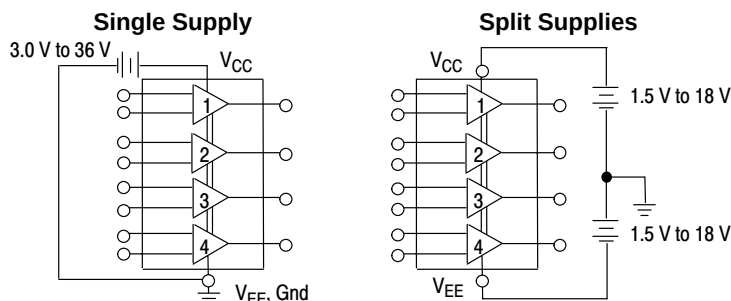


MC3403, MC3303

Single Supply Quad Operational Amplifiers

The MC3403 is a low cost, quad operational amplifier with true differential inputs. The device has electrical characteristics similar to the popular MC1741C. However, the MC3403 has several distinct advantages over standard operational amplifier types in single supply applications. The quad amplifier can operate at supply voltages as low as 3.0 V or as high as 36 V with quiescent currents about one third of those associated with the MC1741C (on a per amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The output voltage range also includes the negative power supply voltage.

- Short Circuit Protected Outputs
- Class AB Output Stage for Minimal Crossover Distortion
- True Differential Input Stage
- Single Supply Operation: 3.0 V to 36 V
- Split Supply Operation: ± 1.5 V to ± 18 V
- Low Input Bias Currents: 500 nA Max
- Four Amplifiers Per Package
- Internally Compensated
- Similar Performance to Popular MC1741C
- Industry Standard Pinouts
- ESD Diodes Added for Increased Ruggedness



MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Power Supply Voltages			
Single Supply	V _{CC}	36	Vdc
Split Supplies	V _{CC} , V _{EE}	± 18	
Input Differential Voltage Range (Note 1)	V _{IDR}	± 36	Vdc
Input Common Mode Voltage Range (Notes 1 and 2)	V _{ICR}	± 18	Vdc
Storage Temperature Range	T _{stg}	-55 to +125	°C
Operating Ambient Temperature Range	T _A		
MC3303		-40 to +85	°C
MC3403		0 to +70	
Junction Temperature	T _J	150	°C

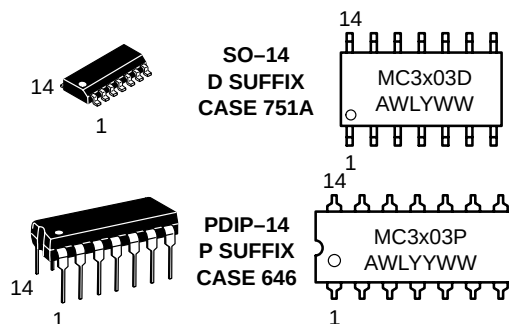
1. Split power supplies.
2. For supply voltages less than ± 18 V, the absolute maximum input voltage is equal to the supply voltage.



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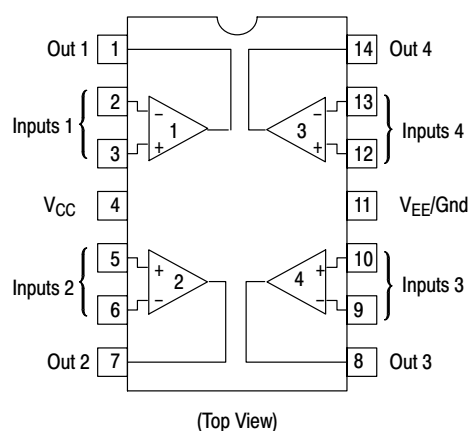
<http://onsemi.com>

MARKING DIAGRAMS



x = 3 or 4
 A = Assembly Location
 WL = Wafer Lot
 YY, Y = Year
 WW = Work Week

PIN CONNECTIONS



ORDERING INFORMATION

Device	Package	Shipping
MC3303D	SO-14	55 Units/Rail
MC3303DR2	SO-14	2500 Tape & Reel
MC3303P	PDIP-14	25 Units/Rail
MC3403D	SO-14	55 Units/Rail
MC3403DR2	SO-14	2500 Tape & Reel
MC3403P	PDIP-14	25 Units/Rail

MC3403, MC3303

ELECTRICAL CHARACTERISTICS ($V_{CC} = +15\text{ V}$, $V_{EE} = -15\text{ V}$ for MC3403; $V_{CC} = +14\text{ V}$, $V_{EE} = \text{Gnd}$ for MC3303 $T_A = 25^\circ\text{C}$, unless otherwise noted.)

Characteristic	Symbol	MC3403			MC3303			Unit
		Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage $T_A = T_{\text{high}}$ to T_{low} (Note 1)	V_{IO}	–	2.0	10	–	2.0	8.0	mV
		–	–	12	–	–	10	
Input Offset Current $T_A = T_{\text{high}}$ to T_{low}	I_{IO}	–	30	50	–	30	75	nA
		–	–	200	–	–	250	
Large Signal Open Loop Voltage Gain $V_O = \pm 10\text{ V}$, $R_L = 2.0\text{ k}\Omega$ $T_A = T_{\text{high}}$ to T_{low}	A_{VOL}	20	200	–	20	200	–	V/mV
		15	–	–	15	–	–	
Input Bias Current $T_A = T_{\text{high}}$ to T_{low}	I_{IB}	–	–200	–500	–	–200	–500	nA
		–	–	–800	–	–	–1000	
Output Impedance $f = 20\text{ Hz}$	z_o	–	75	–	–	75	–	Ω
Input Impedance $f = 20\text{ Hz}$	z_i	0.3	1.0	–	0.3	1.0	–	M Ω
Output Voltage Range $R_L = 10\text{ k}\Omega$ $R_L = 2.0\text{ k}\Omega$ $R_L = 2.0\text{ k}\Omega$, $T_A = T_{\text{high}}$ to T_{low}	V_O	± 12	± 13.5	–	12	12.5	–	V
		± 10	± 13	–	10	12	–	
		± 10	–	–	10	–	–	
Input Common Mode Voltage Range	V_{ICR}	+13 V – V_{EE}	+13 V – V_{EE}	–	+12 V – V_{EE}	+12.5 V – V_{EE}	–	V
Common Mode Rejection $R_S \leq 10\text{ k}\Omega$	CMR	70	90	–	70	90	–	dB
Power Supply Current ($V_O = 0$) $R_L = \infty$	I_{CC} , I_{EE}	–	2.8	7.0	–	2.8	7.0	mA
Individual Output Short-Circuit Current (Note 2)	I_{SC}	± 10	± 20	± 45	± 10	± 30	± 45	mA
Positive Power Supply Rejection Ratio	PSRR+	–	30	150	–	30	150	$\mu\text{V/V}$
Negative Power Supply Rejection Ratio	PSRR–	–	30	150	–	30	150	$\mu\text{V/V}$
Average Temperature Coefficient of Input Offset Current $T_A = T_{\text{high}}$ to T_{low}	$\Delta I_{IO}/\Delta T$	–	50	–	–	50	–	pA/ $^\circ\text{C}$
Average Temperature Coefficient of Input Offset Voltage $T_A = T_{\text{high}}$ to T_{low}	$\Delta V_{IO}/\Delta T$	–	10	–	–	10	–	$\mu\text{V}/^\circ\text{C}$
Power Bandwidth $A_V = 1$, $R_L = 10\text{ k}\Omega$, $V_O = 20\text{ V(p-p)}$, THD = 5%	BWp	–	9.0	–	–	9.0	–	kHz
Small-Signal Bandwidth $A_V = 1$, $R_L = 10\text{ k}\Omega$, $V_O = 50\text{ mV}$	BW	–	1.0	–	–	1.0	–	MHz
Slew Rate $A_V = 1$, $V_i = -10\text{ V}$ to $+10\text{ V}$	SR	–	0.6	–	–	0.6	–	V/ μs
Rise Time $A_V = 1$, $R_L = 10\text{ k}\Omega$, $V_O = 50\text{ mV}$	t_{TLH}	–	0.35	–	–	0.35	–	μs
Fall Time $A_V = 1$, $R_L = 10\text{ k}\Omega$, $V_O = 50\text{ mV}$	t_{TLH}	–	0.35	–	–	0.35	–	μs
Overshoot $A_V = 1$, $R_L = 10\text{ k}\Omega$, $V_O = 50\text{ mV}$	os	–	20	–	–	20	–	%
Phase Margin $A_V = 1$, $R_L = 2.0\text{ k}\Omega$, $V_O = 200\text{ pF}$	ϕ_m	–	60	–	–	60	–	Degrees
Crossover Distortion ($V_{in} = 30\text{ mVpp}$, $V_{out} = 2.0\text{ Vpp}$, $f = 10\text{ kHz}$)	–	–	1.0	–	–	1.0	–	%

- MC3303: $T_{\text{low}} = -40^\circ\text{C}$, $T_{\text{high}} = +85^\circ\text{C}$
MC3403: $T_{\text{low}} = 0^\circ\text{C}$, $T_{\text{high}} = +70^\circ\text{C}$
- Not to exceed maximum package power dissipation.

MC3403, MC3303

ELECTRICAL CHARACTERISTICS ($V_{CC} = 5.0\text{ V}$, $V_{EE} = \text{Gnd}$, $T_A = 25^\circ\text{C}$, unless otherwise noted.)

Characteristic	Symbol	MC3403			MC3303			Unit
		Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	V_{IO}	–	2.0	10	–	–	10	mV
Input Offset Current	I_{IO}	–	30	50	–	–	75	nA
Input Bias Current	I_{IB}	–	–200	–500	–	–	–500	nA
Large Signal Open Loop Voltage Gain $R_L = 2.0\text{ k}\Omega$	A_{VOL}	10	200	–	10	200	–	V/mV
Power Supply Rejection Ratio	PSRR	–	–	150	–	–	150	$\mu\text{V/V}$
Output Voltage Range (Note 3) $R_L = 10\text{ k}\Omega$, $V_{CC} = 5.0\text{ V}$ $R_L = 10\text{ k}\Omega$, $5.0 \leq V_{CC} \leq 30\text{ V}$	V_{OR}	3.3 $V_{CC}-2.0$	3.5 $V_{CC}-1.7$	– –	3.3 $V_{CC}-2.0$	3.5 $V_{CC}-1.7$	– –	Vpp
Power Supply Current	I_{CC}	–	2.5	7.0	–	2.5	7.0	mA
Channel Separation $f = 1.0\text{ kHz to } 20\text{ kHz}$ (Input Referenced)	CS	–	–120	–	–	–120	–	dB

3. Output will swing to ground with a $10\text{ k}\Omega$ pull down resistor.

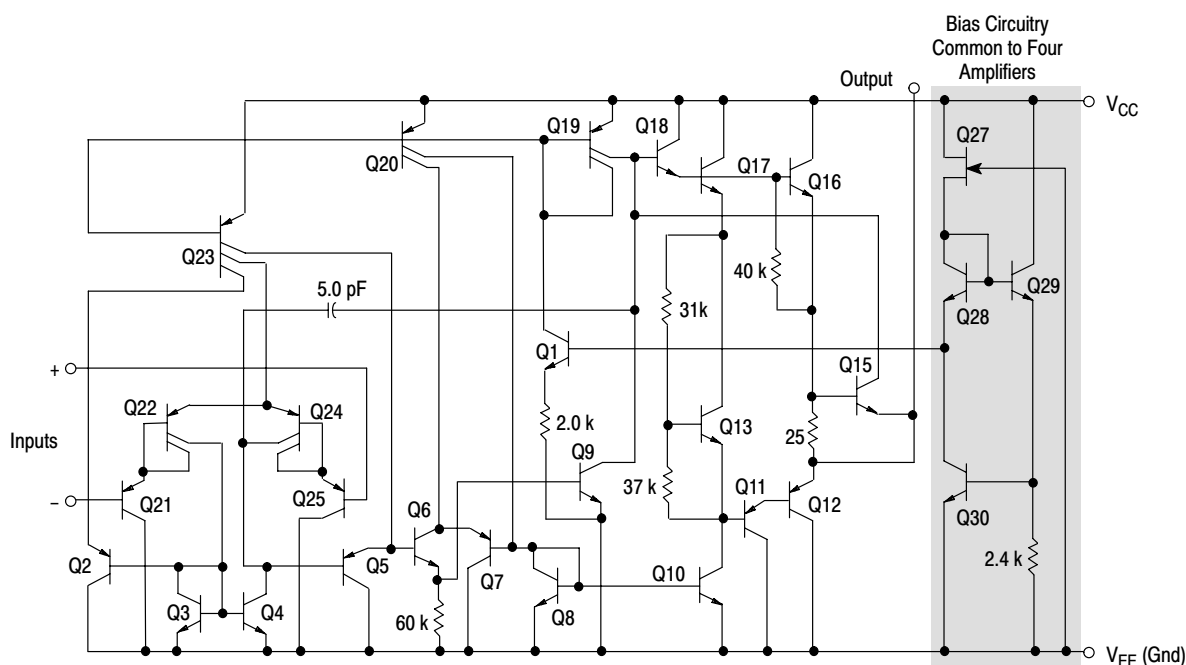


Figure 1. Representative Schematic Diagram
(1/4 of Circuit Shown)

CIRCUIT DESCRIPTION

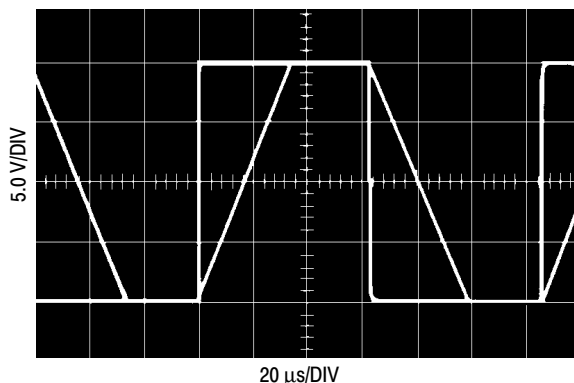


Figure 2. Inverter Pulse Response

The MC3403/3303 is made using four internally compensated, two-stage operational amplifiers. The first stage of each consists of differential input device Q24 and Q22 with input buffer transistors Q25 and Q21 and the differential to single ended converter Q3 and Q4. The first

stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0 pF) can be employed, thus saving chip area. The transconductance reduction is accomplished by splitting the collectors of Q24 and Q22. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single-ended converter. The second stage consists of a standard current source load amplifier stage.

The output stage is unique because it allows the output to swing to ground in single supply operation and yet does not exhibit any crossover distortion in split supply operation. This is possible because Class AB operation is utilized.

Each amplifier is biased from an internal voltage regulator which has a low temperature coefficient, thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.

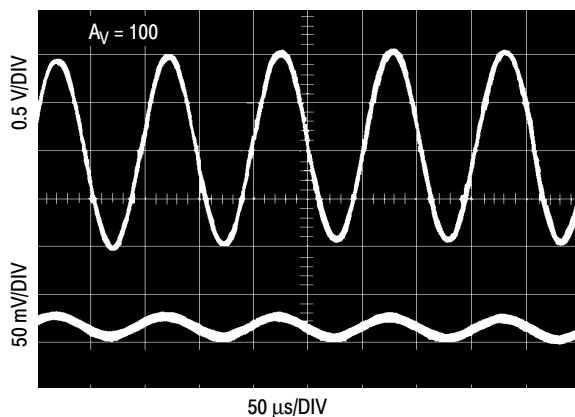


Figure 3. Sine Wave Response

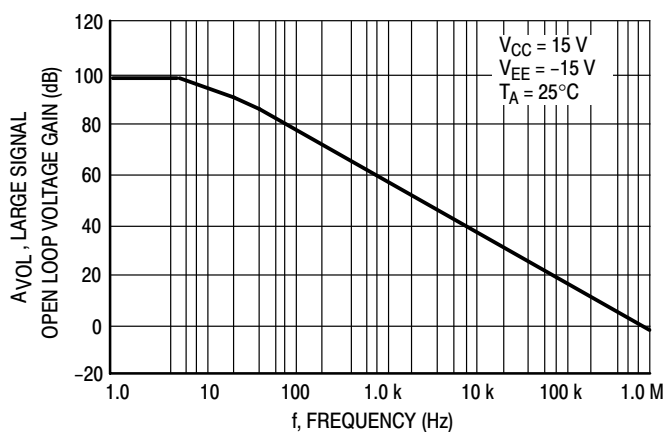


Figure 4. Open Loop Frequency Response

MC3403, MC3303

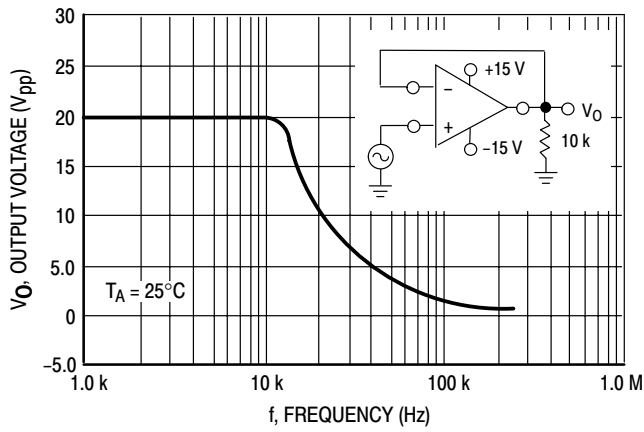


Figure 5. Power Bandwidth

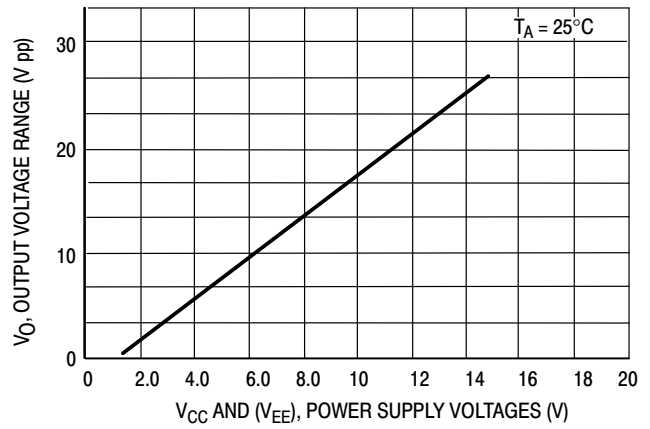


Figure 6. Output Swing versus Supply Voltage

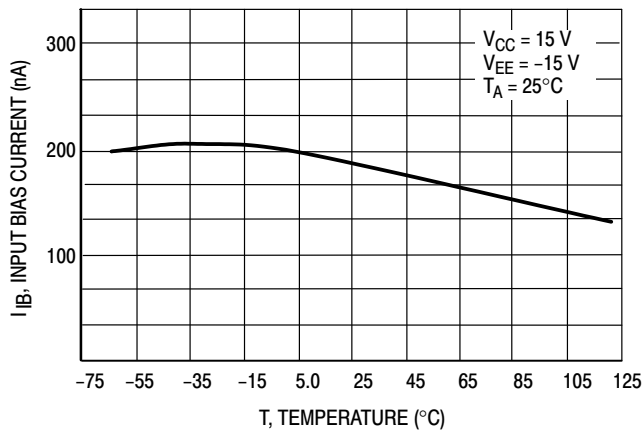


Figure 7. Input Bias Current versus Temperature

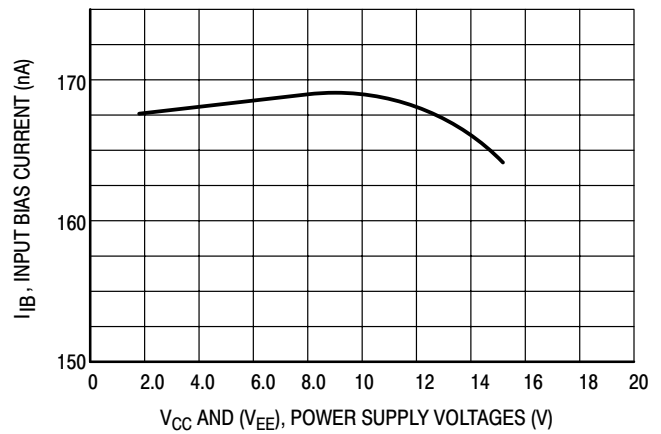


Figure 8. Input Bias Current versus Supply Voltage

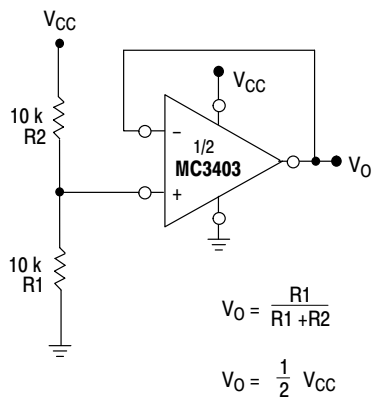


Figure 9. Voltage Reference

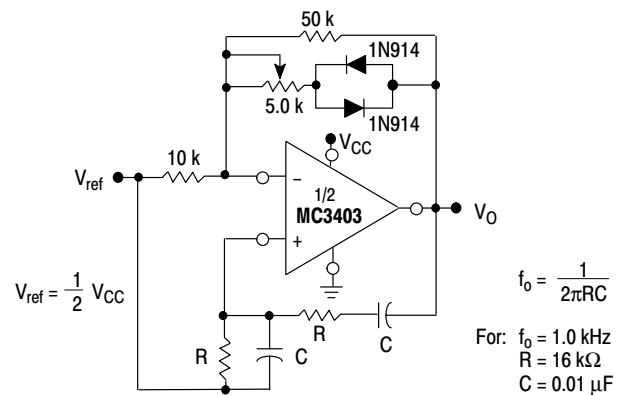


Figure 10. Wien Bridge Oscillator

MC3403, MC3303

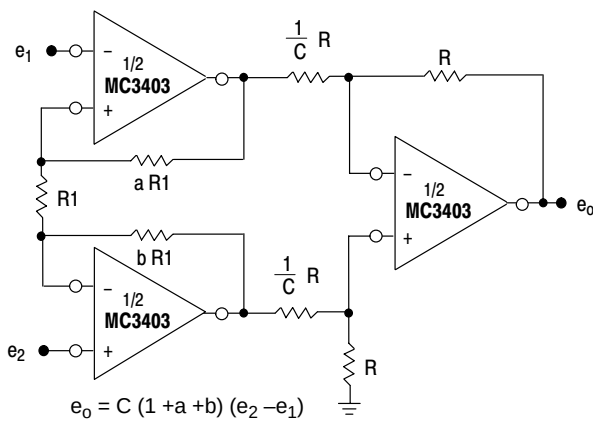


Figure 11. High Impedance Differential Amplifier

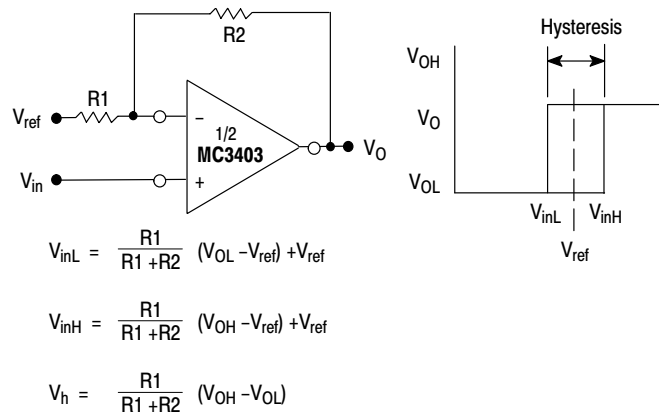


Figure 12. Comparator with Hysteresis

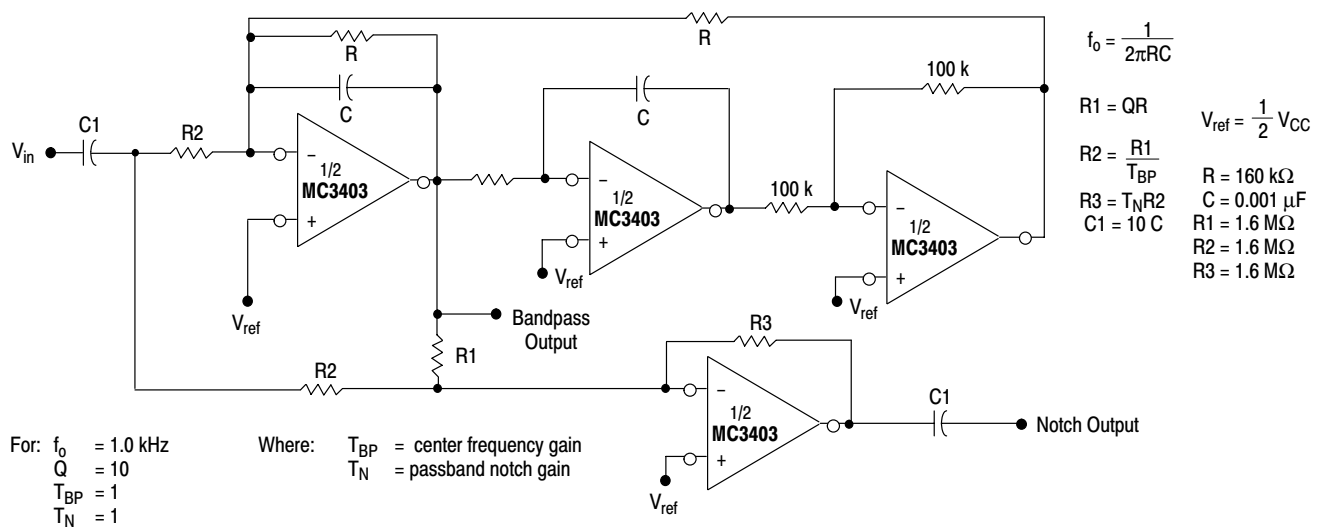


Figure 13. Bi-Quad Filter

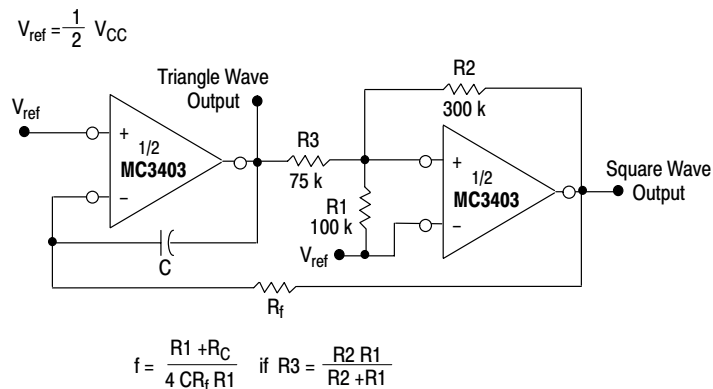
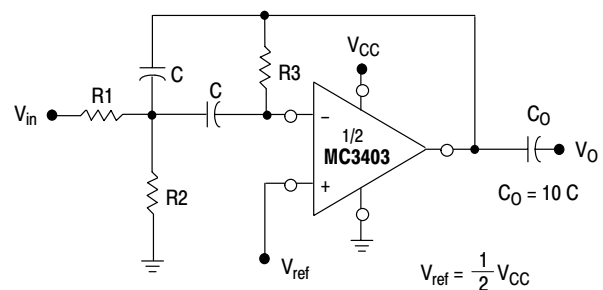


Figure 14. Function Generator



Given: f_0 = center frequency
 $A(f_0)$ = gain at center frequency

Choose value f_0 , C

$$\text{Then: } R_3 = \frac{Q}{\pi f_0 C} \quad R_1 = \frac{R_3}{2 A(f_0)} \quad R_2 = \frac{R_1 R_5}{4 Q^2 R_1 - R_5}$$

For less than 10% error from operational amplifier where f_0 and BW are expressed in Hz.

$$\frac{Q_0 f_0}{BW} < 0.1$$

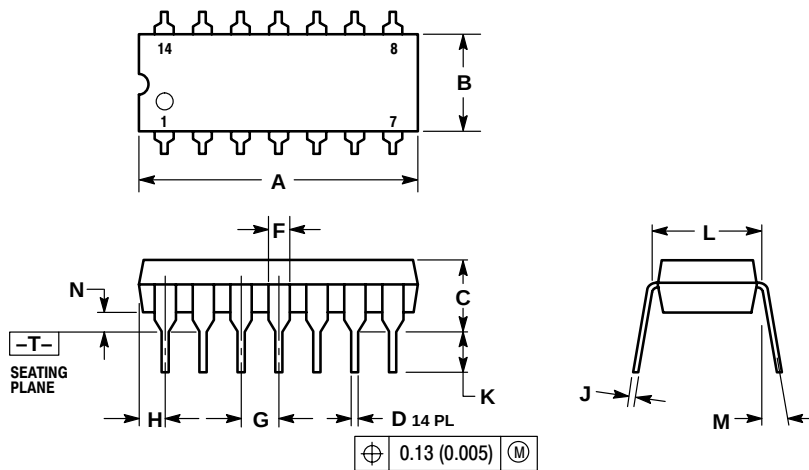
If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

Figure 15. Multiple Feedback Bandpass Filter

MC3403, MC3303

PACKAGE DIMENSIONS

PDIP-14
P SUFFIX
CASE 646-06
ISSUE M

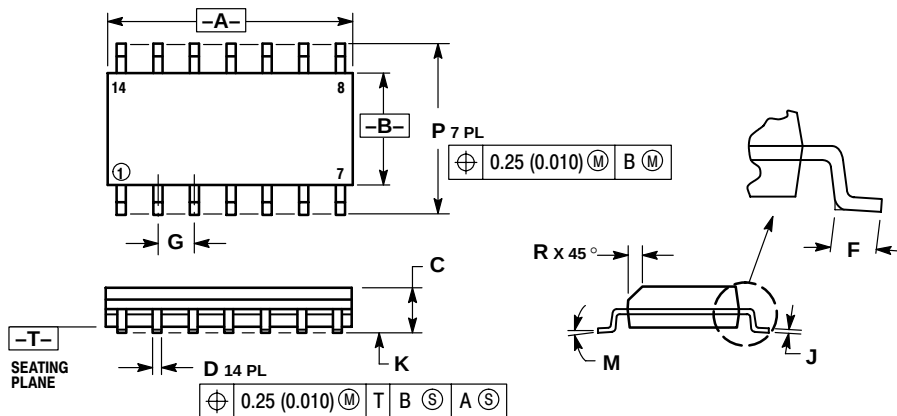


NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
4. DIMENSION B DOES NOT INCLUDE MOLD FLASH.
5. ROUNDED CORNERS OPTIONAL.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.715	0.770	18.16	18.80
B	0.240	0.260	6.10	6.60
C	0.145	0.185	3.69	4.69
D	0.015	0.021	0.38	0.53
F	0.040	0.070	1.02	1.78
G	0.100 BSC		2.54 BSC	
H	0.052	0.095	1.32	2.41
J	0.008	0.015	0.20	0.38
K	0.115	0.135	2.92	3.43
L	0.290	0.310	7.37	7.87
M	---	10°	---	10°
N	0.015	0.039	0.38	1.01

SO-14
D SUFFIX
CASE 751A-03
ISSUE F



NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	8.55	8.75	0.337	0.344
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050 BSC	
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	0°	7°	0°	7°
P	5.80	6.20	0.228	0.244
R	0.25	0.50	0.010	0.019

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