Preferred Device

# Power MOSFET 42 Amps, 60 Volts N-Channel TO-220

This Power MOSFET is designed to withstand high energy in the avalanche and commutation modes. Designed for low voltage, high speed switching applications in power supplies, converters and power motor controls, these devices are particularly well suited for bridge circuits where diode speed and commutating safe operating areas are critical and offer additional safety margin against unexpected voltage transients.

- On-resistance Area Product about One-half that of Standard MOSFETs with New Low Voltage, Low R<sub>DS(on)</sub> Technology
- Faster Switching than E–FET Predecessors
- Avalanche Energy Specified
- I<sub>DSS</sub> and V<sub>DS(on)</sub> Specified at Elevated Temperature
- Static Parameters are the Same for both TMOS V and TMOS E-FET

## **MAXIMUM RATINGS** (T<sub>C</sub> = 25°C unless otherwise noted)

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Rating	Symbol	Value	Unit	
Drain-Source Voltage	V <sub>DSS</sub>	60	Vdc	
Drain–Gate Voltage ( $R_{GS}$ = 1.0 M $\Omega$ )	V <sub>DGR</sub>	60	Vdc	
Gate–Source Voltage – Continuous – Non–Repetitive (t <sub>p</sub> ≤ 10 ms)	V <sub>GS</sub> V <sub>GSM</sub>	± 20 ± 25	Vdc Vpk	
Drain Current – Continuous @ 25°C – Continuous @ 100°C – Single Pulse (t <sub>p</sub> ≤ 10 μs)	I <sub>D</sub> I <sub>D</sub> I <sub>DM</sub>	42 30 147	Adc Apk	
Total Power Dissipation @ 25°C Derate above 25°C	PD	125 0.83	Watts W/°C	
Operating and Storage Temperature Range	T <sub>J</sub> , T <sub>stg</sub>	-55 to 175	°C	
$            Single Pulse Drain-to-Source Avalanche \\             Energy - Starting T_J = 25^\circ C \\              (V_{DD} = 25 Vdc, V_{GS} = 10 Vdc \\              I_L = 42 Apk, L = 0.454 \ \mu H, R_G = 25 \ \Omega )            $	EAS	400	mJ	
Thermal Resistance – Junction to Case – Junction to Ambient	$R_{ heta JC}$ $R_{ heta JA}$	1.2 62.5	°C/W	
Maximum Lead Temperature for Soldering Purposes, 1/8" from case for 10 seconds	ΤL	260	°C	



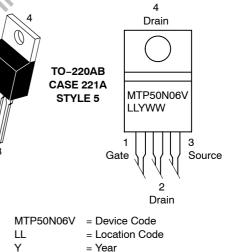
# **ON Semiconductor™**

http://onsemi.com

42 AMPERES 60 VOLTS R<sub>DS(on)</sub> = 28 mΩ

N-Channel

#### MARKING DIAGRAM & PIN ASSIGNMENT



= Work Week

## **ORDERING INFORMATION**

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Device	Package	Shipping
MTP50N06V	TO-220AB	50 Units/Rail

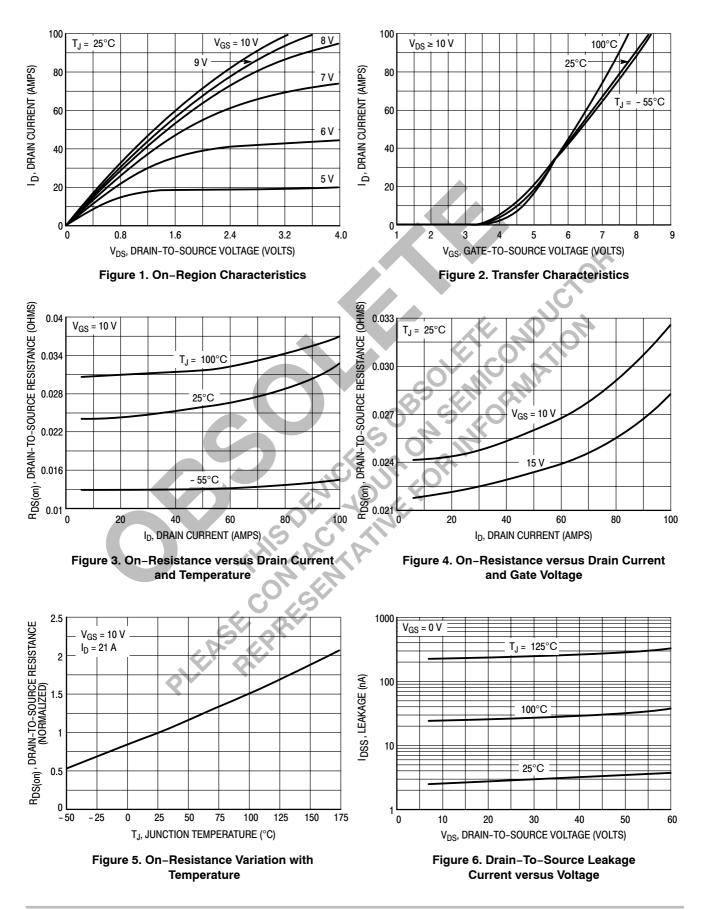
Preferred devices are recommended choices for future use and best overall value.

#### **ELECTRICAL CHARACTERISTICS** ( $T_J = 25^{\circ}C$ unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit	
OFF CHARACTERISTICS		r			1	
Drain-Source Breakdown Voltage (V <sub>GS</sub> = 0 Vdc, I <sub>D</sub> = 250 μAdc) Temperature Coefficient (Positive)	V <sub>(BR)DSS</sub>	60 -	- 69	- -	Vdc mV/°C	
Zero Gate Voltage Drain Current ( $V_{DS} = 60 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}$ ) ( $V_{DS} = 60 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, T_J = 150^{\circ}\text{C}$ )	I <sub>DSS</sub>		-	10 100	μAdc	
Gate-Body Leakage Current ( $V_{GS} = \pm 20$ Vdc, $V_{DS} = 0$ )	I <sub>GSS</sub>	_	-	100	nAdc	
DN CHARACTERISTICS (Note 1)					•	
Gate Threshold Voltage (V <sub>DS</sub> = V <sub>GS</sub> , I <sub>D</sub> = 250 μAdc) Temperature Coefficient (Negative)	V <sub>GS(th)</sub>	2.0	2.7 3.0	4.0 -	Vdc mV/°C	
Static Drain–Source On–Resistance ( $V_{GS}$ = 10 Vdc, $I_D$ = 21 Adc)	R <sub>DS(on)</sub>	-	0.025	0.028	Ohm	
$\label{eq:constraint} \begin{array}{l} \text{Drain-Source On-Voltage (V}_{\text{GS}} = 10 \text{ Vdc}) \\ (I_{\text{D}} = 42 \text{ Adc}) \\ (I_{\text{D}} = 21 \text{ Adc}, \text{ T}_{\text{J}} = 150^{\circ}\text{C}) \end{array}$	V <sub>DS(on)</sub>	-	1.4 -	1.7 1.6	Vdc	
Forward Transconductance (V <sub>DS</sub> = 6.25 Vdc, I <sub>D</sub> = 20 Adc)	9FS	16	23	-	mhos	
DYNAMIC CHARACTERISTICS		4. 3	0	2		
Input Capacitance	C <sub>iss</sub>		1644	2320	pF	
Output Capacitance $(V_{DS} = 25 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, f = 1.0 \text{ MHz})$	C <sub>oss</sub>	G	465	660		
Reverse Transfer Capacitance	C <sub>rss</sub>		112	230	1	
SWITCHING CHARACTERISTICS (Note 2)	6					
Turn-On Delay Time	t <sub>d(on)</sub>		12	20	ns	
Rise Time $(V_{DD} = 25 \text{ Vdc}, I_D = 42 \text{ Adc}, V_{GS} = 10 \text{ Vdc},$	tr	-	122	250		
Turn-Off Delay Time $R_G = 9.1 \Omega$	t <sub>d(off)</sub>	-	64	110		
Fall Time	t <sub>f</sub>	-	54	90		
Gate Charge	Q <sub>T</sub>	-	47	70	nC	
(See Figure 8) (V <sub>DS</sub> = 48 Vdc, I <sub>D</sub> = 42 Adc,	Q <sub>1</sub>	-	9	-		
V <sub>GS</sub> = 10 Vdc)	Q <sub>2</sub>	-	21	_		
	Q <sub>3</sub>	-	16	_		
SOURCE-DRAIN DIODE CHARACTERISTICS						
Forward On–Voltage (Note 1) $ (I_S = 42 \text{ Adc}, V_{GS} = 0 \text{ Vdc}) $ $ (I_S = 42 \text{ Adc}, V_{GS} = 0 \text{ Vdc}, T_J = 150^{\circ}\text{C}) $	V <sub>SD</sub>	-	1.06 0.99	2.5 _	Vdc	
Reverse Recovery Time	t <sub>rr</sub>	-	84	-	ns	
(See Figure 14)	ta	-	73	-		
(I <sub>S</sub> = 42 Adc, V <sub>GS</sub> = 0 Vdc, dI <sub>S</sub> /dt = 100 A/µs)	t <sub>b</sub>	-	11	-		
Reverse Recovery Stored Charge	Q <sub>RR</sub>	-	0.28	-	μC	
NTERNAL PACKAGE INDUCTANCE		1			1	
Internal Drain Inductance (Measured from contact screw on tab to center of die) (Measured from the drain lead 0.25″ from package to center of die)	LD	_	3.5 4.5	-	nH	
Internal Source Inductance (Measured from the source lead 0.25" from package to source bond pad)	L <sub>S</sub>	_	7.5	-	nH	

2. Switching characteristics are independent of operating junction temperature.

## **TYPICAL ELECTRICAL CHARACTERISTICS**



#### POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals ( $\Delta t$ ) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ( $I_{G(AV)}$ ) can be made from a rudimentary analysis of the drive circuit so that

 $t = Q/I_{G(AV)}$ 

During the rise and fall time interval when switching a resistive load,  $V_{GS}$  remains virtually constant at a level known as the plateau voltage,  $V_{SGP}$ . Therefore, rise and fall times may be approximated by the following:

 $t_r = Q_2 x R_G / (V_{GG} - V_{GSP})$ 

 $t_f = Q_2 x R_G / V_{GSP}$ 

where

 $V_{GG}$  = the gate drive voltage, which varies from zero to  $V_{GG}$ 

 $R_G$  = the gate drive resistance

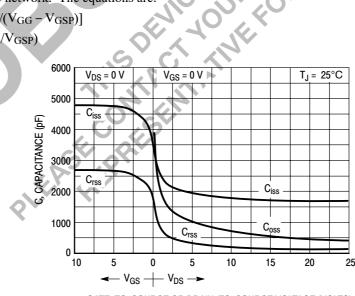
and  $Q_2$  and  $V_{GSP}$  are read from the gate charge curve.

During the turn–on and turn–off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

 $t_{d(on)} = R_G C_{iss} In [V_{GG}/(V_{GG} - V_{GSP})]$  $t_{d(off)} = R_G C_{iss} In (V_{GG}/V_{GSP})$  The capacitance ( $C_{iss}$ ) is read from the capacitance curve at a voltage corresponding to the off–state condition when calculating  $t_{d(on)}$  and is read at a voltage corresponding to the on–state when calculating  $t_{d(off)}$ .

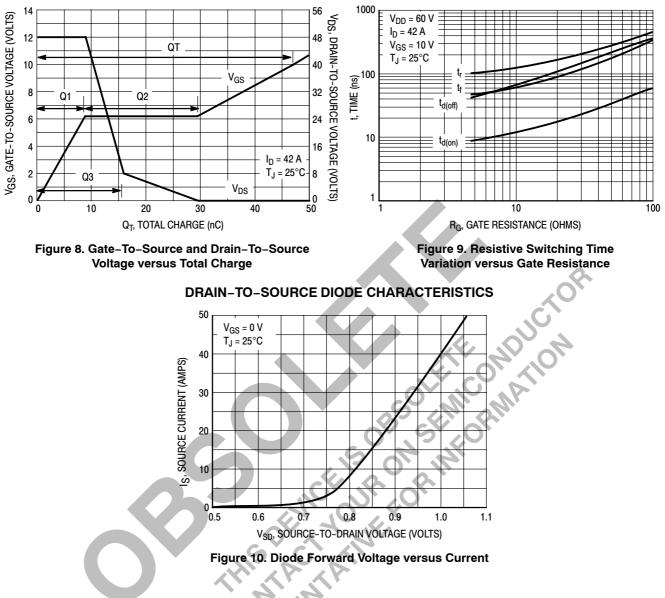
At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.



GATE-TO-SOURCE OR DRAIN-TO-SOURCE VOLTAGE (VOLTS)

Figure 7. Capacitance Variation



## SAFE OPERATING AREA

The Forward Biased Safe Operating Area curves define the maximum simultaneous drain-to-source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature ( $T_C$ ) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance–General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current ( $I_{DM}$ ) nor rated voltage ( $V_{DSS}$ ) is exceeded and the transition time ( $t_r, t_f$ ) do not exceed 10  $\mu$ s. In addition the total power averaged over a complete switching cycle must not exceed ( $T_{J(MAX)} - T_C$ )/( $R_{\theta JC}$ ).

A Power MOSFET designated E–FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E–FETs can withstand the stress of drain–to–source avalanche at currents up to rated pulsed current ( $I_{DM}$ ), the energy rating is specified at rated continuous current ( $I_D$ ), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 12). Maximum energy at currents below rated continuous  $I_D$  can safely be assumed to equal the values indicated.

## SAFE OPERATING AREA

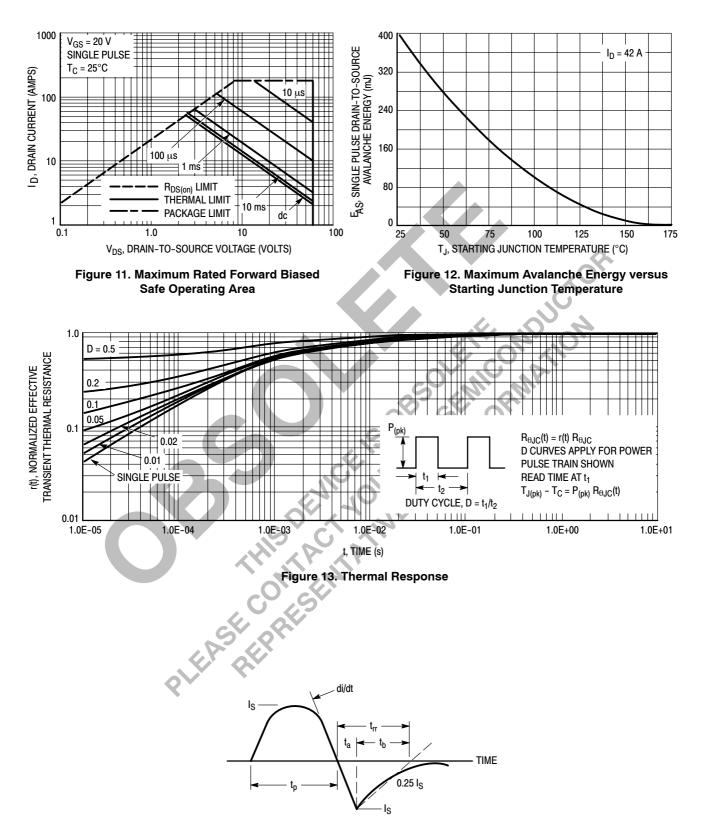
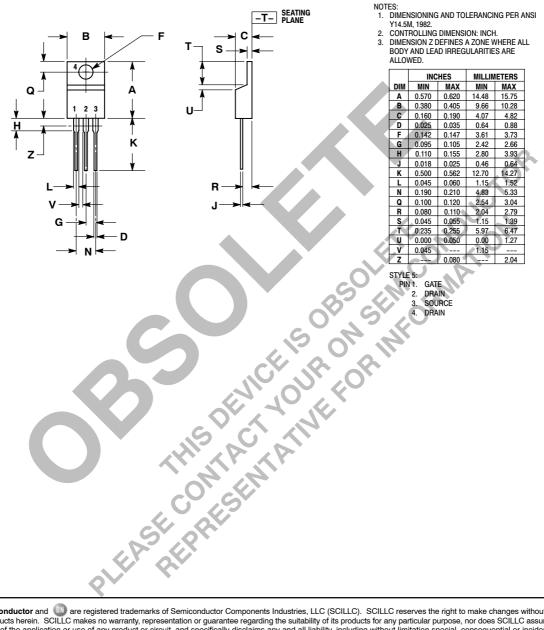


Figure 14. Diode Reverse Recovery Waveform

#### PACKAGE DIMENSIONS

TO-220 THREE-LEAD TO-220AB CASE 221A-09 ISSUE AA



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