negative Supply: -4.5V
23, 33, 37, 40, 43, 44	N.C.	No Connection
24, 27, 29, 32	OVCC	Output Driver Ground: 0V
25	RDON	Negative Recovered Data Output, differential ECL output: 155Mbps or 622Mbps.
26	RDOP	Positive Recovered Data Output, differential ECL output: 155Mbps or 622Mbps.
28	CRP	Clock-Reference Output Divide-by-4. ECL low-power single-ended: 38Mbps or 155Mbps.
30	RCON	Negative Recovered Clock Output, differential ECL output: 155Mbps or 622Mbps.
31	RCOP	Positive Recovered Clock Output, differential ECL output: 155Mbps or 622Mbps.
41	PHADJ	Phase Adjust. This is an analog adjustment that varies the static phase between the input data and the recovered clock. If not used, this input should be grounded. The range is from -1V to 1V.
42	VR	Phase Reference Voltage: 0V. The PHADJ pin compares to this voltage. Set to ground.

MAX3270

bandwidth. The input data stream is sampled by quadrature components of the VCO clock, generating a difference frequency. Depending on the rotation of the difference frequency, the PFD will drive the VCO so that the difference frequency is driven to zero. Once frequency acquisition is obtained, the frequency detector will return to a neutral state.

The PLL is a second-order transfer function whose bandwidth is set by the loop filter. The VCO is integrated into the PLL and always operates at 622MHz. The center frequency is tightly controlled by laser trimming, limiting frequency drift when lock is lost. 155Mbps or 622Mbps mode is selected by the clock-rate select (CRS) pin. CRS selects the inputs to multiplexer MUX2.

A frequency detector is also incorporated into the PLL. Frequency detection aids in the acquisition of the input data; this frequency-aided acquisition is necessary during start-up conditions, since the input data stream and VCO difference frequency may be outside the PLL

The internal VCO can be bypassed with an external clock applied to the EXC input. The external clock select (EXCS) controls the input selections to multiplexers MUX1 and MUX2.



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Design Procedure

Selecting the Data Rate

The MAX3270 is intended for use in SDH/SONET systems operating at 155.52Mbps or 622.08Mbps data rates. TTL inputs (CRS and EXCS) are provided for selecting the recovered clock rate (Table 1). It is also possible to switch to an externally supplied clock by enabling the EXC input. The EXC input is a high-speed single-ended ECL interface capable of handling serial clock rates of 155MHz and 622MHz.

Table 1. MAX3270 Logic Table

EXCS	CRS	RCOP/RCON	CRP
0	1	155.52Mbps	38.88Mbps
0	0	622.08Mbps	155.52Mbps
1	0	EXC	EXC/4
1	1	EXC/4	EXC/16

Setting the Loop Filter

The loop filter within the PLL consist of a transconductance amplifier and the external filter elements R_f and C_f (Figure 2). The closed-loop bandwidth of a PLL can be approximated by:

$$K_D K_O G_m R_f$$

where K_D is the gain of the phase detector, K_O is the gain of the VCO, and G_m is the transconductance of the filter amplifier. Because this filter is an integrator, a zero in the open-loop gain is required for stability. This zero is set by the following equation:

$$\omega_z = 1 / (R_f C_f)$$

where the recommended external values are $R_f = 20\Omega$ and $C_f = 2.2\mu F$. To decrease the PLL's closed-loop bandwidth, reduce the value of R_f . Decreasing this bandwidth will improve the MAX3270's jitter transfer performance but reduce jitter tolerance. The MAX3270 has been designed (using the recommended values of R_f and C_f) to meet the Bellcore and CCITT specifications for jitter tolerance of a Network Element. Carefully consider the application if a reduction in loop bandwidth is desired. By reducing R_f an order of magnitude, the PLL's bandwidth becomes more sensitive to the internal tolerances of the IC. As a result, the loop bandwidth may have a wider variation. If R_f is reduced, then C_f should also be increased to maintain loop stability and minimize jitter peaking.

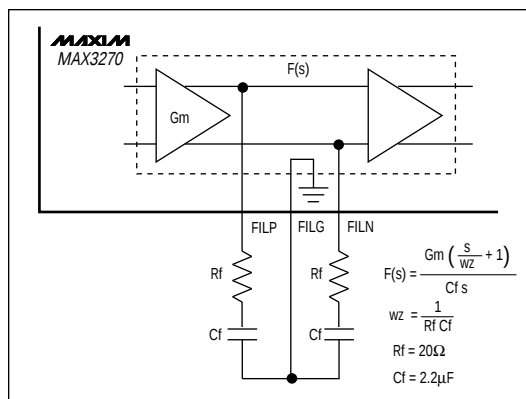


Figure 2. Loop Filter

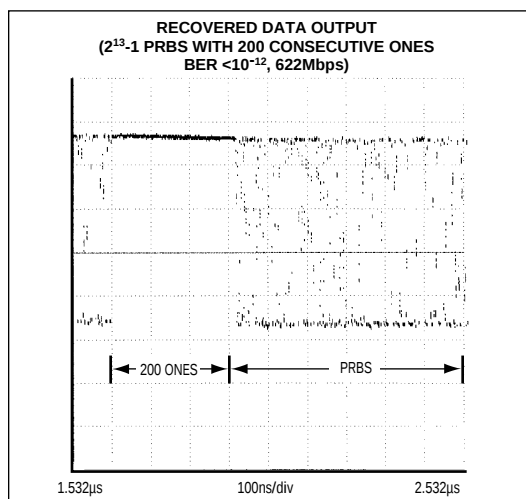


Figure 3. Recovered Data Output

The MAX3270 is optimally designed to acquire lock and to provide a bit-error rate (BER) of less than 10^{-12} for long strings of consecutive zeros or ones. Using the recommended external values for $R_f = 20\Omega$ and $C_f = 2.2\mu F$, measured results show that the MAX3270 can tolerate more than 200 consecutive ones or zeros. Figure 3 shows a bit stream of $2^{13} - 1$ PRBS with 200 consecutive ones.

155Mbps/622Mbps Clock Recovery and Data Retiming IC with Fully Integrated Phase/Frequency Detector

MAX3270

Input and Output Termination

The MAX3270 data and clock I/Os (SDIP, SDIN, RDOP, RDON, RCOP, RCON, and EXC) are open emitters, designed to interface with ECL signal levels. It is important to bias these ports appropriately. A circuit that provides a Thevenin equivalent of 50Ω to $-2V$ should be used with fixed-impedance transmission lines for proper termination. Figure 4 shows some typical input and output termination methods.

The serial data input signals (SDIP and SDIN) are the differential inputs to an emitter coupled pair. As a result, the MAX3270 can accept differential input signal levels as low as 250mV. The serial input (SDIP) can also be driven single-ended by externally biasing SDIN to the center of the voltage swing (approximately $-1.3V$). Make sure that the differential inputs and outputs each see the same termination impedance for balanced operation.

CRP is also an open-emitter ECL output, but it requires a termination resistor of 450Ω to $-4.5V$. If this output is not used, reduce power by connecting CRP to V_{EE} through a resistor valued at $10k\Omega$ or more.

The MAX3270's performance can be greatly affected by circuit board layout and design. Use good high-frequency design techniques, including minimizing ground inductances and using fixed-impedance transmission lines on the data and clock signals. Power-

supply decoupling should be placed as close to the V_{EE} and V_{TTL} pins as possible. $AV_{EE}1$, $AV_{EE}2$ and GV_{EE} should each have their own bypass/decoupling elements, independent of each other and any other $-4.5V$ supply. Make sure to isolate the inputs from the outputs to reduce feedthrough.

Applications Information

Lock Detection

The MAX3270 has an output (FM) that monitors the input voltage to the VCO. FM is an analog output that can be used as a flag to indicate that the PLL is locked. Under normal operation, the loop is locked and the FM output is approximately equal to $0V$. When the PLL is unlocked, the VCO will drift. The FM output monitors this drift and will equal approximately $\pm 1V$ in the limit.

Phase Adjust

In some applications, the optimum alignment point between the recovered clock and the serial data is not at the center of the eye diagram. The MAX3270 has a PHADJ input that can be used in these applications to introduce a phase difference between the recovered clock and the serial data. When no phase difference is desired, this input should be set to $0V$. The VR pin is the reference input for PHADJ and is normally tied to GND.

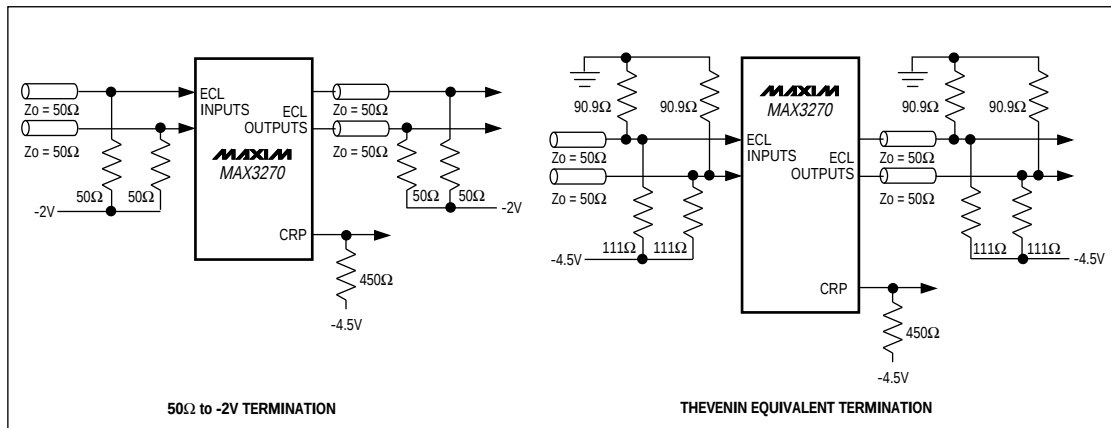
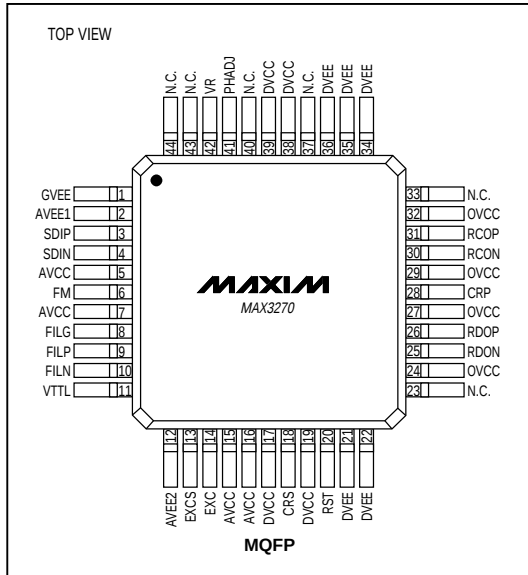


Figure 4. Typical Input and Output Terminations

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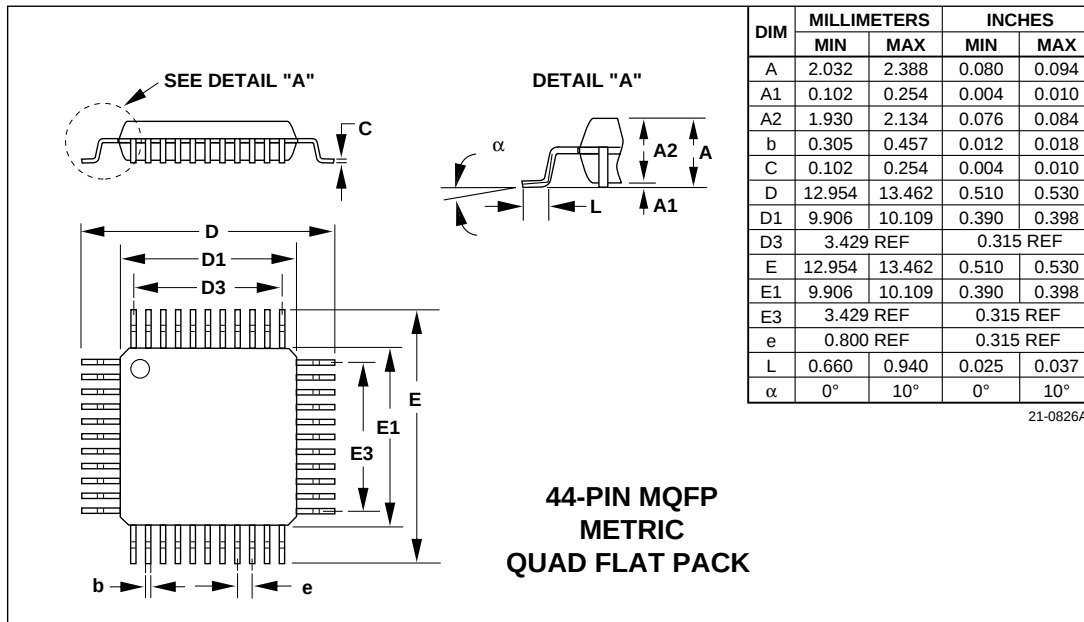
Pin Configuration



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Package Information

MAX3270



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