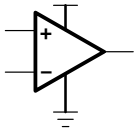
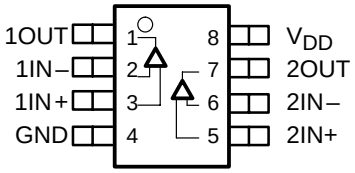
- **High Output Drive . . . >300 mA**
- **Rail-To-Rail Output**
- **Unity-Gain Bandwidth . . . 2.7 MHz**
- **Slew Rate . . . 1.5 V/μs**
- **Supply Current . . . 700 μA/Per Channel**
- **Supply Voltage Range . . . 2.5 V to 6 V**
- **Specified Temperature Range:**
  - **T_A = 0°C to 70°C . . . Commercial Grade**
  - **T_A = –40°C to 125°C . . . Industrial Grade**
- **Universal OpAmp EVM**

Operational Amplifier



TLV4112
D, DGN, OR P PACKAGE
(TOP VIEW)



description

The TLV411x single supply operational amplifiers provide output currents in excess of 300 mA at 5 V. This enables standard pin-out amplifiers to be used as high current buffers or in coil driver applications. The TLV4110 and TLV4113 come with a shutdown feature.

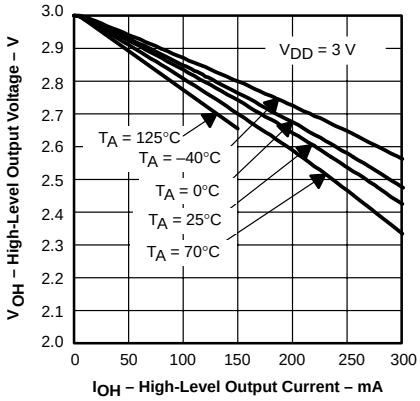
The TLV411x is available in the ultra small MSOP PowerPAD™ package, which offers the exceptional thermal impedance required for amplifiers delivering high current levels.

All TLV411x devices are offered in PDIP, SOIC (single and dual) and MSOP PowerPAD (dual).

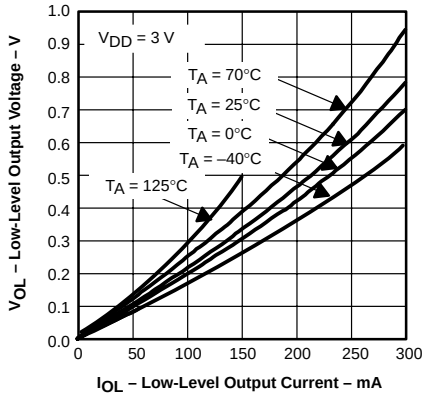
FAMILY PACKAGE TABLE

DEVICE	NUMBER OF CHANNELS	PACKAGE TYPES			SHUTDOWN	UNIVERSAL EVM BOARD
		MSOP	PDIP	SOIC		
TLV4110	1	8	8	8	Yes	Refer to the EVM Selection Guide (Lit# SLOU060)
TLV4111	1	8	8	8	—	
TLV4112	2	8	8	8	—	
TLV4113	2	10	14	14	Yes	

HIGH-LEVEL OUTPUT VOLTAGE vs HIGH-LEVEL OUTPUT CURRENT



LOW-LEVEL OUTPUT VOLTAGE vs LOW-LEVEL OUTPUT CURRENT



TLV4110 AND TLV4111 AVAILABLE OPTIONS

T_A	PACKAGED DEVICES			
	SMALL OUTLINE (D)†‡	MSOP		PLASTIC DIP (P)
		SMALL OUTLINE (DGN)†	SYMBOL	
0°C to 70°C	TLV4110CD	TLV4110CDGN	xxTIAHL	TLV4110CP
	TLV4111CD	TLV4111CDGN	xxTIAHN	TLV4111CP
–40°C to 125°C	TLV4110ID	TLV4110IDGN	xxTIAHM	TLV4110IP
	TLV4111ID	TLV4111IDGN	xxTIAHO	TLV4111IP

† This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLV4110CDR).
‡ In the SOIC package, the maximum RMS output power is thermally limited to 350 mW; 700 mW peaks can be driven, as long as the RMS value is less than 350 mW.

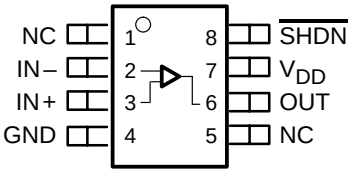
TLV4112 AND TLV4113 AVAILABLE OPTIONS

T_A	PACKAGED DEVICES					
	SMALL OUTLINE (D)†‡	MSOP				PLASTIC DIP (P)
		SMALL OUTLINE (DGN)†	SYMBOL	SMALL OUTLINE (DGQ)†	SYMBOL	
0°C to 70°C	TLV4112CD	TLV4112DGN	xxTIAHP	—	—	TLV4112CP
	TLV4113CD	—	—	TLV4113CDGQ	xxTIAHR	TLV4113CN
–40°C to 125°C	TLV4112ID	TLV4112IDGN	xxTIAHQ	—	—	TLV4112IP
	TLV4113ID	—	—	TLV4113IDGQ	xxTIAHS	TLV4113IN

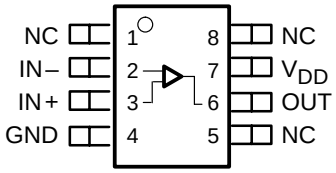
† This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLV4112CDR).
‡ In the SOIC package, the maximum RMS output power is thermally limited to 350 mW; 700 mW peaks can be driven, as long as the RMS value is less than 350 mW.

TLV411x PACKAGE PINOUTS

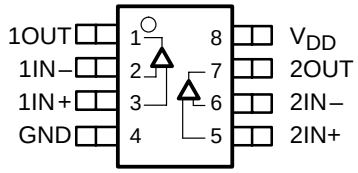
TLV4110
D, DGN OR P PACKAGE
(TOP VIEW)



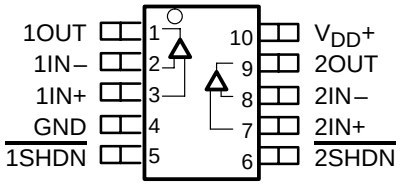
TLV4111
D, DGN OR P PACKAGE
(TOP VIEW)



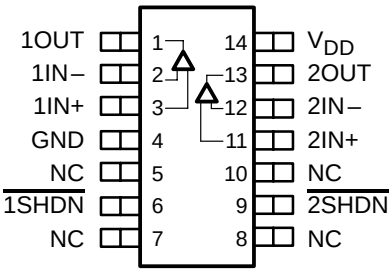
TLV4112
D, DGN, OR P PACKAGE
(TOP VIEW)



TLV4113
DGQ PACKAGE
(TOP VIEW)



TLV4113
D OR N PACKAGE
(TOP VIEW)



NC – No internal connection

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V_{DD} (see Note 1) . . . . . 7 V
Differential input voltage, V_{ID} . . . . . $\pm V_{DD}$
Input voltage range, V_I . . . . . $\pm V_{DD}$
Output current, I_O (see Note 2) . . . . . 800 mA
Continuous /RMS output current, I_O (each output of amplifier): $T_J \le 105°C$ . . . . . 350 mA
$T_J \le 150°C$ . . . . . 110 mA
Peak output current, I_O (each output of amplifier: $T_J \le 105°C$ . . . . . 500 mA
$T_J \le 150°C$ . . . . . 155 mA
Continuous total power dissipation . . . . . See Dissipation Rating Table
Operating free-air temperature range, T_A: C suffix . . . . . 0°C to 70°C
I suffix . . . . . –40°C to 125°C
Maximum junction temperature, T_J . . . . . 150°C
Storage temperature range, T_{stg} . . . . . –65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds . . . . . 260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. All voltage values, except differential voltages, are with respect to GND.
2. To prevent permanent damage the die temperature must not exceed the maximum junction temperature.

DISSIPATION RATING TABLE

PACKAGE	θ_{JC} (°C/W)	θ_{JA} (°C/W)	$T_A \le 25°C$ POWER RATING	$T_A = 125°C$ POWER RATING
D (8)	38.3	176	710 mW	142 mW
D (14)	26.9	122.3	1022 mW	204.4 mW
DGN (8)‡	4.7	52.7	2.37 W	474.4 mW
DGQ (10)‡	4.7	52.3	2.39 W	478 mW
P (8)	41	104	1200 mW	240.4 mW
N (14)	32	78	1600 mW	320.5 mW

‡ See The Texas Instruments document, *PowerPAD Thermally Enhanced Package Application Report* (literature number SLMA002), for more information on the PowerPAD package. The thermal data was measured on a PCB layout based on the information in the section entitled *Texas Instruments Recommended Board for PowerPAD* on page 33 of the before mentioned document.

recommended operating conditions

			MIN	MAX	UNIT
Supply voltage, V_{DD}			2.5	6	V
Common-mode input voltage range, V_{ICR}			0	V_{DD}–1.5	V
Operating free-air temperature, T_A	C-suffix		0	70	°C
	I-suffix		–40	125	
Shutdown turnon/off voltage level§	V(on)	V_{DD} = 3 V	2.1		V
		V_{DD} = 5 V	3.8		
	V(off)	V_{DD} = 3 V		0.9	
		V_{DD} = 5 V		1.65	

§ Relative to GND

electrical characteristics at recommend operating conditions, V_{DD} = 3 V and 5 V (unless otherwise noted)

dc performance

PARAMETER		TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNITS
V_{IO}	Input offset voltage	$V_{IC} = V_{DD}/2$, R_L = 100 Ω,	$V_O = V_{DD}/2$, R_S = 50 Ω	25°C		175	3500	µV
				Full range			4000	
αVIO	Offset voltage draft			25°C		3		µV/°C
CMRR	Common-mode rejection ratio	V_{DD} = 3 V, R_S = 50 Ω	V_{IC} = 0 to 2 V,	25°C		63		dB
		V_{DD} = 5 V, R_S = 50 Ω	V_{IC} = 0 to 4 V,	25°C		68		
A_{VD}	Large-signal differential voltage amplification	V_{DD} = 3 V, $V_{O(PP)}$=0 to 1V	R_L=100 Ω	25°C	78	84		dB
				Full range	67			
			R_L=10 kΩ	25°C	85	100		
				Full range	75			
		V_{DD} = 5 V, $V_{O(PP)}$=0 to 3V	R_L=100 Ω	25°C	88	94		
				Full range	75			
			R_L=10 kΩ	25°C	90	110		
				Full range	85			

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

input characteristics

PARAMETER		TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNITS
I_{IO}	Input offset current	$V_{IC} = V_{DD}/2$		25°C		0.3	25	pA
			TLV411xC	Full range			50	
			TLV411xI				250	
I_{IB}	Input bias current	$V_O = V_{DD}/2$, R_S = 50 Ω		25°C		0.3	50	
			TLV411xC	Full range			100	
			TLV411xI				500	
$r_{i(d)}$	Differential input resistance			25°C		1000		GΩ
C_{IC}	Common-mode input capacitance	f = 100 Hz		25°C		5		pF

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

electrical characteristics at specified free-air temperature, V_{DD} = 3 V and 5 V (unless otherwise noted) (continued)

output characteristics

PARAMETER		TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNITS
V_{OH}	High-level output voltage	V_{DD} = 3 V, V_{IC} = V_{DD}/2	I_{OH} = –10 mA	25°C	2.7	2.97		V
				Full range	2.7			
			I_{OH} =–100 mA	25°C	2.6	2.73		
				Full range	2.6			
		V_{DD} = 5 V, V_{IC} = V_{DD}/2	I_{OH} = –10 mA	25°C	4.7	4.96		V
				Full range	4.7			
			I_{OH} = –100 mA	25°C	4.6	4.76		
				Full range	4.6			
			I_{OH} = –200 mA	25°C	4.45	4.6		
				–40°C to 85°C	4.35			
V_{OL}	Low-level output voltage	V_{DD} = 3 V and 5 V, V_{IC} = V_{DD}/2	I_{OL} = 10 mA	25°C		0.03	0.1	V
				Full range			0.1	
			I_{OL} = 100 mA	25°C		0.33	0.4	
				Full range			0.55	
		V_{DD} = 5 V, V_{IC} = V_{DD}/2	I_{OL} = 200 mA	25°C		0.38	0.6	
				–40°C to 85°C			0.7	
I_O	Output current‡	Measured at 0.5 V from rail	V_{DD} = 3 V	25°C		±220		mA
			V_{DD} = 5 V			±320		
I_{OS}	Short-circuit output current‡	Sourcing		25°C		800		mA
		Sinking				800		

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.
‡ When driving output currents in excess of 200 mA, the MSOP PowerPAD package is required for thermal dissipation.

power supply

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNITS
I_{DD}	Supply current (per channel)	V_O = V_{DD}/2	25°C		700	1000	μA
			Full range			1500	
PSRR	Power supply rejection ratio (ΔV_{DD} / ΔV_{IO})	V_{DD} =2.7 to 3.3 V, No load, V_{IC} = V_{DD}/2 V	25°C	70	82		dB
			Full range	65			
		V_{DD} =4.5 to 5.5 V, No load, V_{IC} = V_{DD}/2 V	25°C	70	79		
			Full range	65			

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

electrical characteristics at specified free-air temperature, V_{DD} = 3 V and 5 V (unless otherwise noted) (continued)

dynamic performance

PARAMETER		TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNITS
GBWP	Gain bandwidth product	R_L=100 Ω	C_L=10 pF	25°C		2.7		MHz
SR	Slew rate at unity gain	$V_{O(pp)}$ = 2 V, R_L = 100 Ω, C_L = 10 pF	V_{DD} = 3 V	25°C	0.8	1.57		V/µs
				Full range	0.55			
			V_{DD} = 5 V	25°C	1	1.57		
				Full range	0.7			
ϕM	Phase margin	R_L = 100 Ω,	C_L = 10 pF	25°C		66		
	Gain margin					16		dB
t_s	Settling time	$V_{(STEP)pp}$ = 1 V, A_V = –1, C_L = 10 pF, R_L = 100 Ω	0.1%	25°C		0.7		µs
			0.01%			1.3		

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.

noise/distortion performance

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNITS
THD+N	Total harmonic distortion plus noise	$V_{O(pp)}$ = V_{DD}/2 V, R_L = 100 Ω, f = 100 Hz	A_V = 1	25°C		0.025		
			A_V = 10			0.035		
			A_V = 100			0.15		
V_n	Equivalent input noise voltage	f = 100 Hz				55		nV/√Hz
		f = 10 kHz				10		
I_n	Equivalent input noise current	f = 1 kHz				0.31		fA/√Hz

shutdown characteristics

PARAMETER		TEST CONDITIONS	T_A†	MIN	TYP	MAX	UNITS
$I_{DD(SHDN)}$	Supply current in shutdown mode (per channel) (TLV4110, TLV4113)	$\overline{SHDN}$ = 0 V	25°C		3.4	10	µA
			Full range			15	
$t_{(ON)}$	Amplifier turnon time‡	R_L = 100 Ω	25°C		1		µs
$t_{(Off)}$	Amplifier turnoff time‡				3.3		

† Full range is 0°C to 70°C for C suffix and –40°C to 125°C for I suffix. If not specified, full range is –40°C to 125°C.
‡ Disable time and enable time are defined as the interval between application of the logic signal to $\overline{SHDN}$ and the point at which the supply current has reached half its final value.

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	vs Common-mode input voltage	1, 2
CMRR	Common-mode rejection ratio	vs Frequency	3
V_{OH}	High-level output voltage	vs High-level output current	4, 6
V_{OL}	Low-level output voltage	vs Low-level output current	5, 7
Z_o	Output impedance	vs Frequency	8
I_{DD}	Supply current	vs Supply voltage	9
k_{SVR}	Power supply voltage rejection ratio	vs Frequency	10
A_{VD}	Differential voltage amplification and phase	vs Frequency	11
	Gain-bandwidth product	vs Supply voltage	12
SR	Slew rate	vs Supply voltage	13
		vs Temperature	14
	Total harmonic distortion+noise	vs Frequency	15
V_n	Equivalent input voltage noise	vs Frequency	16
	Phase margin	vs Capacitive load	17
	Voltage-follower signal pulse response		18, 19
	Inverting large-signal pulse response		20, 21
	Small-signal inverting pulse response		22
	Crosstalk	vs Frequency	23
	Shutdown forward and reverse isolation		24
	Shutdown supply current	vs Free-air temperature	25
	Shutdown supply current/output voltage		26

TYPICAL CHARACTERISTICS

INPUT OFFSET VOLTAGE vs COMMON-MODE INPUT VOLTAGE

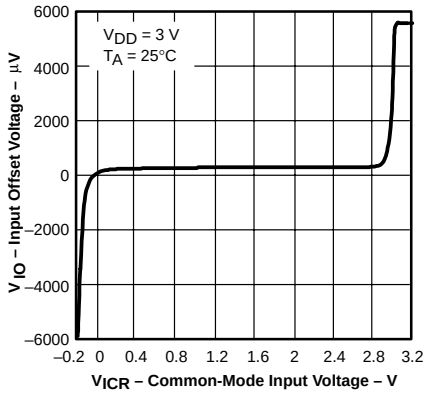


Figure 1

INPUT OFFSET VOLTAGE vs COMMON-MODE INPUT VOLTAGE

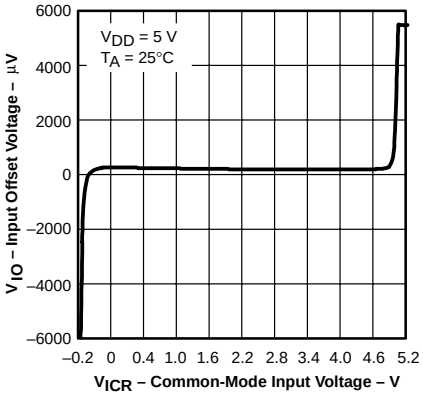


Figure 2

COMMON-MODE REJECTION RATIO vs FREQUENCY

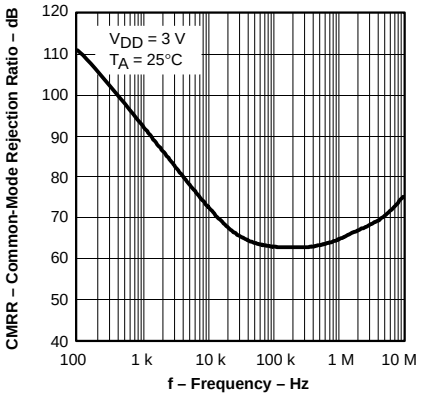


Figure 3

HIGH-LEVEL OUTPUT VOLTAGE vs HIGH-LEVEL OUTPUT CURRENT

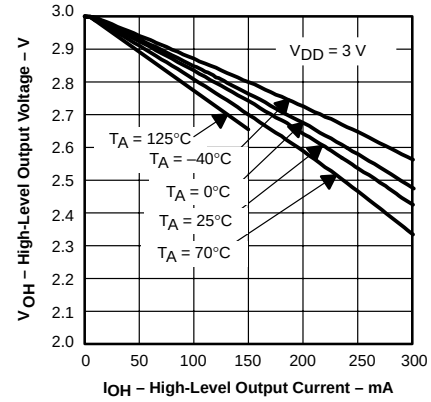


Figure 4

LOW-LEVEL OUTPUT VOLTAGE vs LOW-LEVEL OUTPUT CURRENT

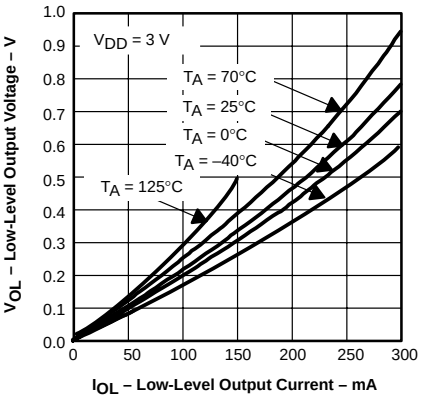


Figure 5

HIGH-LEVEL OUTPUT VOLTAGE vs HIGH-LEVEL OUTPUT CURRENT

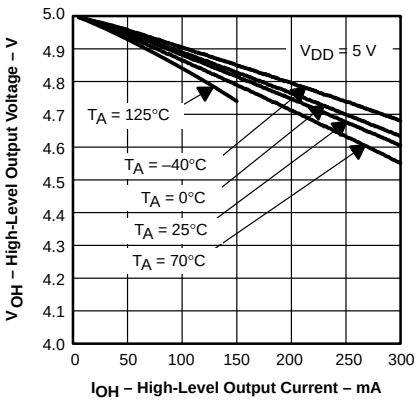


Figure 6

LOW-LEVEL OUTPUT VOLTAGE vs LOW-LEVEL OUTPUT CURRENT

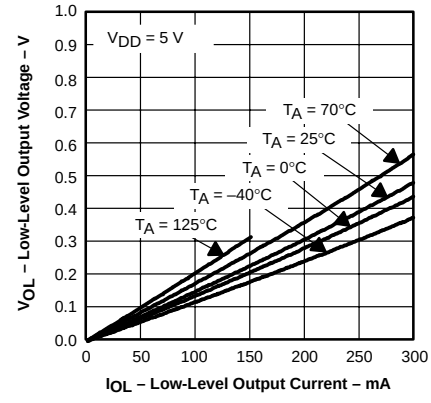


Figure 7

OUTPUT IMPEDANCE vs FREQUENCY

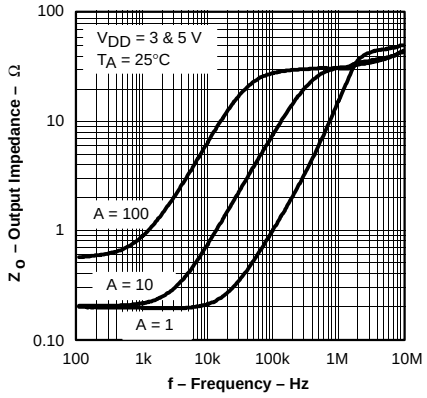


Figure 8

SUPPLY CURRENT vs SUPPLY VOLTAGE

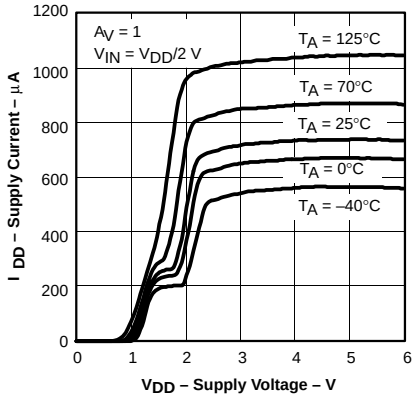


Figure 9

TYPICAL CHARACTERISTICS

POWER SUPPLY REJECTION RATIO vs FREQUENCY

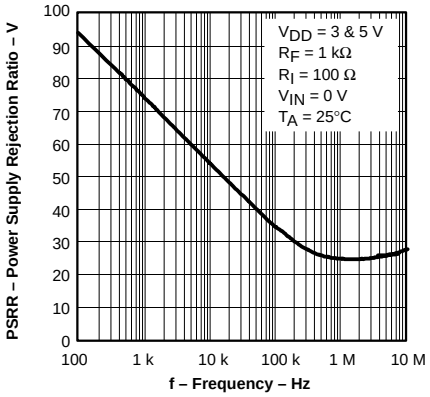


Figure 10

DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE vs FREQUENCY

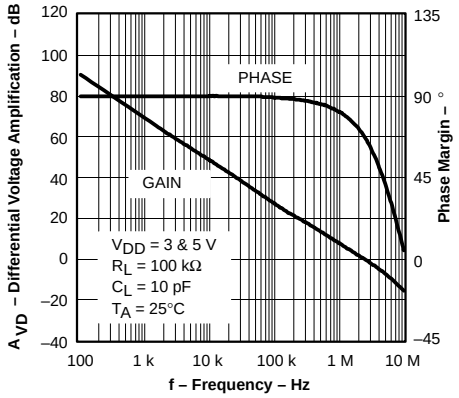


Figure 11

GAIN-BANDWIDTH PRODUCT vs SUPPLY VOLTAGE

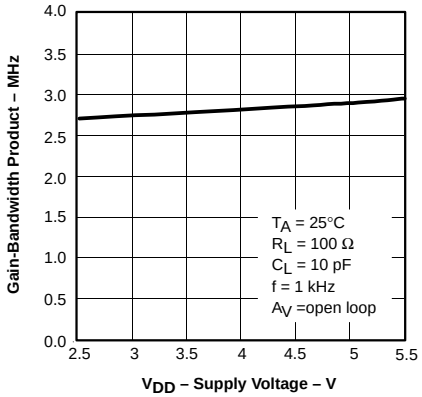


Figure 12

SLEW RATE vs SUPPLY VOLTAGE

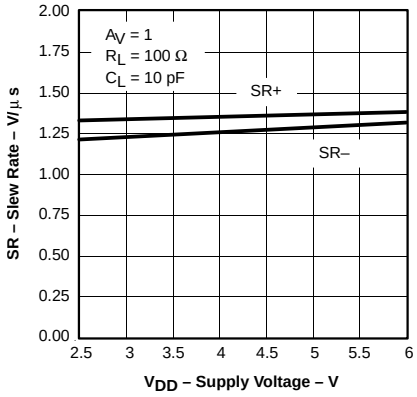


Figure 13

SLEW RATE vs TEMPERATURE

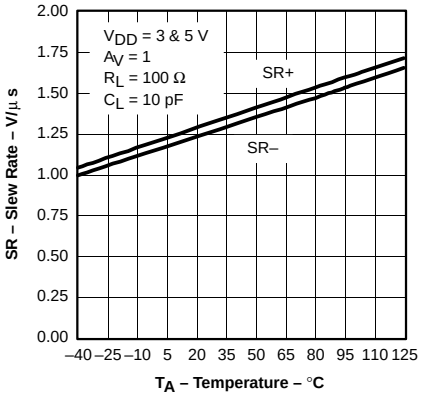


Figure 14

TOTAL HARMONIC DISTORTION+NOISE vs FREQUENCY

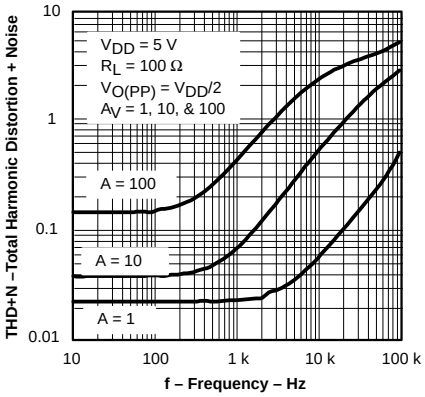


Figure 15

EQUIVALENT INPUT VOLTAGE NOISE vs FREQUENCY

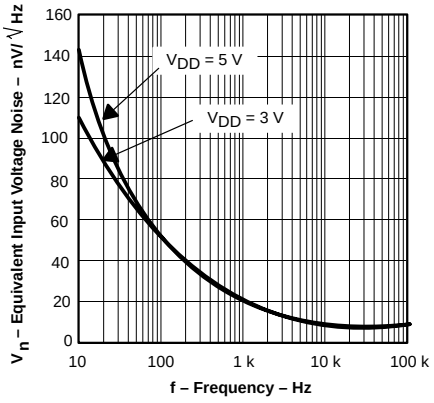


Figure 16

PHASE MARGIN vs CAPACITIVE LOAD

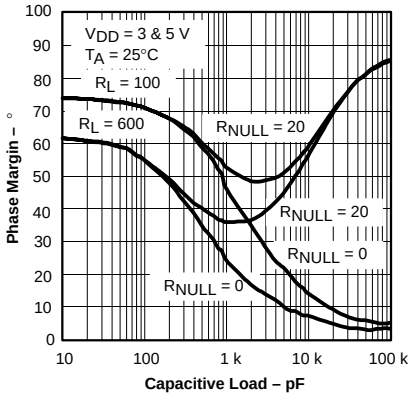


Figure 17

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE

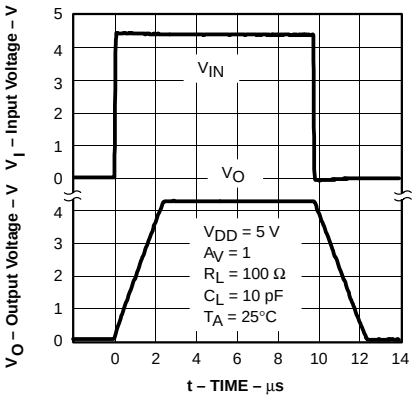


Figure 18

VOLTAGE-FOLLOWER SMALL-SIGNAL PULSE RESPONSE

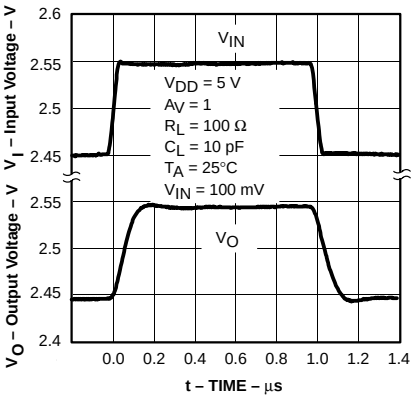


Figure 19

INVERTING LARGE-SIGNAL PULSE RESPONSE

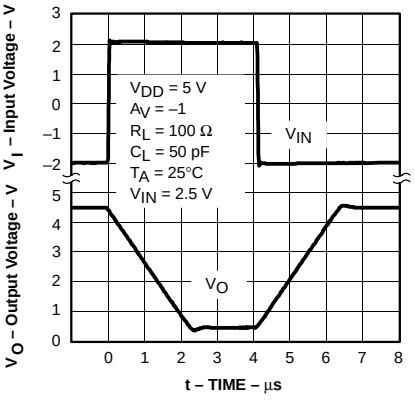


Figure 20

INVERTING LARGE-SIGNAL PULSE RESPONSE

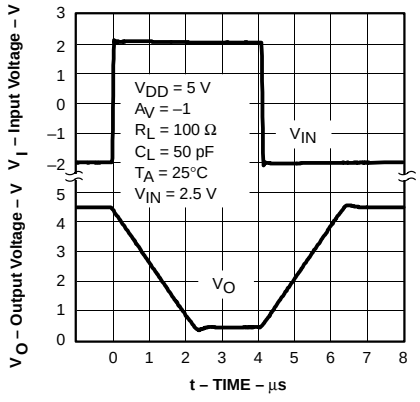


Figure 21

SMALL-SIGNAL INVERTING PULSE RESPONSE

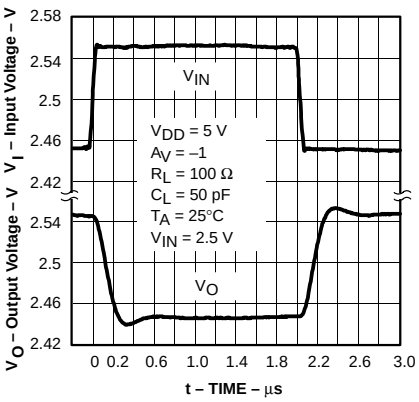


Figure 22

CROSSTALK vs FREQUENCY

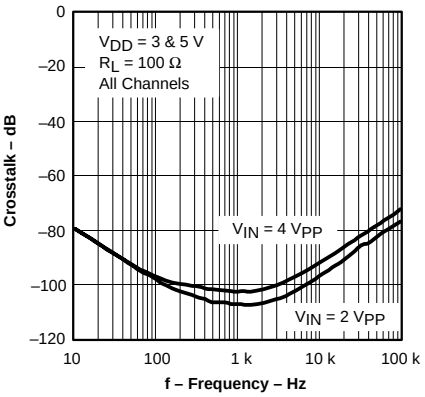


Figure 23

SHUTDOWN FORWARD AND REVERSE ISOLATION

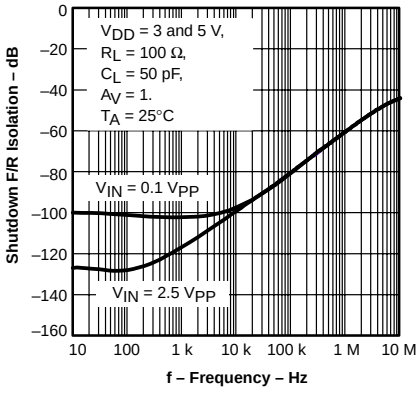


Figure 24

SHUTDOWN SUPPLY CURRENT vs FREE-AIR TEMPERATURE

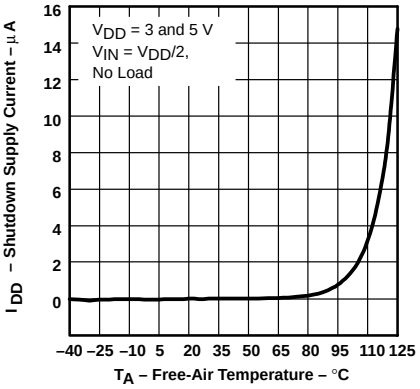


Figure 25

TYPICAL CHARACTERISTICS

SHUTDOWN SUPPLY CURRENT / OUTPUT VOLTAGE

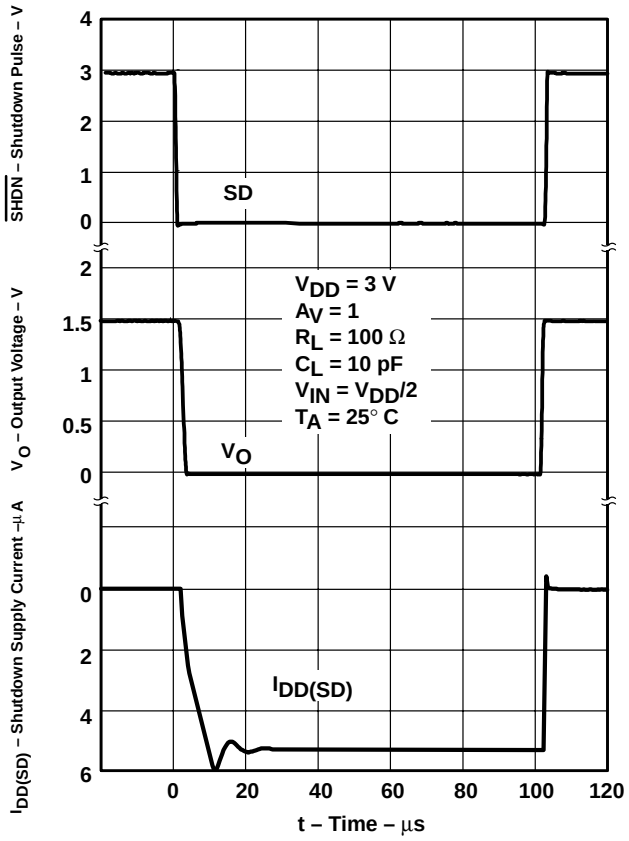


Figure 26

APPLICATION INFORMATION

shutdown function

Two members of the TLV411x family (TLV4110/3) have a shutdown terminal for conserving battery life in portable applications. When the shutdown terminal is tied low, the supply current is reduced to just nano amps per channel, the amplifier is disabled, and the outputs are placed in a high impedance mode. In order to save power in shutdown mode, an external pullup resistor is required, therefore, to enable the amplifier the shutdown terminal must be pulled high. When the shutdown terminal is left floating, care should be taken to ensure that parasitic leakage current at the shutdown terminal does not inadvertently place the operational amplifier into shutdown.

driving a capacitive load

When the amplifier is configured in this manner, capacitive loading directly on the output will decrease the device's phase margin leading to high frequency ringing or oscillations. Therefore, for capacitive loads of greater than 1 nF, it is recommended that a resistor be placed in series (R_{NULL}) with the output of the amplifier, as shown in Figure 27. A maximum value of 20 Ω should work well for most applications.

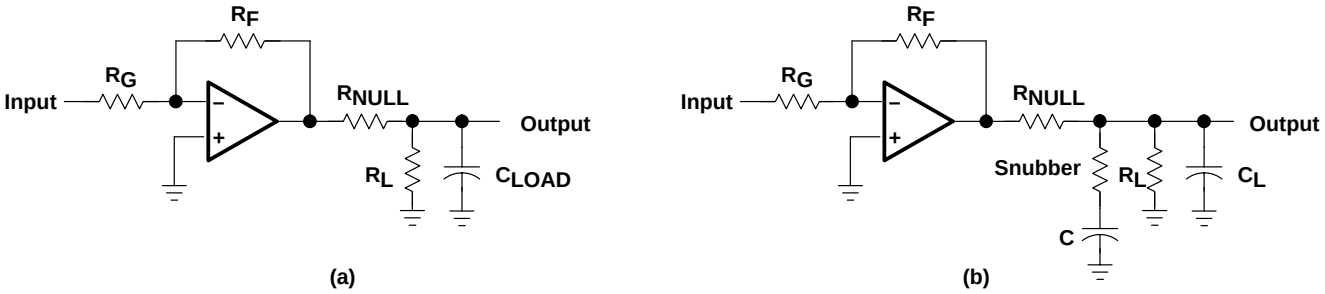
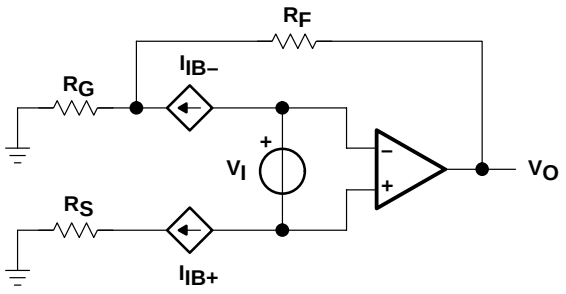


Figure 27. Driving a Capacitive Load

offset voltage

The output offset voltage, (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:



$$V_{OO} = V_{IO}\left(1 + \left(\frac{R_F}{R_G}\right)\right) \pm I_{IB+}\, R_S \left(1 + \left(\frac{R_F}{R_G}\right)\right) \pm I_{IB-}\, R_F$$

Figure 28. Output Offset Voltage Model

APPLICATION INFORMATION

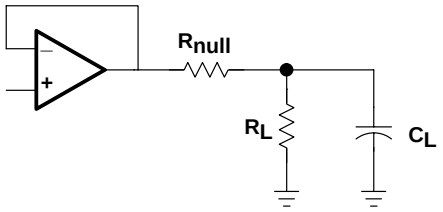


Figure 29

general power design considerations

When driving heavy loads at high junction temperatures there is an increased probability of electromigration affecting the long term reliability of ICs. Therefore for this not to be an issue either:

- The output current must be limited (at these high junction temperatures).

or

- The junction temperature must be limited.

The maximum continuous output current at a die temperature 150°C will be 1/3 of the current at 105°C.

The junction temperature will be dependent on the ambient temperature around the IC, thermal impedance from the die to the ambient and power dissipated within the IC.

$$T_J = T_A + \theta_{JA} \times P_{DIS}$$

Where:

P_{DIS} is the IC power dissipation and is equal to the output current multiplied by the voltage dropped across the output of the IC.
θ_{JA} is the thermal impedance between the junction and the ambient temperature of the IC.
T_J is the junction temperature.
T_A is the ambient temperature.

Reducing one or more of these factors results in a reduced die temperature. The 8-pin SOIC (small outline integrated circuit) has a thermal impedance from junction to ambient of 176°C/W. For this reason it is recommended that the maximum power dissipation of the 8-pin SOIC package be limited to 350 mW, with peak dissipation of 700 mW as long as the RMS value is less than 350 mW.

The use of the MSOP PowerPAD™ dramatically reduces the thermal impedance from junction to case. And with correct mounting, the reduced thermal impedance greatly increases the IC's permissible power dissipation and output current handling capability. For example, the power dissipation of the PowerPAD™ is increased to above 1 W. Sinusoidal and pulse-width modulated output signals also increase the output current capability. The equivalent dc current is proportional to the square-root of the duty cycle:

$$I_{DC(EQ)} = I_{Cont} \times \sqrt{(\text{duty cycle})}$$

CURRENT DUTY CYCLE AT PEAK RATED CURRENT	EQUIVALENT DC CURRENT AS A PERCENTAGE OF PEAK
100	100
70	84
50	71

Note that with an operational amplifier, a duty cycle of 70% would often result in the op amp sourcing current 70% of the time and sinking current 30%, therefore, the equivalent dc current would still be 0.84 times the continuous current rating at a particular junction temperature.

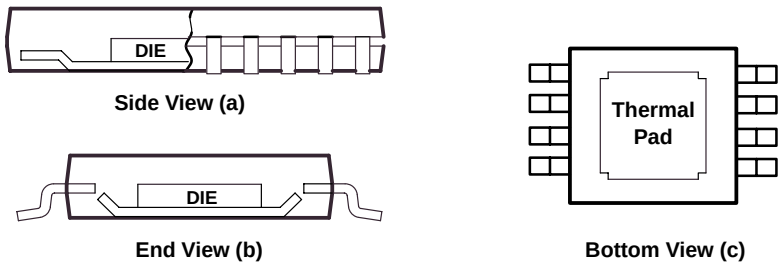
APPLICATION INFORMATION

general PowerPAD design considerations

The TLV411x is available in a thermally-enhanced PowerPAD family of packages. These packages are constructed using a downset leadframe upon which the die is mounted [see Figure 30(a) and Figure 30(b)]. This arrangement results in the lead frame being exposed as a thermal pad on the underside of the package [see Figure 30(c)]. Because this thermal pad has direct thermal contact with the die, excellent thermal performance can be achieved by providing a good thermal path away from the thermal pad.

The PowerPAD package allows for both assembly and thermal management in one manufacturing operation. During the surface-mount solder operation (when the leads are being soldered), the thermal pad can also be soldered to a copper area underneath the package. Through the use of thermal paths within this copper area, heat can be conducted away from the package into either a ground plane or other heat dissipating device.

The PowerPAD package represents a breakthrough in combining the small area and ease of assembly of surface mount with the, heretofore, awkward mechanical methods of heatsinking.



NOTE A: The thermal pad is electrically isolated from all terminals in the package.

Figure 30. Views of Thermally Enhanced DGN Package

Although there are many ways to properly heatsink the PowerPAD package, the following steps illustrate the recommended approach.

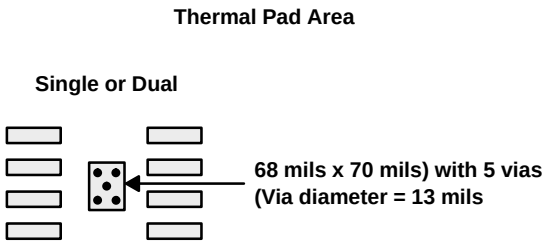


Figure 31. PowerPAD PCB Etch and Via Pattern

APPLICATION INFORMATION

general PowerPAD design considerations (continued)

1. Prepare the PCB with a top side etch pattern as shown in Figure 31. There should be etch for the leads as well as etch for the thermal pad.
2. Place five holes (dual) or nine holes (quad) in the area of the thermal pad. These holes should be 13 mils in diameter. Keep them small so that solder wicking through the holes is not a problem during reflow.
3. Additional vias may be placed anywhere along the thermal plane outside of the thermal pad area. This helps dissipate the heat generated by the TLV411x IC. These additional vias may be larger than the 13-mil diameter vias directly under the thermal pad. They can be larger because they are not in the thermal pad area to be soldered so that wicking is not a problem.
4. Connect all holes to the internal ground plane.
5. When connecting these holes to the ground plane, do not use the typical web or spoke via connection methodology. Web connections have a high thermal resistance connection that is useful for slowing the heat transfer during soldering operations. This makes the soldering of vias that have plane connections easier. In this application, however, low thermal resistance is desired for the most efficient heat transfer. Therefore, the holes under the TLV411x PowerPAD package should make their connection to the internal ground plane with a complete connection around the entire circumference of the plated-through hole.
6. The top-side solder mask should leave the terminals of the package and the thermal pad area with its five holes (dual) or nine holes (quad) exposed. The bottom-side solder mask should cover the five or nine holes of the thermal pad area. This prevents solder from being pulled away from the thermal pad area during the reflow process.
7. Apply solder paste to the exposed thermal pad area and all of the IC terminals.
8. With these preparatory steps in place, the TLV411x IC is simply placed in position and run through the solder reflow operation as any standard surface-mount component. This results in a part that is properly installed.

For a given θ_{JA}, the maximum power dissipation is shown in Figure 32 and is calculated by the following formula:

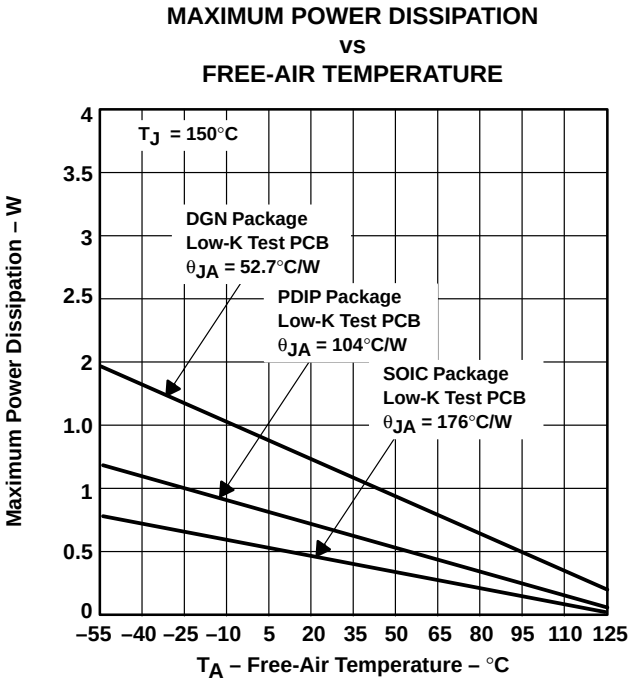
$$P_D = \left(\frac{T_{MAX} - T_A}{\theta_{JA}} \right)$$

Where:

P_D = Maximum power dissipation of TLV411x IC (watts)
T_{MAX} = Absolute maximum junction temperature (150°C)
T_A = Free-ambient air temperature (°C)
$\theta_{JA} = \theta_{JC} + \theta_{CA}$
θ_{JC} = Thermal coefficient from junction to case
θ_{CA} = Thermal coefficient from case to ambient air (°C/W)

APPLICATION INFORMATION

general PowerPAD design considerations (continued)



NOTE A: Results are with no air flow and using JEDEC Standard Low-K test PCB.

Figure 32. Maximum Power Dissipation vs Free-Air Temperature

The next consideration is the package constraints. The two sources of heat within an amplifier are quiescent power and output power. The designer should never forget about the quiescent heat generated within the device, especially multi-amplifier devices. Because these devices have linear output stages (Class A-B), most of the heat dissipation is at low output voltages with high output currents.

The other key factor when dealing with power dissipation is how the devices are mounted on the PCB. The PowerPAD devices are extremely useful for heat dissipation. But, the device should always be soldered to a copper plane to fully use the heat dissipation properties of the PowerPAD. The SOIC package, on the other hand, is highly dependent on how it is mounted on the PCB. As more trace and copper area is placed around the device, θ_{JA} decreases and the heat dissipation capability increases. The currents and voltages shown in these graphs are for the total package. For the dual or quad amplifier packages, the sum of the RMS output currents and voltages should be used to choose the proper package.

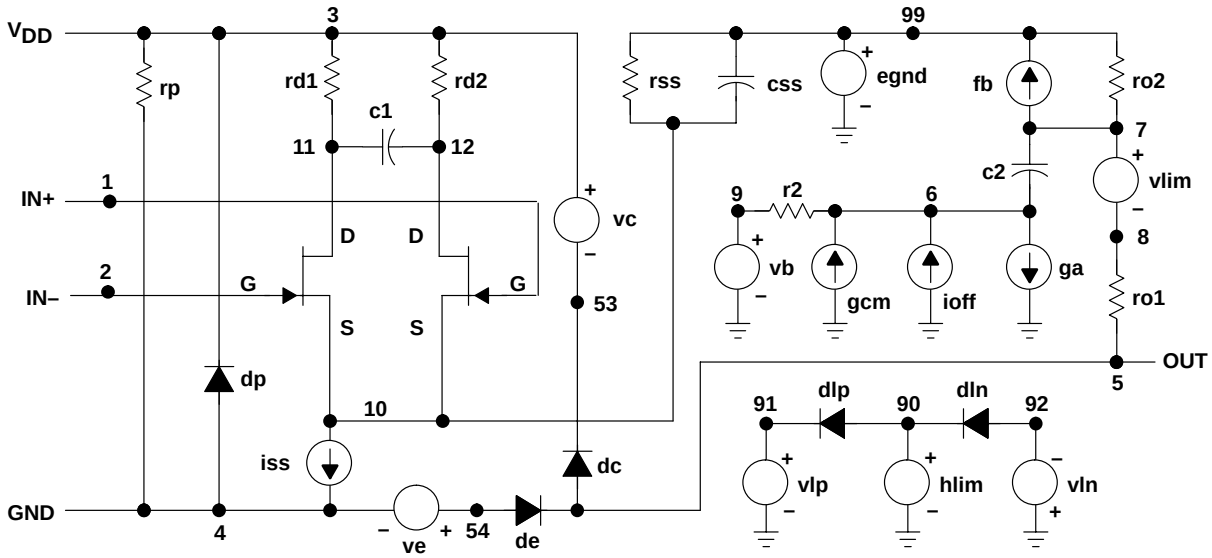
APPLICATION INFORMATION

macromodel information

Macromodel information provided was derived using Microsim *Parts*™, the model generation software used with Microsim *PSpice*™. The Boyle macromodel (see Note 3) and subcircuit in Figure 33 are generated using the TLV411x typical electrical and operating characteristics at T_A = 25°C. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 3: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers," *IEEE Journal of Solid-State Circuits,* SC-9, 353 (1974).




```
* TLV4112_5V operational amplifier "macromodel" subcircuit
* updated using Model Editor release 9.1 on 01/18/00 at 15:50
Model Editor is an OrCAD product.
*
* connections:   non–inverting input
*                | inverting input
*                | | positive power supply
*                | | | negative power supply
*                | | | | output
*                | | | | |
.subckt TLV4112_5V      1 2 3 4 5
*
c1      11      12      2.2439E–12
c2      6       7       10.000E–12
css     10      99      454.55E–15
dc      5       53      dy
de      54      5       dy
dlp     90      91      dx
dln     92      90      dx
dp      4       3       dx
egnd    99      0       poly(2) (3,0) (4,0) 0 .5 .5
fb      7       99      poly(5) vb vc ve vlp vln 0
+ 33.395E6 –1E3 1E3 33E6 –33E6
ga      6       0       11      12      168.39E–6
gcm     0       6       10      99      168.39E–12
iss     10      4       dc      13.800E–6
hlim    90      0       vlim 1K
ioff    0       6       dc      75E–9
j1      11      2       10 jx1
J2      12      1       10 jx2
r2      6       9       100.00E3
rd1     3       11      5.9386E3
rd2     3       12      5.9386E3
ro1     8       5       10
ro2     7       99      10
rp      3       4       3.3333E3
rss     10      99      14.493E6
vb      9       0       dc 0
vc      3       53      dc .86795
ve      54      4       dc .86795
vlim    7       8       dc 0
vlp     91      0       dc 300
vln     0       92      dc 300
.model  dx      D(Is=800.00E–18)
.model  dy      D(Is=800.00E–18 Rs=1m Cjo=10p)
.model  jx1     NJF(Is=150.00E–12 Beta=2.0547E–3 +Vto=–1)
.model  jx2     NJF(Is=150.00E–12 Beta=2.0547E–3 + Vto=–1)
.ends
*$
```

Figure 33. Boyle Macromodel and Subcircuit

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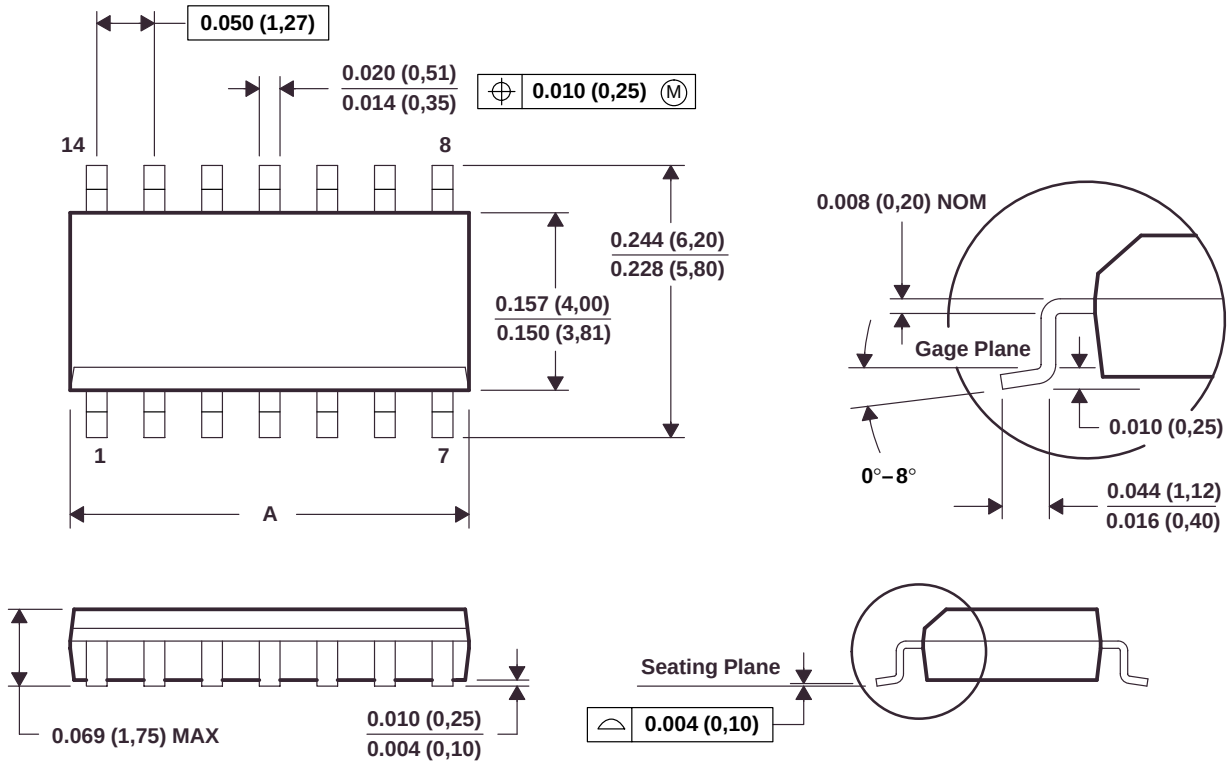
MECHANICAL DATA

D (R-PDSO-G)** **PLASTIC SMALL-OUTLINE PACKAGE**

14 PIN SHOWN



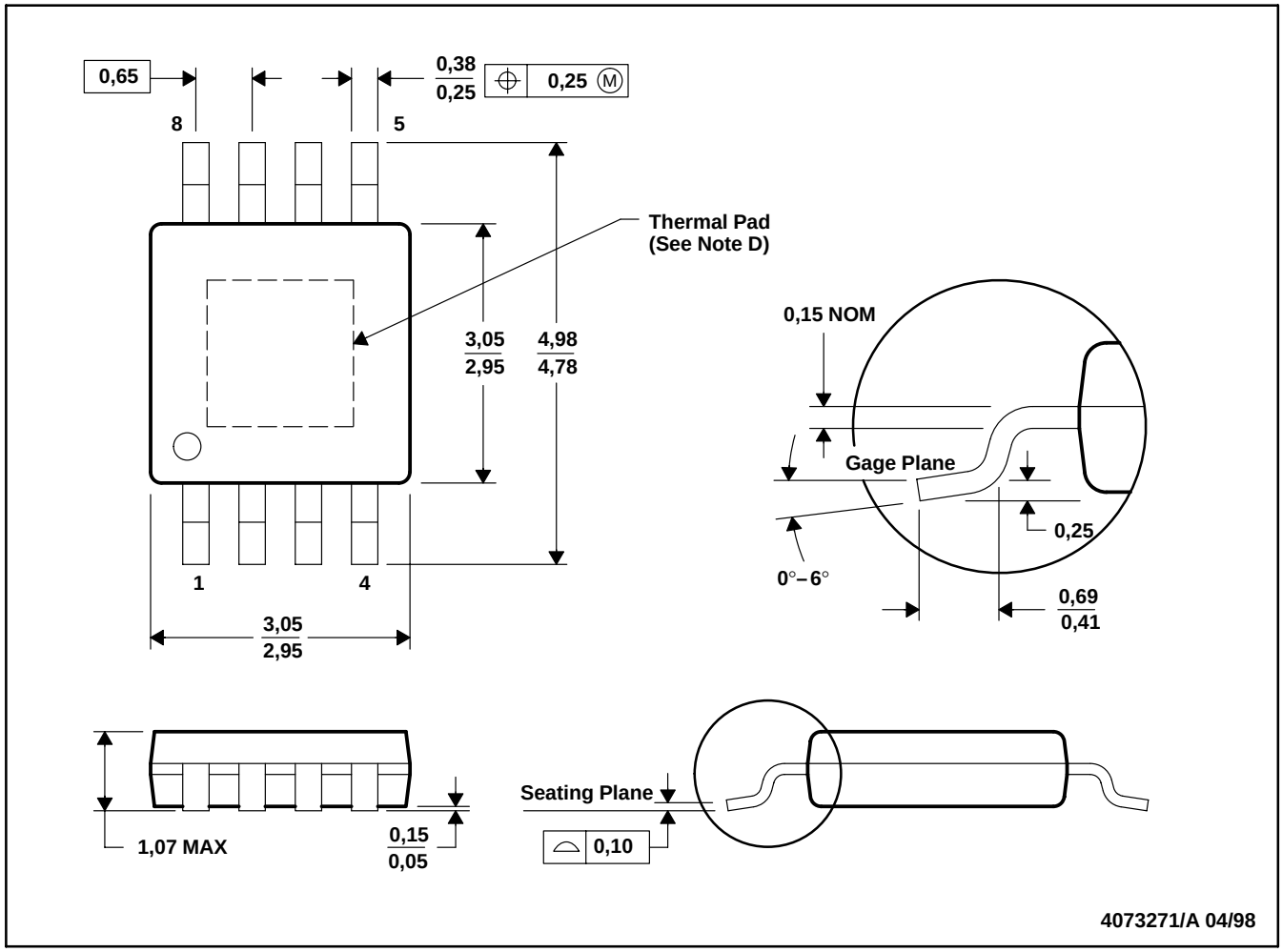
DIM \ PINS **	8	14
A MAX	0.197 (5,00)	0.344 (8,75)
A MIN	0.189 (4,80)	0.337 (8,55)

4040047/D 10/96

NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
D. Falls within JEDEC MS-012

MECHANICAL INFORMATION

DGN (S-PDSO-G8) **PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE**

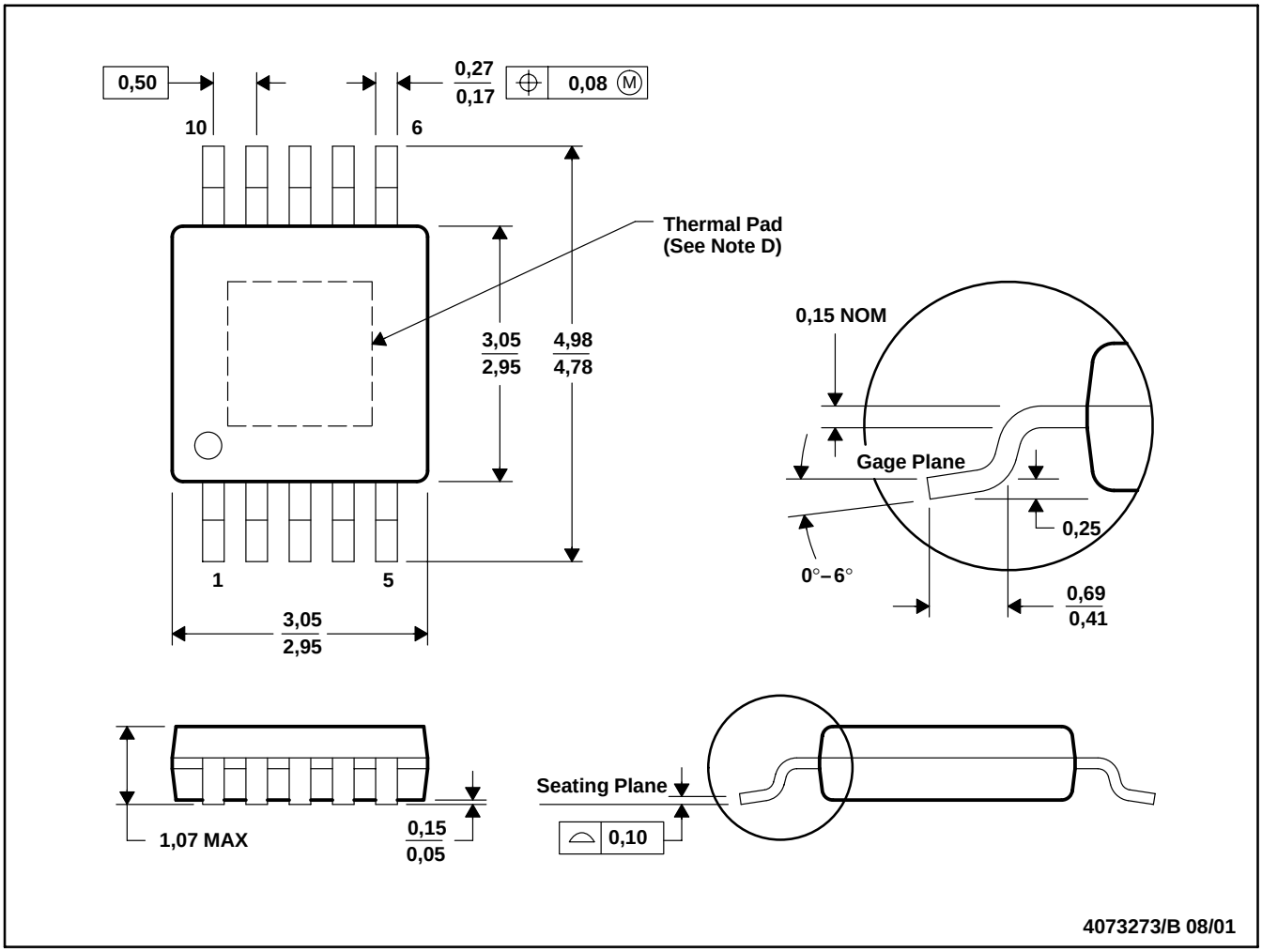


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions include mold flash or protrusions.
D. The package thermal performance may be enhanced by attaching an external heat sink to the thermal pad. This pad is electrically and thermally connected to the backside of the die and possibly selected leads.
E. Falls within JEDEC MO-187

PowerPAD is a trademark of Texas Instruments.

MECHANICAL INFORMATION

DGQ (S-PDSO-G10) **PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE**



NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion.
D. The package thermal performance may be enhanced by bonding the thermal pad to an external thermal plane. This pad is electrically and thermally connected to the backside of the die and possibly selected leads.
E. Falls within JEDEC MO-187

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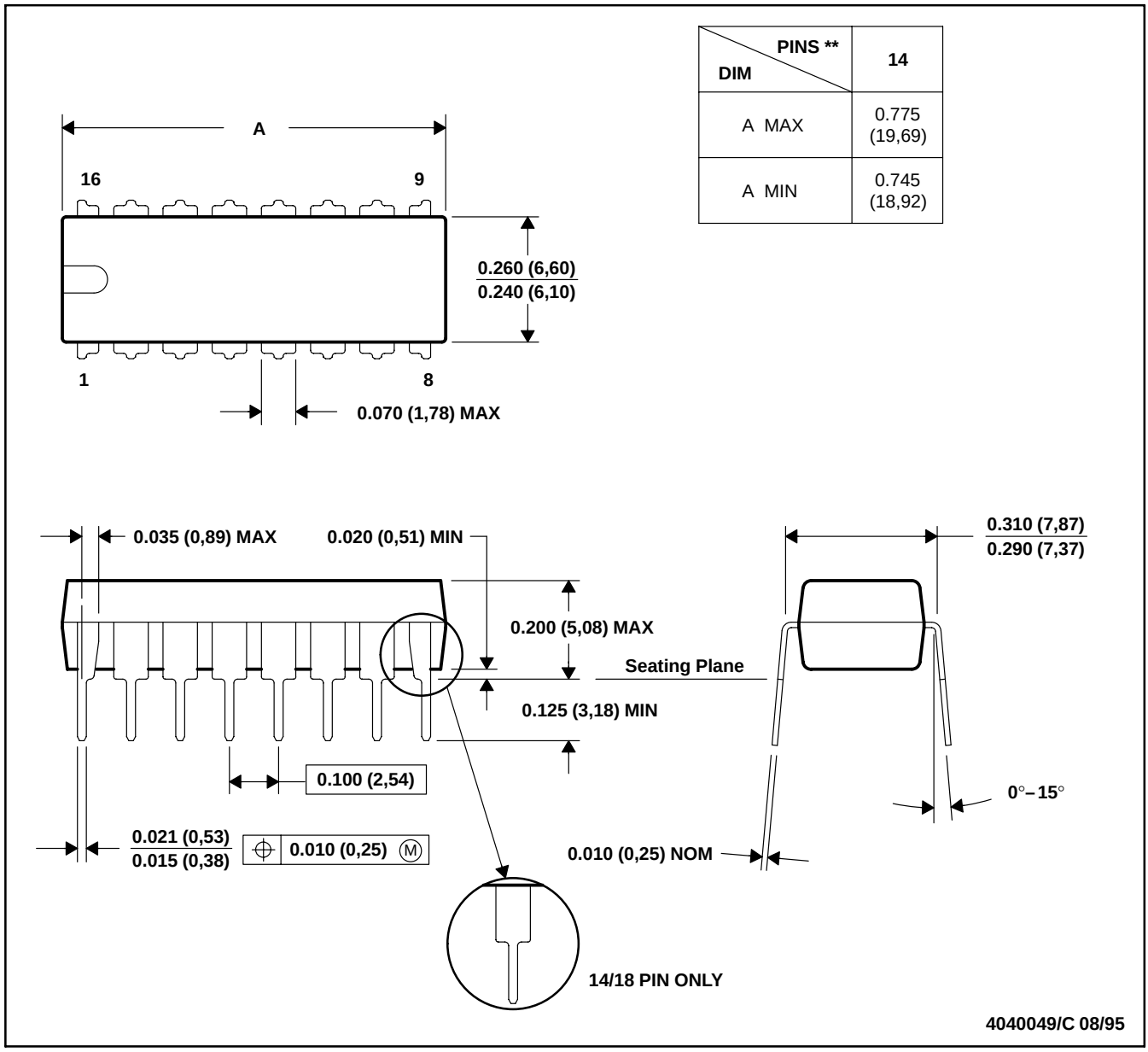
MECHANICAL INFORMATION

N (R-PDIP-T)** **PLASTIC DUAL-IN-LINE PACKAGE**

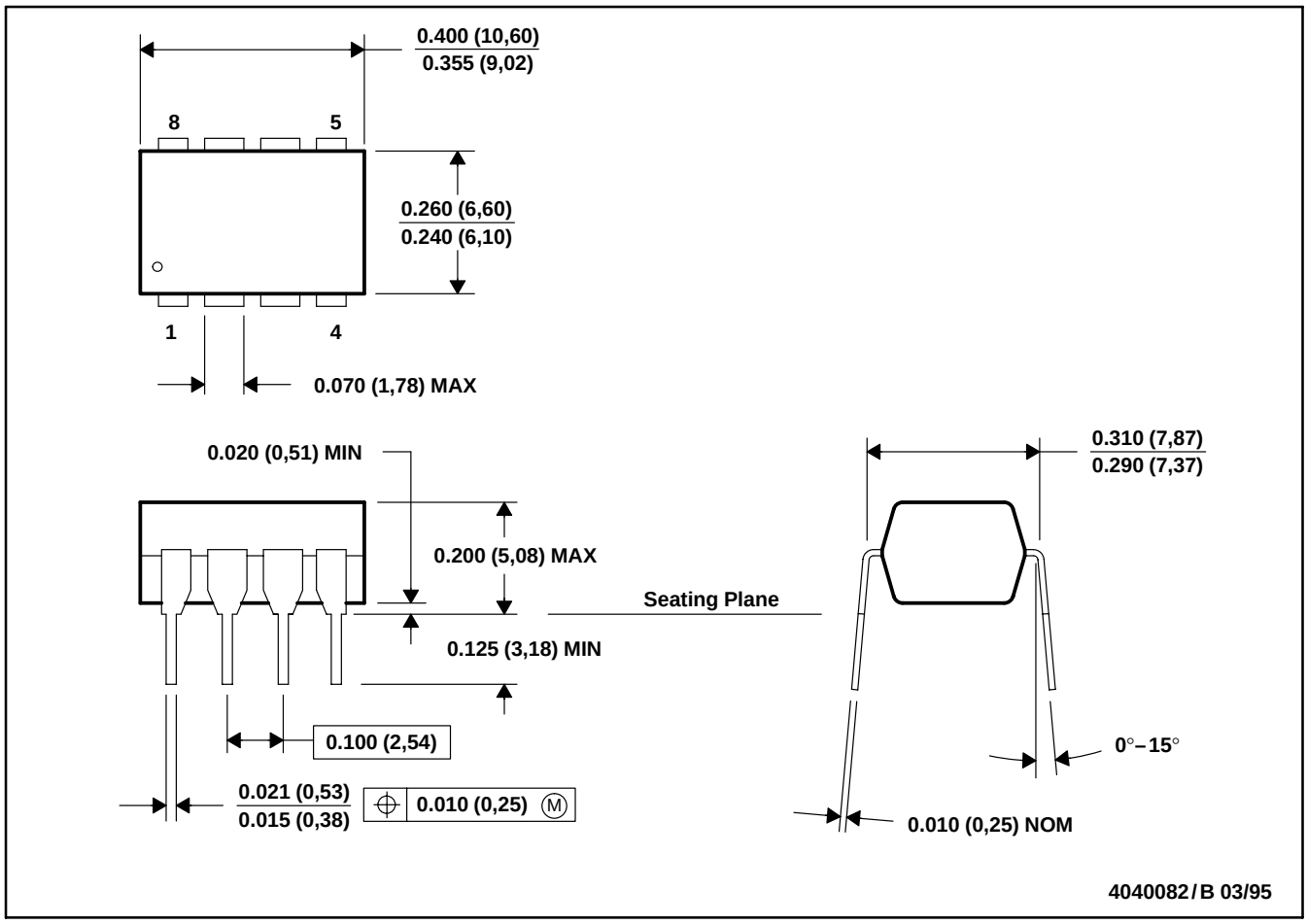
16 PIN SHOWN



NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-001 (20 pin package is shorter then MS-001.)

MECHANICAL INFORMATION

P (R-PDIP-T8) **PLASTIC DUAL-IN-LINE PACKAGE**



NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-001

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