

CMOS, Rail-to-Rail, I/O OPERATIONAL AMPLIFIERS

FEATURES

- **RAIL-TO-RAIL INPUT AND OUTPUT**
- **WIDE SUPPLY RANGE:**
Single Supply: 4V to 12V
Dual Supplies: ± 2 to ± 6
- **LOW QUIESCENT CURRENT: 160 μ A**
- **FULL-SCALE CMRR: 90dB**
- **LOW OFFSET: 160 μ V**
- **HIGH SPEED:**
OPA703: 1MHz, 0.6V/ μ s
OPA704: 3MHz, 3V/ μ s
- **MicroSIZE PACKAGES:**
SOT23-5, MSOP-8, TSSOP-14
- **LOW INPUT BIAS CURRENT: 1pA**

APPLICATIONS

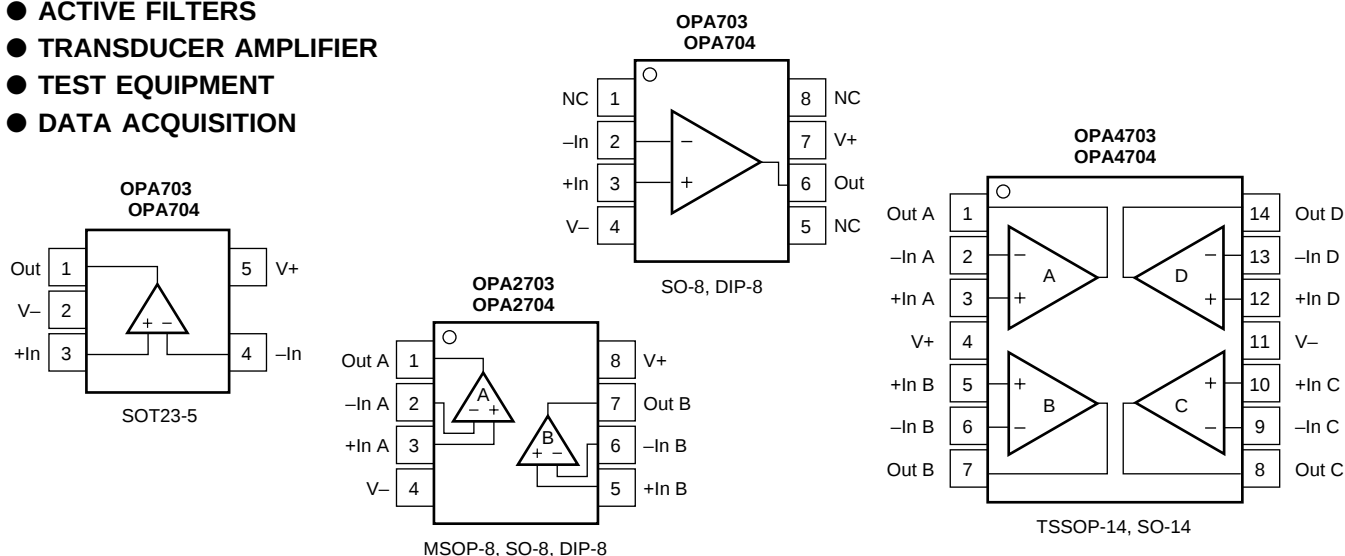
- **AUTOMOTIVE APPLICATIONS:**
Audio, Sensor Applications, Security Systems
- **PORTABLE EQUIPMENT**
- **ACTIVE FILTERS**
- **TRANSDUCER AMPLIFIER**
- **TEST EQUIPMENT**
- **DATA ACQUISITION**

DESCRIPTION

The OPA703 and OPA704 series op amps are optimized for applications requiring rail-to-rail input and output swing. Single, dual, and quad versions are offered in a variety of packages. While the quiescent current is less than 200 μ A per amplifier, the OPA703 still offers excellent dynamic performance (1MHz GBW and 0.6V/ μ s SR) and unity-gain stability. The OPA704 is optimized for gains of 5 or greater and provides 3MHz GBW and 3V/ μ s slew rate.

The OPA703 and OPA704 series are fully specified and guaranteed over the supply range of ± 2 V to ± 6 V. Input swing extends 300mV beyond the rail and the output swings to within 40mV of the rail.

The single versions (OPA703 and OPA704) are available in the MicroSIZE SOT23-5 and in the standard SO-8 surface-mount, as well as the DIP-8 packages. Dual versions (OPA2703 and OPA2704) are available in the MSOP-8, SO-8, and DIP-8 packages. The quad OPA4703 and OPA4704 are available in the TSSOP-14 and SO-14 packages. All are specified for operation from -40°C to $+85^{\circ}\text{C}$.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

Supply Voltage, V+ to V-.....	13.2V
Signal Input Terminals, Voltage ⁽²⁾	(V-) -0.3V to (V+) +0.3V
Current ⁽²⁾	10mA
Output Short-Circuit ⁽³⁾	Continuous
Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C
Junction Temperature.....	+150°C
Lead Temperature (soldering, 10s).....	+300°C

NOTES: (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. (2) Input terminals are diode-clamped to the power supply rails. Input signals that can swing more than 0.3V beyond the supply rails should be current-limited to 10mA or less. (3) Short-circuit to ground, one amplifier per package.



ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PACKAGE/ORDERING INFORMATION

PRODUCT	DESCRIPTION	MINIMUM RECOMMENDED GAIN	PACKAGE	PACKAGE DRAWING NUMBER	PACKAGE MARKING	ORDERING NUMBER ⁽¹⁾	TRANSPORT MEDIA
OPA703NA	Single, GBW = 1MHz	1	SOT23-5	331	A03	OPA703NA/250	Tape and Reel
OPA703UA	Single, GBW = 1MHz	1	SO-8	182	OPA703UA	OPA703NA/3K OPA703UA	Tape and Reel Rails
OPA703PA	Single, GBW = 1MHz	1	DIP-8	006	OPA703PA	OPA703UA/2K5 OPA703PA	Tape and Reel Rails
OPA2703EA	Dual, GBW = 1MHz	1	MSOP-8	337	B03	OPA2703EA/250	Tape and Reel
OPA2703UA	Dual, GBW = 1MHz	1	SO-8	182	OPA2703UA	OPA2703EA/2K5 OPA2703UA	Tape and Reel Rails
OPA2703PA	Dual, GBW = 1MHz	1	DIP-8	006	OPA2703PA	OPA2703UA/2K5 OPA2703PA	Tape and Reel Rails
OPA4703EA	Quad, GBW = 1MHz	1	TSSOP-14	357	OPA4703EA	OPA4703EA/250	Tape and Reel
OPA4703UA	Quad, GBW = 1MHz	1	SO-14	235	OPA4703UA	OPA4703EA/2K5 OPA4703UA	Tape and Reel Rails
OPA4703PA	Quad, GBW = 1MHz	1	SO-14	235	OPA4703UA	OPA4703UA/2K5	Tape and Reel
OPA704NA	Single, GBW = 5MHz	5	SOT23-5	331	A04	OPA704NA/250	Tape and Reel
OPA704UA	Single, GBW = 5MHz	5	SO-8	182	OPA704UA	OPA704NA/3K OPA704UA	Tape and Reel Rails
OPA704PA	Single, GBW = 5MHz	5	DIP-8	006	OPA704PA	OPA704UA/2K5 OPA704PA	Tape and Reel Rails
OPA2704EA	Dual, GBW = 5MHz	5	MSOP-8	337	B04	OPA2703EA/250	Tape and Reel
OPA2704UA	Dual, GBW = 5MHz	5	SO-8	182	OPA2704UA	OPA2703EA/2K5 OPA2704UA	Tape and Reel Rails
OPA2704PA	Dual, GBW = 5MHz	5	DIP-8	006	OPA2704PA	OPA2704UA/2K5 OPA2704PA	Tape and Reel Rails
OPA4704EA	Quad, GBW = 5MHz	5	TSSOP-14	357	OPA4704EA	OPA4704EA/250	Tape and Reel
OPA4704UA	Quad, GBW = 5MHz	5	SO-14	235	OPA4704UA	OPA4704EA/2K5 OPA4704UA	Tape and Reel Rails
OPA4704PA	Quad, GBW = 5MHz	5	SO-14	235	OPA4704UA	OPA4704UA/2K5	Tape and Reel

NOTE: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /3K indicates 3000 devices per reel). Ordering 3000 pieces of "OPA703NA/3K" will get a single 3000-piece Tape and Reel.

OPA703 ELECTRICAL CHARACTERISTICS: $V_S = 4V$ to $12V$

Boldface limits apply over the specified temperature range, $T_A = -40^{\circ}C$ to $+85^{\circ}C$

At $T_A = +25^{\circ}C$, $R_L = 20k\Omega$ connected to $V_S/2$ and $V_{OUT} = V_S/2$, unless otherwise noted.

PARAMETER	CONDITION	OPA703NA, UA, PA OPA2703EA, UA, PA OPA4703EA, UA			UNITS
		MIN	TYP	MAX	
OFFSET VOLTAGE Input Offset Voltage Drift vs Power Supply Over Temperature Channel Separation, dc $f = 1kHz$	V_{OS} dV_{OS}/dT PSRR $V_S = \pm 5V, V_{CM} = 0V$ $T_A = -40^{\circ}C$ to $+85^{\circ}C$ $V_S = \pm 2V$ to $\pm 6V, V_{CM} = 0V$ $V_S = \pm 2V$ to $\pm 6V, V_{CM} = 0V$ $R_L = 20k\Omega$		± 160 ± 4 20 1 98	± 750 100 200	μV $\mu V/^{\circ}C$ $\mu V/V$ $\mu V/V$ $\mu V/V$ dB
INPUT VOLTAGE RANGE Common-Mode Voltage Range Common-Mode Rejection Ratio over Temperature over Temperature	V_{CM} CMRR $V_S = \pm 5V, (V-) - 0.3V < V_{CM} < (V+) + 0.3V$ $V_S = \pm 5V, (V-) < V_{CM} < (V+)$ $V_S = \pm 5V, (V-) - 0.3V < V_{CM} < (V+) - 2V$ $V_S = \pm 5V, (V-) < V_{CM} < (V+) - 2V$	$(V-) - 0.3$ 70 68 80 74	90 96	$(V+) + 0.3$	V dB dB dB dB
INPUT BIAS CURRENT Input Bias Current Input Offset Current	I_B I_{OS} $V_S = \pm 5V, V_{CM} = 0V$ $V_S = \pm 5V, V_{CM} = 0V$		± 1 ± 0.5	± 10 ± 10	pA pA
INPUT IMPEDANCE Differential Common-Mode			$4 \cdot 10^9 \parallel 4$ $5 \cdot 10^{12} \parallel 4$		$\Omega \parallel pF$ $\Omega \parallel pF$
NOISE Input Voltage Noise, $f = 0.1Hz$ to $10Hz$ Input Voltage Noise Density, $f = 1kHz$ Current Noise Density, $f = 1kHz$	e_n i_n $V_S = \pm 5V, V_{CM} = 0V$ $V_S = \pm 5V, V_{CM} = 0V$ $V_S = \pm 5V, V_{CM} = 0V$		6 45 2.5		$\mu Vp-p$ $nV/\sqrt{Hz}$ $fA/\sqrt{Hz}$
OPEN-LOOP GAIN Open-Loop Voltage Gain over Temperature over Temperature	A_{OL} $R_L = 100k\Omega, (V-)+0.1V < V_O < (V+)-0.1V$ $R_L = 20k\Omega, (V-)+0.075V < V_O < (V+)-0.075V$ $R_L = 20k\Omega, (V-)+0.075V < V_O < (V+)-0.075V$ $R_L = 5k\Omega, (V-)+0.15V < V_O < (V+)-0.15V$ $R_L = 5k\Omega, (V-)+0.15V < V_O < (V+)-0.15V$	100 96 100 96	120 110 110		dB dB dB dB dB
OUTPUT Voltage Output Swing from Rail over Temperature over Temperature Output Current Short-Circuit Current Capacitive Load Drive	I_{OUT} I_{SC} C_{LOAD} $R_L = 100k\Omega, A_{OL} > 80dB$ $R_L = 20k\Omega, A_{OL} > 100dB$ $R_L = 20k\Omega, A_{OL} > 96dB$ $R_L = 5k\Omega, A_{OL} > 100dB$ $R_L = 5k\Omega, A_{OL} > 96dB$ $ V_S - V_{OUT} < 1V$		40 ± 10 ± 40	75 75 150 150	mV mV mV mV mV mA mA
See Typical Performance Curves					
FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Settling Time, 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise	GBW SR t_S THD+N $C_L = 100pF$ $G = +1$ $V_S = \pm 5V, G = +1$ $V_S = \pm 5V, 5V$ Step, $G = +1$ $V_S = \pm 5V, 5V$ Step, $G = +1$ $V_{IN} \cdot \text{Gain} = V_S$ $V_S = \pm 5V, V_O = 3Vp-p, G = +1, f = 1kHz$		1 0.6 15 20 3 0.02		MHz V/ μs μs μs μs %
POWER SUPPLY Specified Voltage Range, Single Supply Specified Voltage Range, Dual Supplies Operating Voltage Range Quiescent Current (per amplifier) over Temperature	V_S V_S I_Q $I_O = 0$	4 ± 2	3.6 to 12 160	12 ± 6 200 300	V V V μA μA
TEMPERATURE RANGE Specified Range Operating Range Storage Range Thermal Resistance SOT23-5 Surface-Mount MSOP-8 Surface-Mount TSSOP-14 Surface-Mount SO-8 Surface Mount SO-14 Surface Mount DIP-8	θ_{JA}	-40 -55 -65		85 125 150	$^{\circ}C$ $^{\circ}C$ $^{\circ}C$ $^{\circ}C/W$ $^{\circ}C/W$ $^{\circ}C/W$ $^{\circ}C/W$ $^{\circ}C/W$ $^{\circ}C/W$ $^{\circ}C/W$

OPA704 ELECTRICAL CHARACTERISTICS: $V_S = 4V$ to $12V$

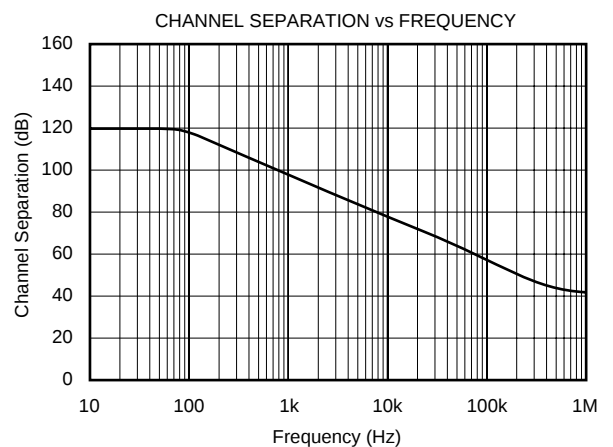
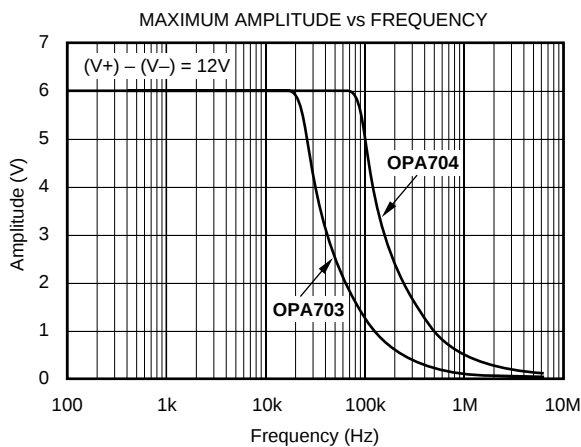
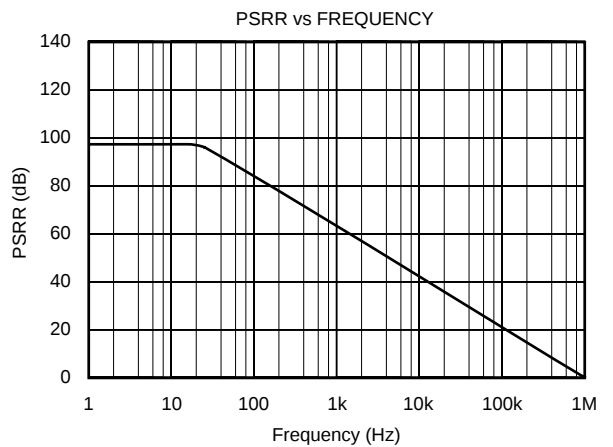
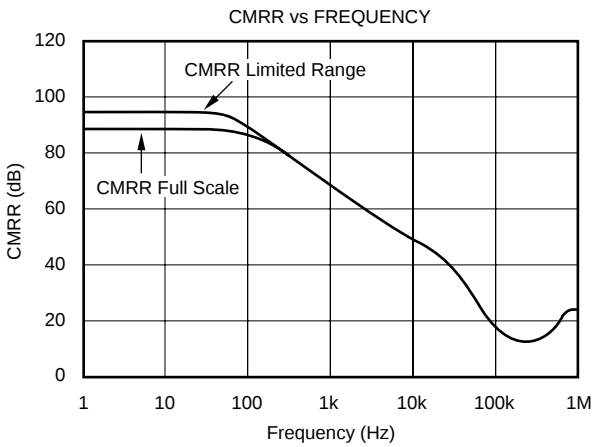
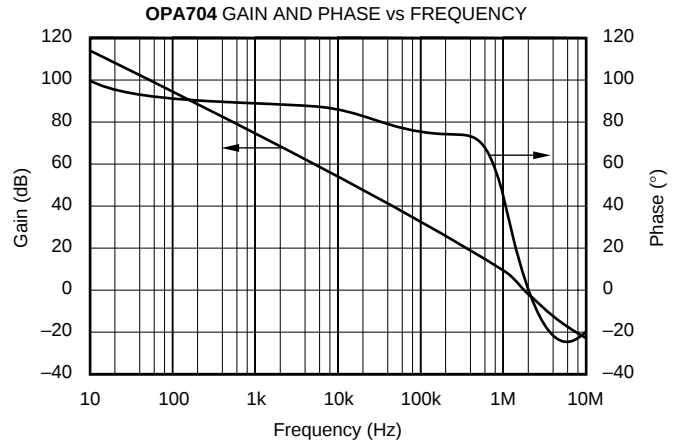
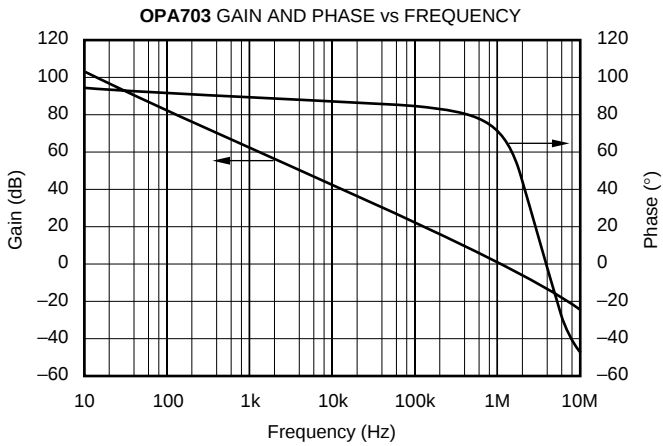
Boldface limits apply over the specified temperature range, $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$

At $T_A = +25^{\circ}\text{C}$, $R_L = 20\text{k}\Omega$ connected to $V_S/2$ and $V_{OUT} = V_S/2$, unless otherwise noted.

PARAMETER	CONDITION	OPA704NA, UA, PA OPA2704EA, UA, PA OPA4704EA, UA			UNITS
		MIN	TYP	MAX	
OFFSET VOLTAGE Input Offset Voltage Drift vs Power Supply Over Temperature Channel Separation, dc $f = 1\text{kHz}$	V_{OS} dV_{OS}/dT PSRR $V_S = \pm 5V$, $V_{CM} = 0V$ $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ $V_S = \pm 2V$ to $\pm 6V$, $V_{CM} = 0V$ $V_S = \pm 2V$ to $\pm 6V$, $V_{CM} = 0V$ $R_L = 20\text{k}\Omega$		± 160 ± 4 20 1 98	± 750 100 200	μV $\mu\text{V}/^{\circ}\text{C}$ $\mu\text{V}/\text{V}$ $\mu\text{V}/\text{V}$ $\mu\text{V}/\text{V}$ dB
INPUT VOLTAGE RANGE Common-Mode Voltage Range Common-Mode Rejection Ratio over Temperature over Temperature	V_{CM} CMRR $V_S = \pm 5V$, $(V-) - 0.3V < V_{CM} < (V+) + 0.3V$ $V_S = \pm 5V$, $(V-) < V_{CM} < (V+)$ $V_S = \pm 5V$, $(V-) - 0.3V < V_{CM} < (V+) - 2V$ $V_S = \pm 5V$, $(V-) < V_{CM} < (V+) - 2V$	$(V-) - 0.3$ 70 68 80 74	90 96	$(V+) + 0.3$	V dB dB dB dB
INPUT BIAS CURRENT Input Bias Current Input Offset Current	I_B I_{OS} $V_S = \pm 5V$, $V_{CM} = 0V$ $V_S = \pm 5V$, $V_{CM} = 0V$		± 1 ± 0.5	± 10 ± 10	pA pA
INPUT IMPEDANCE Differential Common-Mode			$4 \cdot 10^9 \parallel 4$ $5 \cdot 10^{12} \parallel 4$		$\Omega \parallel \text{pF}$ $\Omega \parallel \text{pF}$
NOISE Input Voltage Noise, $f = 0.1\text{Hz}$ to 10Hz Input Voltage Noise Density, $f = 1\text{kHz}$ Current Noise Density, $f = 1\text{kHz}$	e_n i_n $V_S = \pm 5V$, $V_{CM} = 0V$ $V_S = \pm 5V$, $V_{CM} = 0V$ $V_S = \pm 5V$, $V_{CM} = 0V$		6 45 2.5		$\mu\text{Vp-p}$ $\text{nV}/\sqrt{\text{Hz}}$ $\text{fA}/\sqrt{\text{Hz}}$
OPEN-LOOP GAIN Open-Loop Voltage Gain over Temperature over Temperature	A_{OL} $R_L = 100\text{k}\Omega$, $(V-) + 0.1V < V_O < (V+) - 0.1V$ $R_L = 20\text{k}\Omega$, $(V-) + 0.075V < V_O < (V+) - 0.075V$ $R_L = 20\text{k}\Omega$, $(V-) + 0.075V < V_O < (V+) - 0.075V$ $R_L = 5\text{k}\Omega$, $(V-) + 0.15V < V_O < (V+) - 0.15V$ $R_L = 5\text{k}\Omega$, $(V-) + 0.15V < V_O < (V+) - 0.15V$	100 96 100 96	120 110 110		dB dB dB dB dB
OUTPUT Voltage Output Swing from Rail over Temperature over Temperature Output Current Short-Circuit Current Capacitive Load Drive	I_{OUT} I_{SC} C_{LOAD} $R_L = 100\text{k}\Omega$, $A_{OL} > 80\text{dB}$ $R_L = 20\text{k}\Omega$, $A_{OL} > 100\text{dB}$ $R_L = 20\text{k}\Omega$, $A_{OL} > 96\text{dB}$ $R_L = 5\text{k}\Omega$, $A_{OL} > 100\text{dB}$ $R_L = 5\text{k}\Omega$, $A_{OL} > 96\text{dB}$ $ V_S - V_{OUT} < 1V$		40 ± 10 ± 40	75 75 150 150	mV mV mV mV mA mA
FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Settling Time, 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise	GBW SR t_S t_S THD+N $C_L = 100\text{pF}$ $G = +5$ $V_S = \pm 5V$, $G = +5$ $V_S = \pm 5V$, 5V Step, $G = +5$ $V_S = \pm 5V$, 5V Step, $G = +5$ $V_{IN} \cdot \text{Gain} = V_S$ $V_S = \pm 5V$, $V_O = 3V_{p-p}$, $G = +5$, $f = 1\text{kHz}$		3 3 18 21 0.6 0.025		MHz V/ μs μs μs μs %
POWER SUPPLY Specified Voltage Range, Single Supply Specified Voltage Range, Dual Supplies Operating Voltage Range Quiescent Current (per amplifier) over Temperature	V_S V_S I_Q $I_Q = 0$	4 ± 2	3.6 to 12 160	12 ± 6 200 300	V V V μA μA
TEMPERATURE RANGE Specified Range Operating Range Storage Range Thermal Resistance SOT23-5 Surface-Mount MSOP-8 Surface-Mount TSSOP-14 Surface-Mount SO-8 Surface Mount SO-14 Surface Mount DIP-8	θ_{JA}	-40 -55 -65		85 125 150	$^{\circ}\text{C}$ $^{\circ}\text{C}$ $^{\circ}\text{C}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$ $^{\circ}\text{C}/\text{W}$

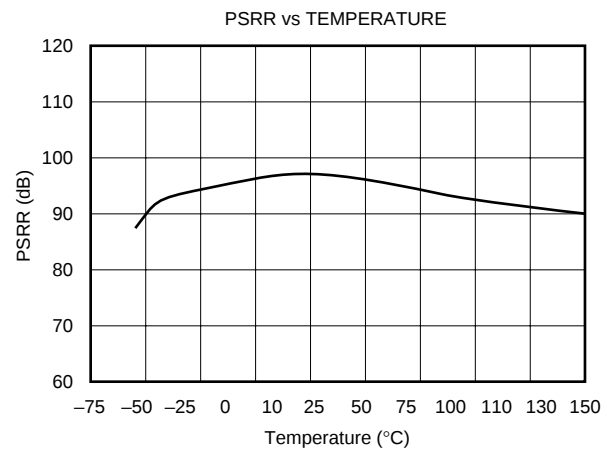
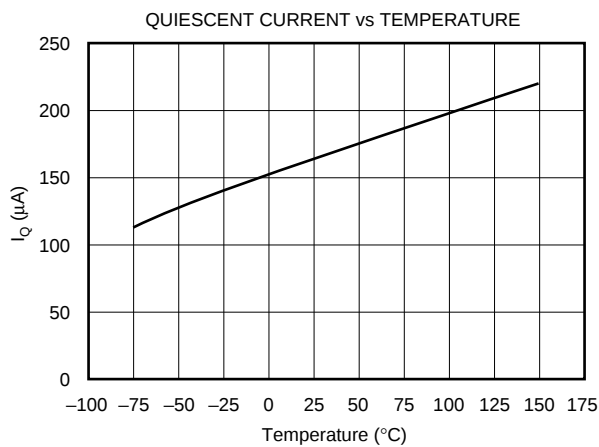
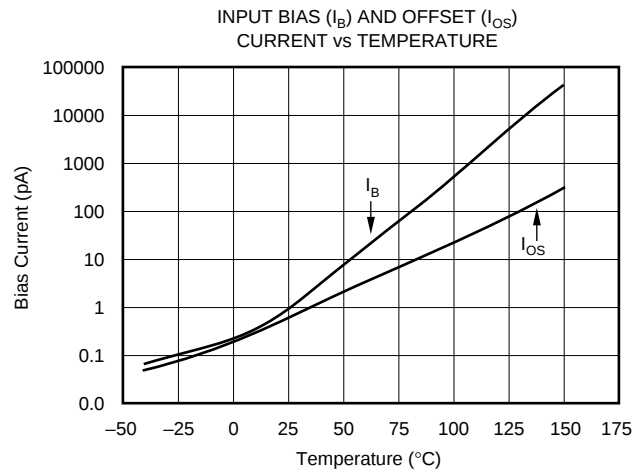
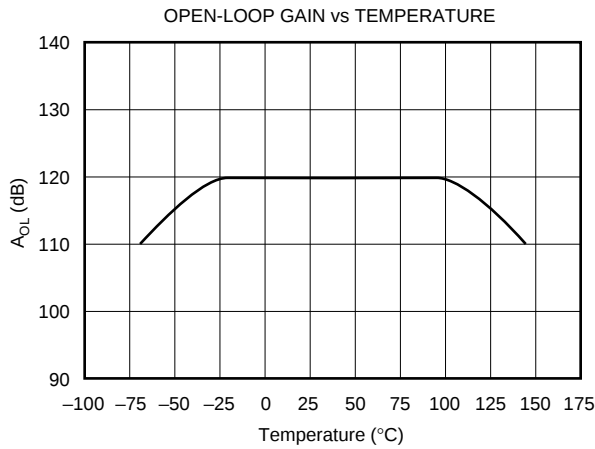
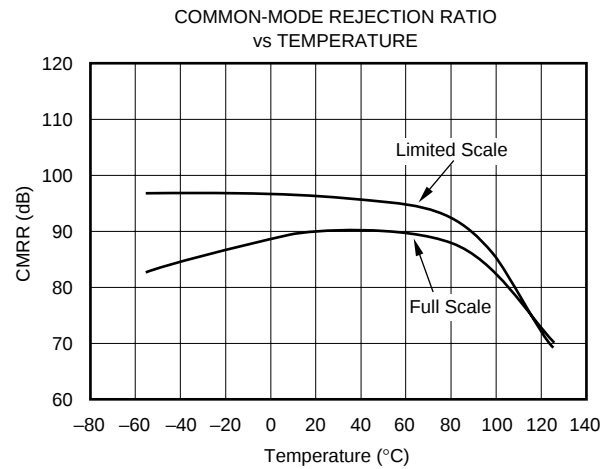
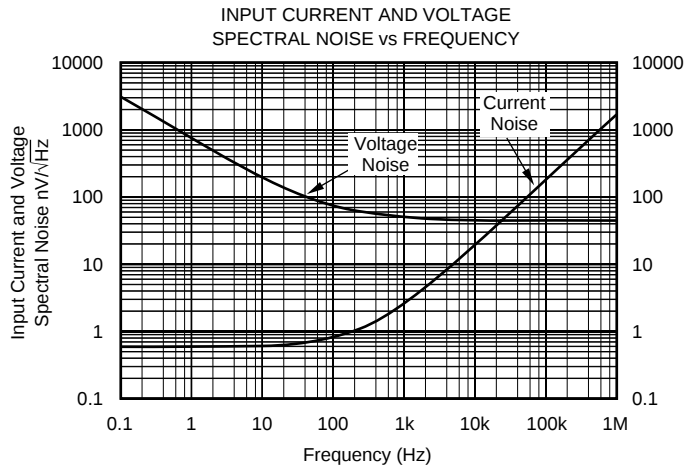
TYPICAL CHARACTERISTICS

At $T_A = +25^\circ\text{C}$, $V_S = \pm 5\text{V}$, and $R_L = 20\text{k}\Omega$, unless otherwise noted.



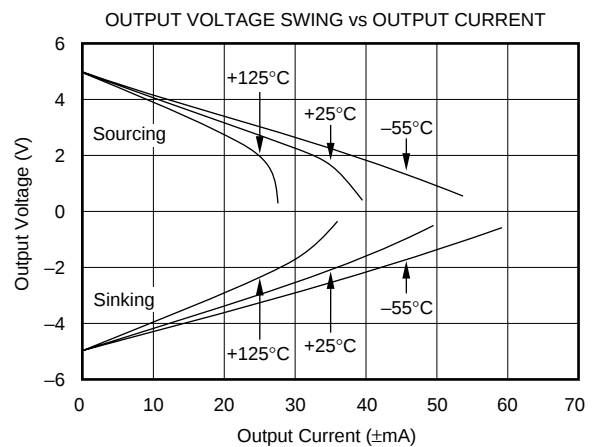
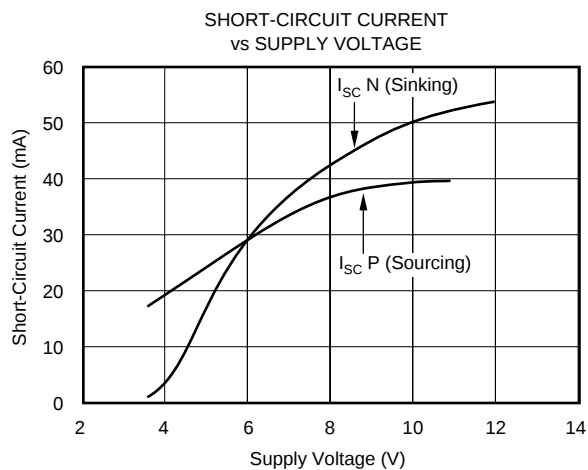
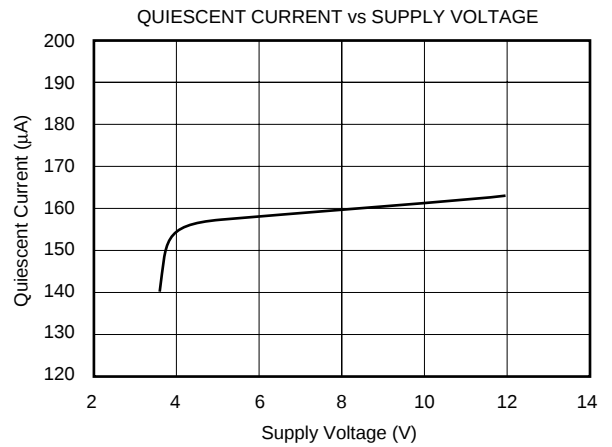
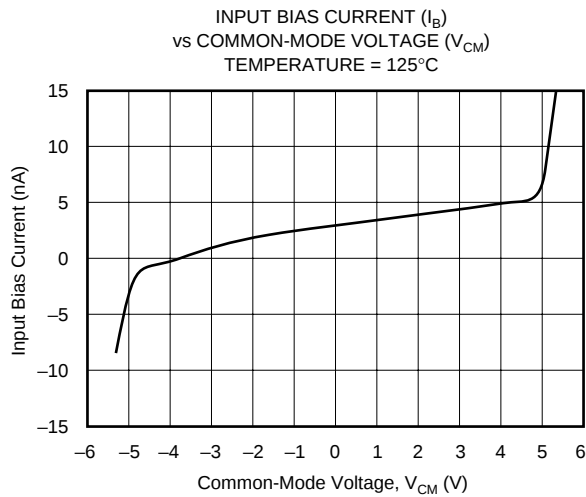
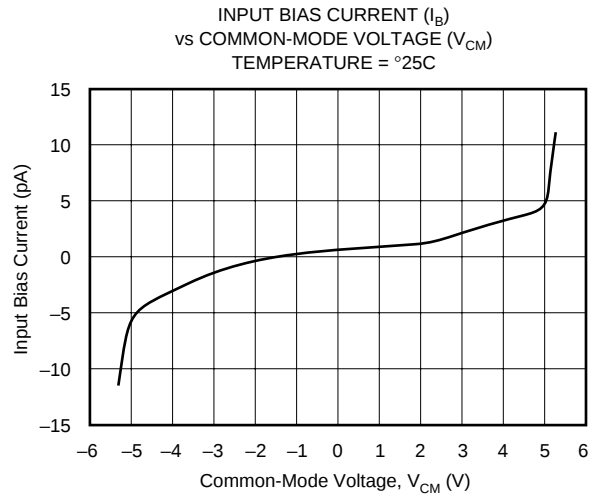
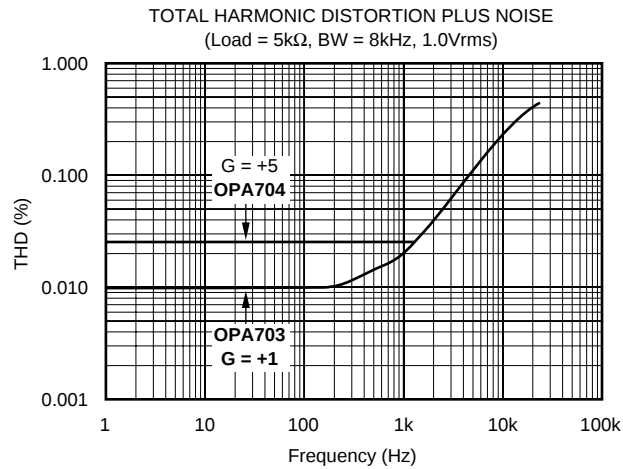
TYPICAL CHARACTERISTICS (Cont.)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 5\text{V}$, and $R_L = 20\text{k}\Omega$, unless otherwise noted.



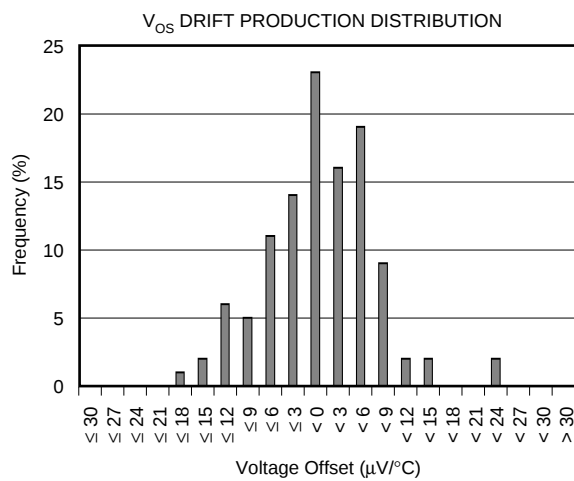
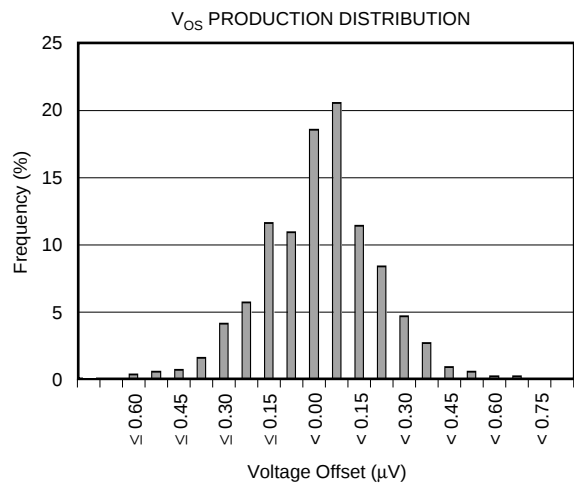
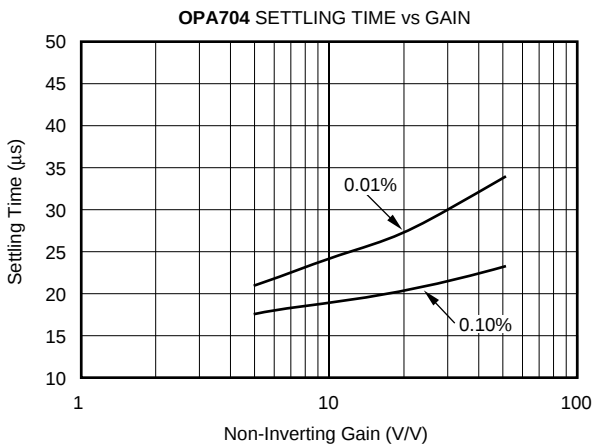
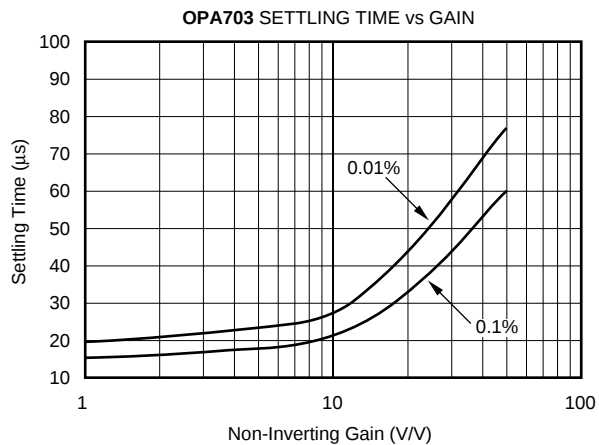
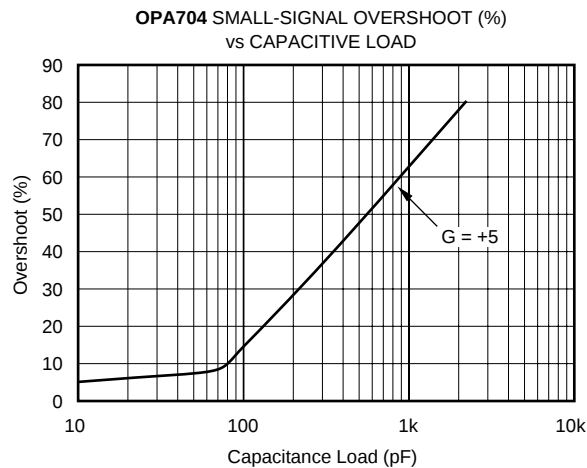
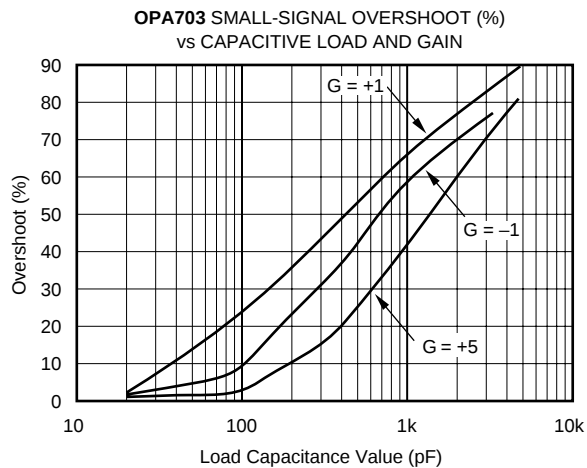
TYPICAL CHARACTERISTICS (Cont.)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 5\text{V}$, and $R_L = 20\text{k}\Omega$, unless otherwise noted.



TYPICAL CHARACTERISTICS (Cont.)

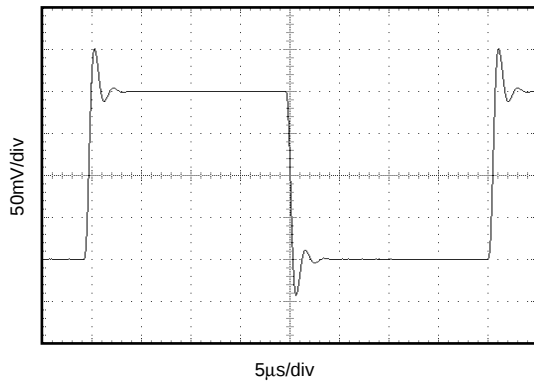
At $T_A = +25^\circ\text{C}$, $V_S = \pm 5\text{V}$, and $R_L = 20\text{k}\Omega$, unless otherwise noted.



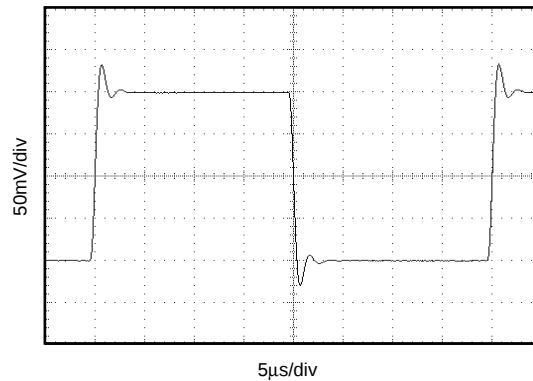
TYPICAL CHARACTERISTICS (Cont.)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 5\text{V}$, and $R_L = 20\text{k}\Omega$, unless otherwise noted.

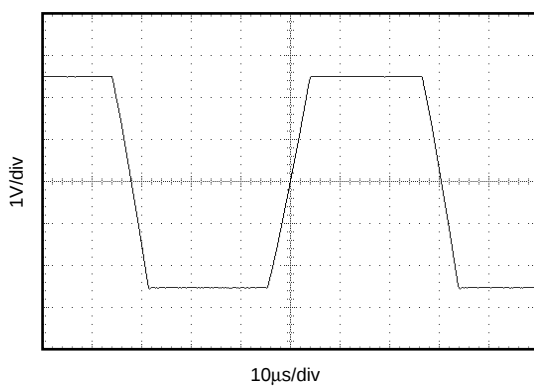
OPA703 SMALL SIGNAL STEP RESPONSE
($G = +1\text{V/V}$, $R_L = 20\text{k}\Omega$, $C_L = 100\text{pF}$)



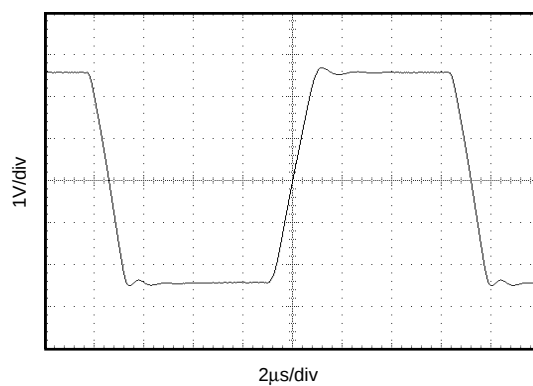
OPA704 SMALL SIGNAL STEP RESPONSE
($G = +5\text{V/V}$, $C_F = 3\text{pF}$, $R_F = 100\text{k}\Omega$,
 $C_L = 100\text{pF}$, $R_L = 20\text{k}\Omega$)



OPA703 LARGE SIGNAL STEP RESPONSE
($G = +1\text{V/V}$, $R_L = 20\text{k}\Omega$, $C_L = 100\text{pF}$)



OPA704 LARGE SIGNAL STEP RESPONSE
($G = +5\text{V/V}$, $R_L = 20\text{k}\Omega$, $C_F = 3\text{pF}$, $C_L = 100\text{pF}$)



APPLICATIONS INFORMATION

OPA703 and OPA704 series op amps can operate on 160 μ A quiescent current from a single (or split) supply in the range of 4V to 12V (± 2 V to ± 6 V), making them highly versatile and easy to use. The OPA703 is unity-gain stable and offers 1MHz bandwidth and 0.6V/ μ s slew rate. The OPA704 is optimized for gains of 5 or greater with a 3MHz bandwidth and 3V/ μ s slew rate.

Rail-to-rail input and output swing helps maintain dynamic range, especially in low supply applications. Figure 1 shows the input and output waveforms for the OPA703 in unity-gain configuration. Operation is from a ± 5 V supply with a 100k Ω load connected to $V_S/2$. The input is a 10Vp-p sinusoid. Output voltage is approximately 10Vp-p.

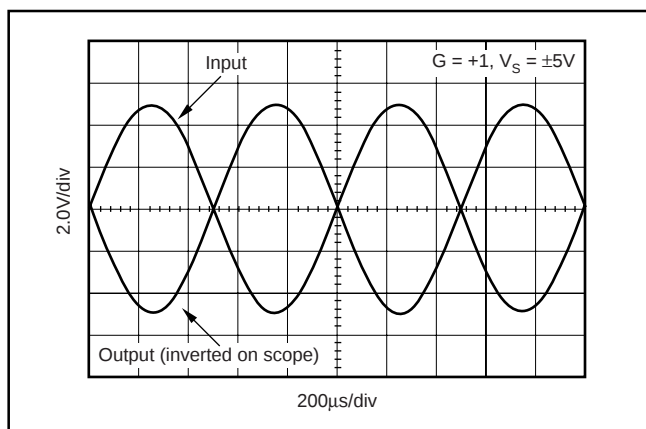


FIGURE 1. Rail-to-Rail Input and Output.

Power-supply pins should be bypassed with 1000pF ceramic capacitors in parallel with 1 μ F tantalum capacitors.

OPERATING VOLTAGE

OPA703 and OPA704 series op amps are fully specified and guaranteed from +4V to +12V over a temperature range of -40°C to $+85^{\circ}\text{C}$. Parameters that vary significantly with operating voltages or temperature are shown in the Typical Performance Curves.

RAIL-TO-RAIL INPUT

The input common-mode voltage range of the OPA703 series extends 300mV beyond the supply rails at room temperature. This is achieved with a complementary input stage—an N-channel input differential pair in parallel with a P-channel differential pair, as shown in Figure 2. The N-channel pair is active for input voltages close to the positive rail, typically $(V_+) - 2.0$ V to 300mV above the positive supply, while the P-channel pair is on for inputs from 300mV below the negative supply to approximately $(V_+) - 1.5$ V. There is a small transition region, typically $(V_+) - 2.0$ V to $(V_+) - 1.5$ V, in which both pairs are on. This 500mV transition region can vary ± 100 mV with process variation. Thus, the transition region (both stages on) can range from $(V_+) - 2.1$ V to $(V_+) - 1.4$ V on the low end, up to $(V_+) - 1.9$ V to $(V_+) - 1.6$ V on the high end. Within the 500mV transition region PSRR, CMRR, offset voltage, and offset drift, and THD may vary compared to operation outside this region.

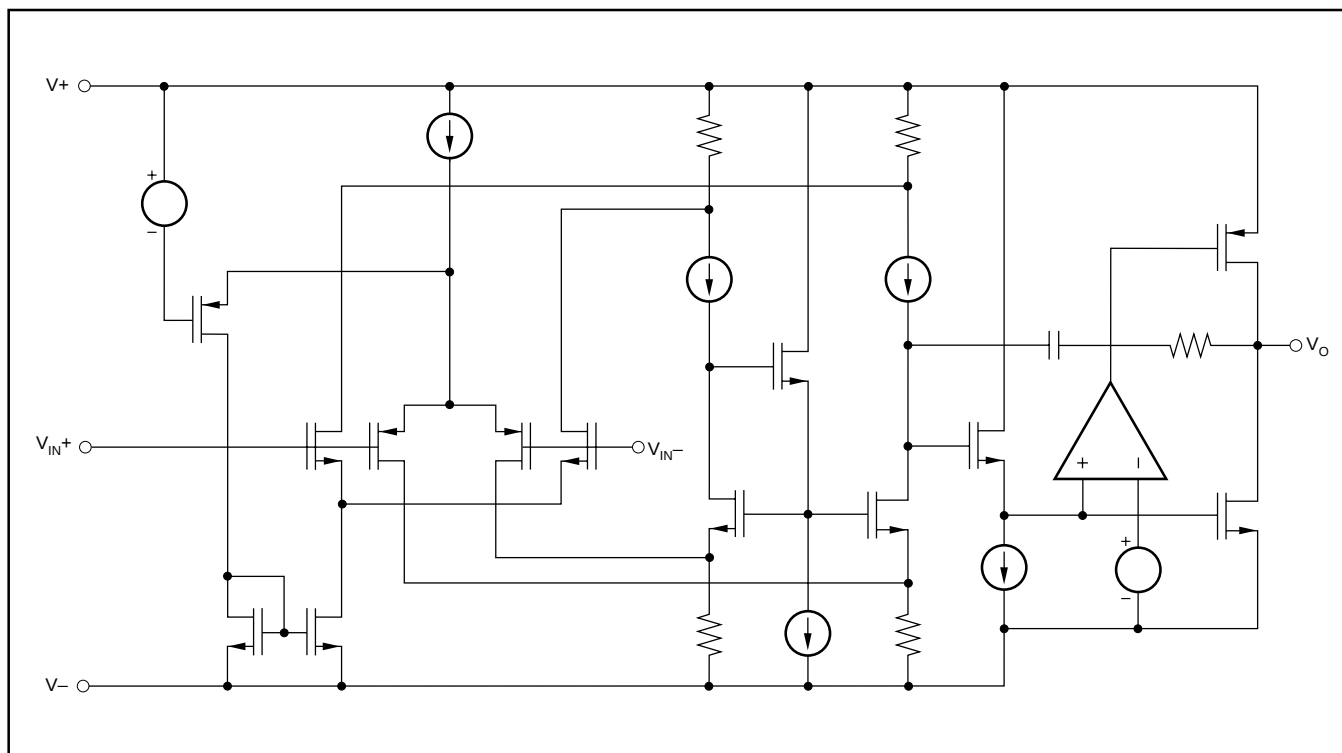


FIGURE 2. Simplified Schematic.

INPUT VOLTAGE

Device inputs are protected by ESD diodes that will conduct if the input voltages exceed the power supplies by more than approximately 300mV. Momentary voltages greater than 300mV beyond the power supply can be tolerated if the current is limited to 10mA. This is easily accomplished with an input resistor, as shown in Figure 3. Many input signals are inherently current-limited to less than 10mA; therefore, a limiting resistor is not always required. The OPA703 features no phase inversion when the inputs extend beyond supplies if the input current is limited, as seen in Figure 4.

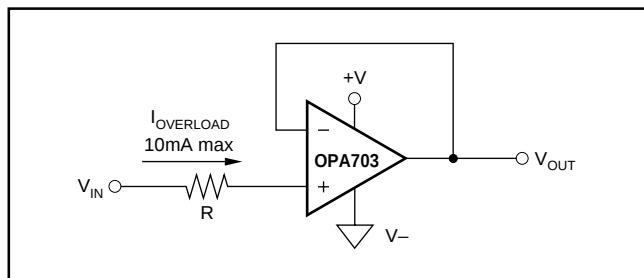


FIGURE 3. Input Current Protection for Voltages Exceeding the Supply Voltage.

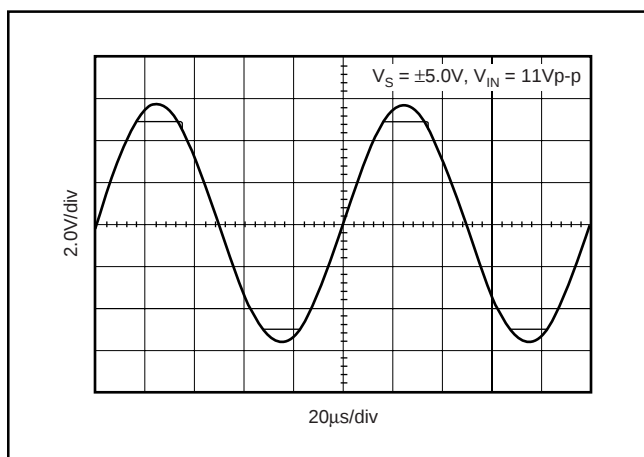


FIGURE 4. OPA703—No Phase Inversion with Inputs Greater than the Power-Supply Voltage.

RAIL-TO-RAIL OUTPUT

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. This output stage is capable of driving 1kΩ loads connected to any point between V+ and ground. For light resistive loads (> 100kΩ), the output voltage can swing to 40mV from the supply rail. With moderate resistive loads (20kΩ), the output can swing to within 75mV from the supply rails while maintaining high open-loop gain (see the typical performance curve “Output Voltage Swing vs Output Current”).

CAPACITIVE LOAD AND STABILITY

The OPA703 and OPA704 series op amps can drive up to 1000pF pure capacitive load. Increasing the gain enhances the amplifier’s ability to drive greater capacitive loads (see the typical performance curve “Small Signal Overshoot vs Capacitive Load”).

One method of improving capacitive load drive in the unity-gain configuration is to insert a 10Ω to 20Ω resistor inside the feedback loop, as shown in Figure 5. This reduces ringing with large capacitive loads while maintaining DC accuracy.

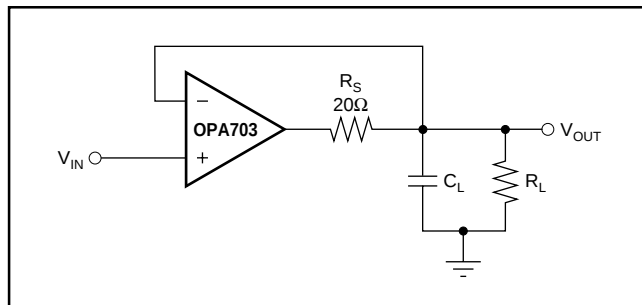


FIGURE 5. Series Resistor in Unity-Gain Buffer Configuration Improves Capacitive Load Drive.

APPLICATION CIRCUITS

Figure 6 shows a G = 5 non-inverting amplifier implemented with the OPA703 and OPA704 op amps. It demonstrates the increased speed characteristics (bandwidth, slew rate and settling time) that can be achieved with the OPA704 family when used in gains of five or greater. Some optimization of feedback capacitor value may be required to achieve best dynamic response. Circuits with closed-loop gains of less than five should use the OPA703 family for good stability and capacitive load drive. The OPA703 can be used in gains greater than five, but will not provide the increased speed benefits of the OPA704 family.

The OPA703 series op amps are optimized for driving medium-speed sampling data converters. The OPA703 op amps buffer the converter’s input capacitance and resulting charge injection while providing signal gain.

Figure 7 shows the OPA2703 in a dual-supply buffered reference configuration for the DAC7644. The DAC7644 is a 16-bit, low-power, quad-voltage output converter. Small size makes the combination ideal for automatic test equipment, data acquisition systems, and other low-power space-limited applications.

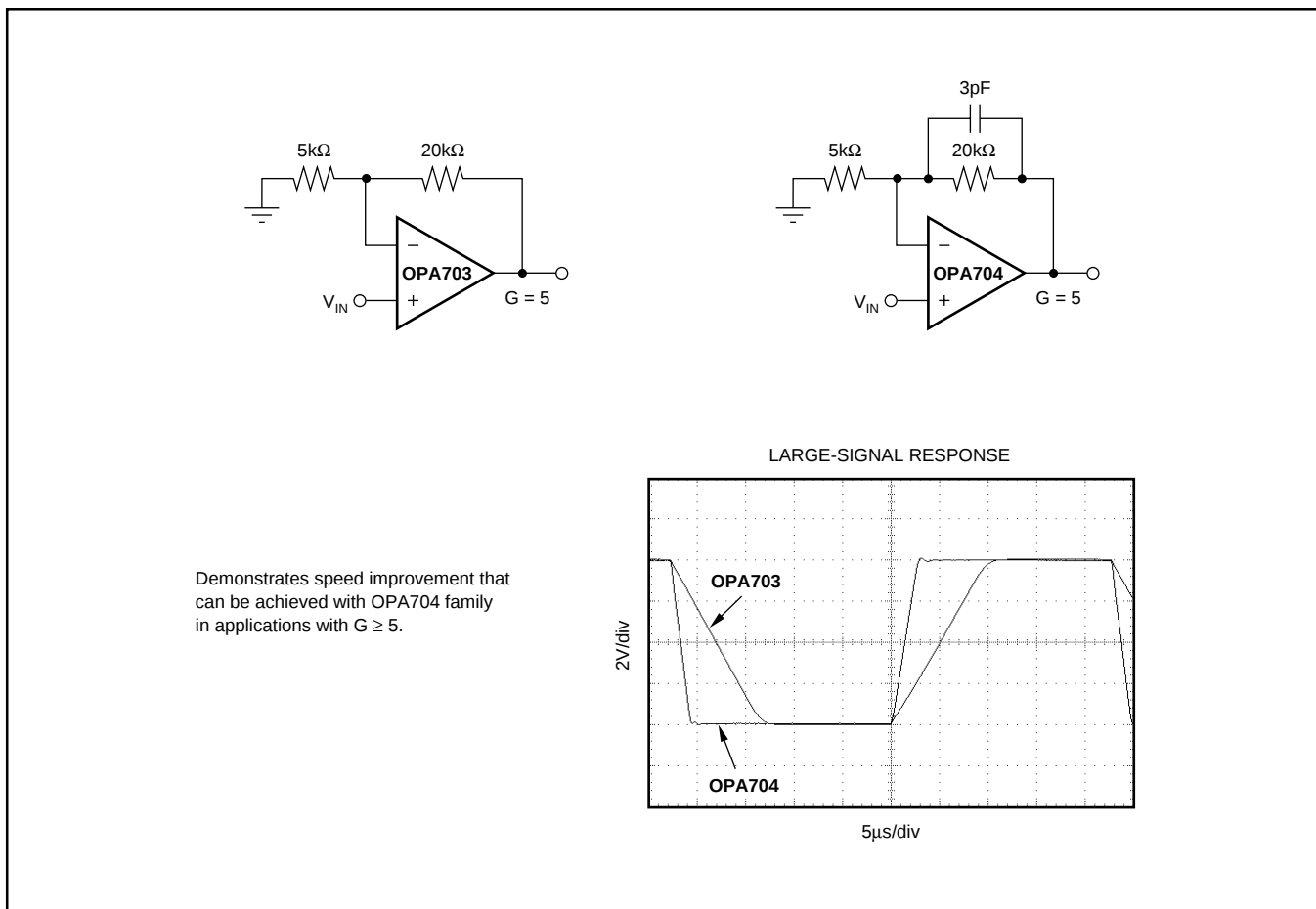


FIGURE 6. OPA704 Provides higher Speed in $G \geq 5$.

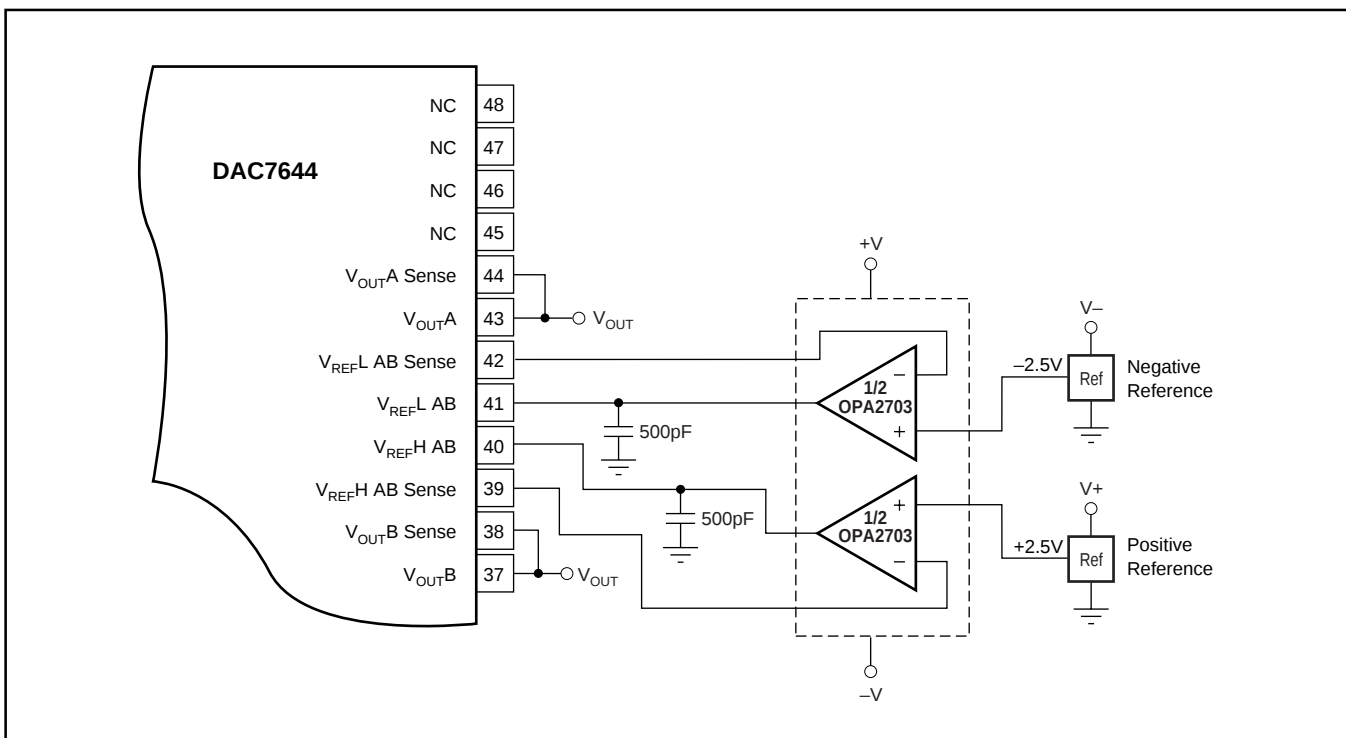


FIGURE 7. OPA703 as Dual Supply Configuration-Buffered References for the DAC7644.

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