

FEATURES

- LOW THERMAL RESISTANCE — 1.1°C/W
- CURRENT FOLDOVER PROTECTION
- EXCELLENT LINEARITY — Class A/B Output
- WIDE SUPPLY RANGE — $\pm 10\text{V}$ to $\pm 45\text{V}$
- HIGH OUTPUT CURRENT — Up to $\pm 15\text{A}$ Peak

APPLICATIONS

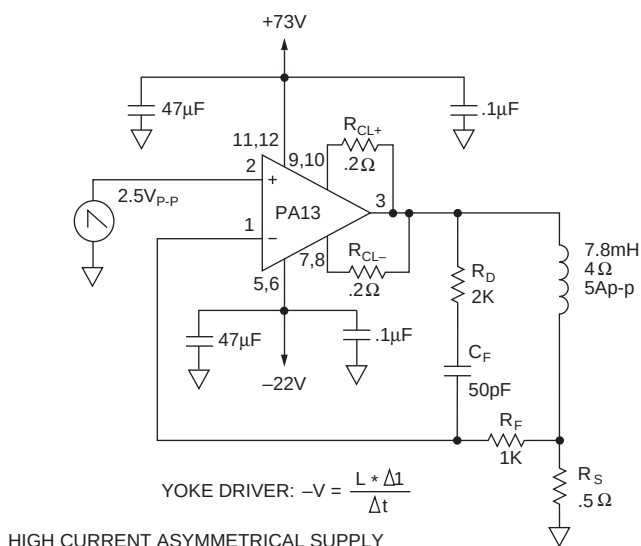
- MOTOR, VALVE AND ACTUATOR CONTROL
- MAGNETIC DEFLECTION CIRCUITS UP TO 10A
- POWER TRANSDUCERS UP TO 100kHz
- TEMPERATURE CONTROL UP TO 360W
- PROGRAMMABLE POWER SUPPLIES UP TO 90V
- AUDIO AMPLIFIERS UP TO 120W RMS

DESCRIPTION

The PA13 is a state of the art high voltage, very high output current operational amplifier designed to drive resistive, inductive and capacitive loads. For optimum linearity, especially at low levels, the output stage is biased for class A/B operation using a thermistor compensated base-emitter voltage multiplier circuit. The safe operating area (SOA) can be observed for all operating conditions by selection of user programmable current limiting resistors. For continuous operation under load, a heatsink of proper rating is recommended.

This hybrid integrated circuit utilizes thick film (cermet) resistors, ceramic capacitors and semiconductor chips to maximize reliability, minimize size and give top performance. Ultrasonically bonded aluminum wires provide reliable interconnections at all operating temperatures. The 12-pin power SIP package is electrically isolated.

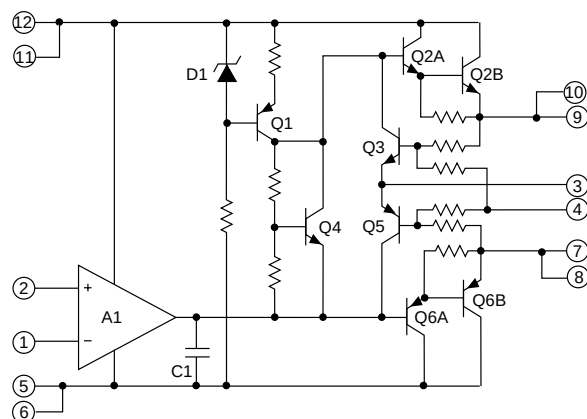
TYPICAL APPLICATION



HIGH CURRENT ASYMMETRICAL SUPPLY



EQUIVALENT SCHEMATIC



POWER RATING

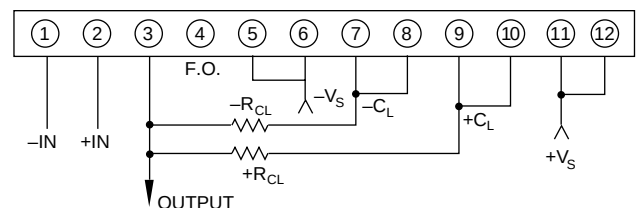
Not all vendors use the same method to rate the power handling capability of a Power Op Amp. APEX rates the internal dissipation, which is consistent with rating methods used by transistor manufacturers and gives conservative results. Rating delivered power is highly application dependent and therefore can be misleading. For example, the 135W internal dissipation rating of the PA13 could be expressed as an output rating of 260W for audio (sine wave) or as 440W if using a single ended DC load. Please note that all vendors rate maximum power using an infinite heatsink.

THERMAL STABILITY

APEX has eliminated the tendency of class A/B output stages toward thermal runaway and thus has vastly increased amplifier reliability. This feature, not found in most other Power Op Amps, was pioneered by APEX in 1981 using thermistors which assure a negative temperature coefficient in the quiescent current. The reliability benefits of this added circuitry far outweigh the slight increase in component count.

EXTERNAL CONNECTIONS

Package: SIP03



ABSOLUTE MAXIMUM RATINGS

SUPPLY VOLTAGE, +Vs to -Vs
 OUTPUT CURRENT, within SOA
 POWER DISSIPATION, internal
 INPUT VOLTAGE, differential
 INPUT VOLTAGE, common mode
 TEMPERATURE, pin solder -10s
 TEMPERATURE, junction¹
 TEMPERATURE RANGE, storage
 OPERATING TEMPERATURE RANGE, case

PA13/PA13A

100V
 15A
 135W
 $\pm V_S - 3V$
 $\pm V_S$
 300°C
 175°C
 -65 to +150°C
 -55 to +125°C

SPECIFICATIONS

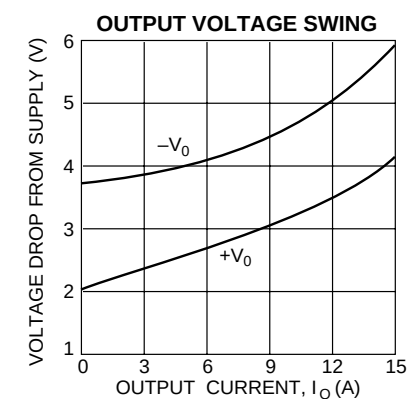
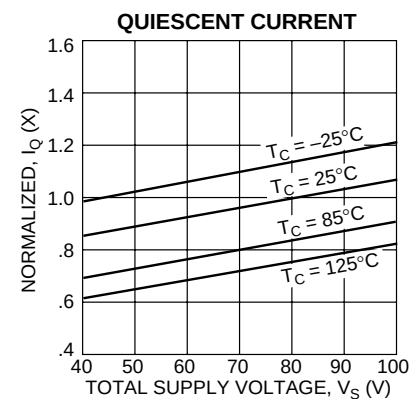
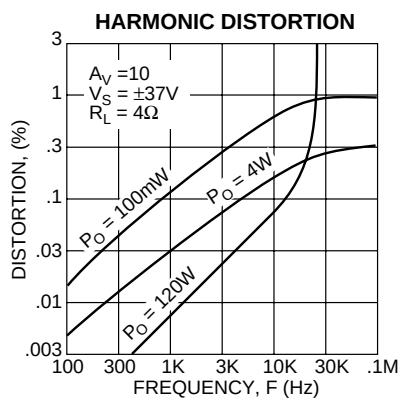
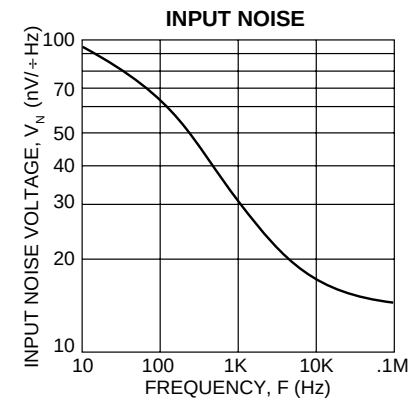
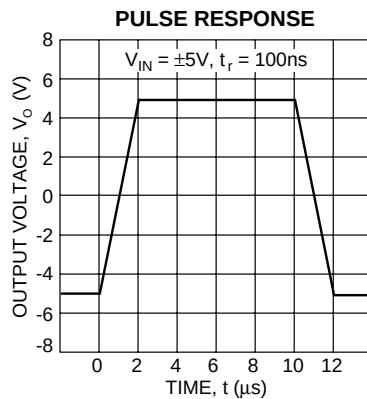
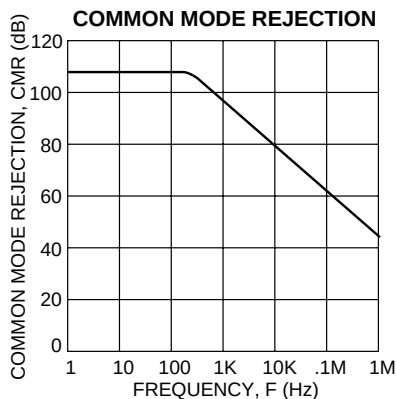
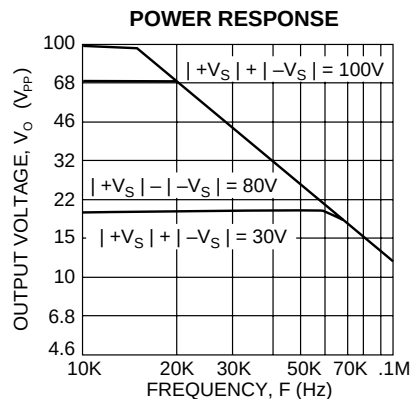
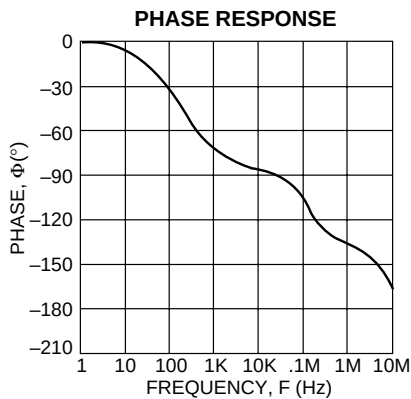
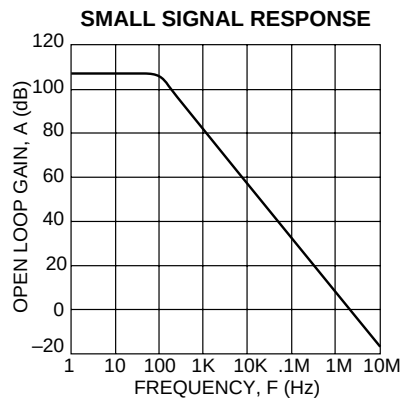
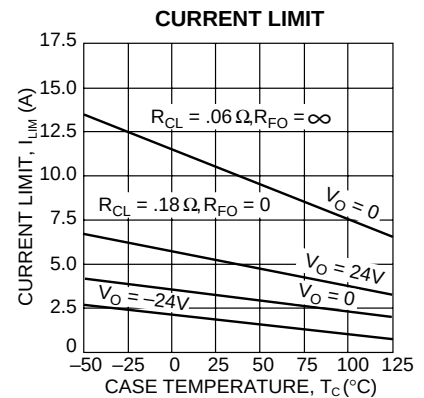
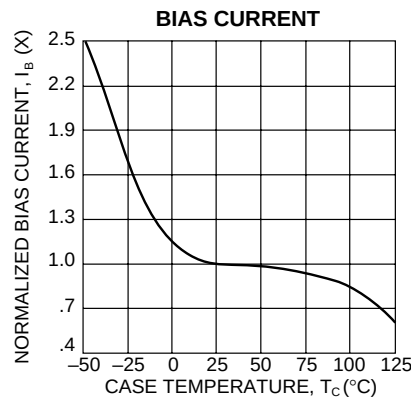
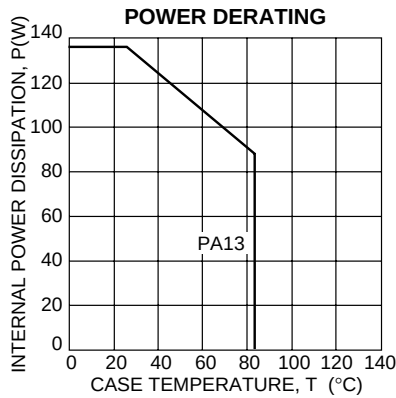
SPECIFICATIONS		PA13			PA13A			
PARAMETER	TEST CONDITIONS ^{2, 5}	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
INPUT								
OFFSET VOLTAGE, initial	T _C = 25°C		±2	±6		±1	±3	mV
OFFSET VOLTAGE, vs. temperature	Full temperature range		±10	±65		*	±40	μV/°C
OFFSET VOLTAGE, vs. supply	T _C = 25°C		±30	±200		*	*	μV/V
OFFSET VOLTAGE, vs. power	T _C = 25°C		±20			*		μV/W
BIAS CURRENT, initial	T _C = 25°C		±12	±30		±10	±20	nA
BIAS CURRENT, vs. temperature	Full temperature range		±50	±500		*	*	pA/°C
BIAS CURRENT, vs. supply	T _C = 25°C		±10			*		pA/V
OFFSET CURRENT, initial	T _C = 25°C		±12	±30		±5	±10	nA
OFFSET CURRENT, vs. temperature	Full temperature range		±50			*		pA/°C
INPUT IMPEDANCE, DC	T _C = 25°C		200			*		MΩ
INPUT CAPACITANCE	T _C = 25°C		3			*		pF
COMMON MODE VOLTAGE RANGE ³	Full temperature range	±V _S −5	±V _S −3		*	*		V
COMMON MODE REJECTION, DC	Full temp. range, V _{CM} = ±V _S −6V	74	100		*	*		dB
GAIN								
OPEN LOOP GAIN at 10Hz	T _C = 25°C, 1KΩ load		110			*		dB
OPEN LOOP GAIN at 10Hz	Full temp. range, 8Ω load	96	108		*	*		dB
GAIN BANDWIDTH PRODUCT @ 1MHz	T _C = 25°C, 8Ω load		4			*		MHz
POWER BANDWIDTH	T _C = 25°C, 8Ω load	13	20		*	*		kHz
PHASE MARGIN	Full temp. range, 8Ω load		20			*		°
OUTPUT								
VOLTAGE SWING ³	T _C = 25°C, PA13 = 10A, PA13A = 15A	±V _S −6			*			V
VOLTAGE SWING ³	T _C = 25°C, I _O = 5A	±V _S −5			*			V
VOLTAGE SWING ³	Full temp. range, I _O = 80mA	±V _S −5			*			V
CURRENT, peak	T _C = 25°C	10			15			A
SETTLING TIME to .1%	T _C = 25°C, 2V step		2			*		μs
SLEW RATE	T _C = 25°C	2.5	4		*	*		V/μs
CAPACITIVE LOAD	Full temperature range, A _V = 1			1.5			*	nF
CAPACITIVE LOAD	Full temperature range, A _V > 10			SOA			*	
POWER SUPPLY								
VOLTAGE	Full temperature range	±10	±40	±45	*	*	*	V
CURRENT, quiescent	T _C = 25°C		25	50		*	*	mA
THERMAL								
RESISTANCE, AC, junction to case ⁴	T _C = −55 to +125°C, F > 60Hz		.6	.7		*	*	°C/W
RESISTANCE, DC, junction to case	T _C = −55 to +125°C		.9	1.1		*	*	°C/W
RESISTANCE, DC, junction to air	T _C = −55 to +125°C		30			*	*	°C/W
TEMPERATURE RANGE, case	Meets full range specification	−25		+85	*		*	°C

NOTES: * The specification of PA13A is identical to the specification for PA13 in the applicable column to the left

1. Long term operation at the maximum junction temperature will result in reduced product life. Derate internal power dissipation to achieve high MTTF.
2. The power supply voltage for all tests is ± 40 , unless otherwise noted as a test condition.
3. $+V_S$ and $-V_S$ denote the positive and negative supply rail respectively. Total V_S is measured from $+V_S$ to $-V_S$.
4. Rating applies if the output current alternates between both output transistors at a rate faster than 60Hz.
5. Full temperature range specifications are guaranteed but not 100% tested.

CAUTION

The exposed substrate contains beryllia (BeO). Do not crush, machine, or subject to temperatures in excess of 850°C to avoid generating toxic fumes.



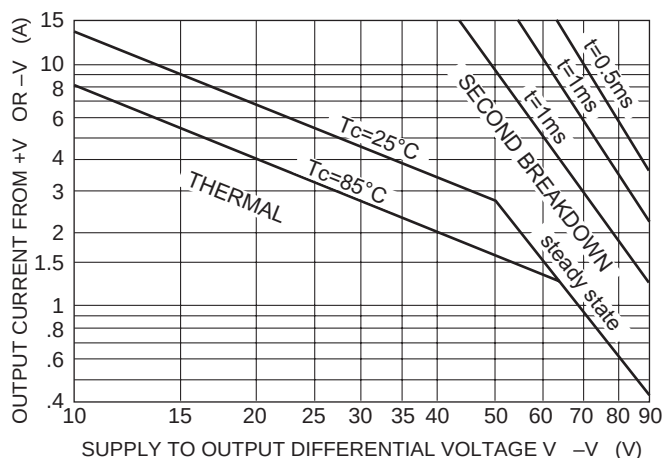
GENERAL

Please read Application Note 1 "General Operating Considerations" which covers stability, supplies, heat sinking, mounting, current limit, SOA interpretation, and specification interpretation. Visit www.apexmicrotech.com for design tools that help automate tasks such as calculations for stability, internal power dissipation, current limit; heat sink selection; Apex's complete Application Notes library; Technical Seminar Workbook; and Evaluation Kits.

SAFE OPERATING AREA (SOA)

The output stage of most power amplifiers has three distinct limitations:

1. The current handling capability of the transistor geometry and the wire bonds.
2. The second breakdown effect which occurs whenever the simultaneous collector current and collector-emitter voltage exceeds specified limits.
3. The junction temperature of the output transistors.



The SOA curves combine the effect of all limits for this Power Op Amp. For a given application, the direction and magnitude of the output current should be calculated or measured and checked against the SOA curves. This is simple for resistive loads but more complex for reactive and EMF generating loads. However, the following guidelines may save extensive analytical efforts.

1. Capacitive and dynamic* inductive loads up to the following maximum are safe with the current limits set as specified.

$\pm V_s$	CAPACITIVE LOAD		INDUCTIVE LOAD	
	$I_{LIM} = 5A$	$I_{LIM} = 10A$	$I_{LIM} = 5A$	$I_{LIM} = 10A$
50V	200 μ F	125 μ F	5mH	2.0mH
40V	500 μ F	350 μ F	15mH	3.0mH
35V	2.0mF	850 μ F	50mH	5.0mH
30V	7.0mF	2.5mF	150mH	10mH
25V	25mF	10mF	500mH	20mH
20V	60mF	20mF	1,000mH	30mH
15V	150mF	60mF	2,500mH	50mH

*If the inductive load is driven near steady state conditions, allowing the output voltage to drop more than 12.5V below the supply rail with $I_{LIM} = 10A$ or 27V below the supply rail with $I_{LIM} = 5A$ while the amplifier is current limiting, the inductor must be capacitively coupled or the current limit must be lowered to meet SOA criteria.

2. The amplifier can handle any EMF generating or reactive load and short circuits to the supply rail or common if the current limits are set as follows at $T_c = 25^\circ C$:

$\pm V_s$	SHORT TO $\pm V_s$ C, L, OR EMF LOAD	SHORT TO COMMON
45V	.43A	3.0A
40V	.65A	3.4A
35V	1.0A	3.9A
30V	1.7A	4.5A
25V	2.7A	5.4A
20V	3.4A	6.7A
15V	4.5A	9.0A

These simplified limits may be exceeded with further analysis using the operating conditions for a specific application.

CURRENT LIMITING

Refer to Application Note 9, "Current Limiting", for details of both fixed and foldover current limit operation. Visit the Apex web site at www.apexmicrotech.com for a copy of Power_design.exe which plots current limits vs. steady state SOA. Beware that current limit should be thought of as a $\pm 20\%$ function initially and varies about 2:1 over the range of $-55^\circ C$ to $125^\circ C$.

For fixed current limit, leave pin 4 open and use equations 1 and 2.

$$R_{CL} = 0.65/I_{CL} \quad (1)$$

$$I_{CL} = 0.65/R_{CL} \quad (2)$$

Where:

I_{CL} is the current limit in amperes.

R_{CL} is the current limit resistor in ohms.

For certain applications, foldover current limit adds a slope to the current limit which allows more power to be delivered to the load without violating the SOA. For maximum foldover slope, ground pin 4 and use equations 3 and 4.

$$I_{CL} = \frac{0.65 + (V_o * 0.014)}{R_{CL}} \quad (3)$$

$$R_{CL} = \frac{0.65 + (V_o * 0.014)}{I_{CL}} \quad (4)$$

Where:

V_o is the output voltage in volts.

Most designers start with either equation 1 to set R_{CL} for the desired current at 0v out, or with equation 4 to set R_{CL} at the maximum output voltage. Equation 3 should then be used to plot the resulting foldover limits on the SOA graph. If equation 3 results in a negative current limit, foldover slope must be reduced. This can happen when the output voltage is the opposite polarity of the supply conducting the current.

In applications where a reduced foldover slope is desired, this can be achieved by adding a resistor (R_{FO}) between pin 4 and ground. Use equations 4 and 5 with this new resistor in the circuit.

$$I_{CL} = \frac{0.65 + \frac{V_o * 0.14}{10.14 + R_{FO}}}{R_{CL}} \quad (5)$$

$$R_{CL} = \frac{0.65 + \frac{V_o * 0.14}{10.14 + R_{FO}}}{I_{CL}} \quad (6)$$

Where:

R_{FO} is in K ohms.