## DATA SHEET

## TEA6320 <br> Sound fader control circuit

Preliminary specification
File under Integrated Circuits，IC01

## FEATURES

- Source selector for four stereo and one mono inputs
- Interface for noise reduction circuits
- Interface for external equalizer
- Volume, balance and fader control
- Special loudness characteristic automatically controlled in combination with volume setting
- Bass and treble control
- Mute control at audio signal zero crossing
- Fast mute control via $\mathrm{I}^{2} \mathrm{C}$-bus
- Fast mute control via pin
- ${ }^{2} \mathrm{C}$-bus control for all functions
- Power supply with internal power-on reset.


## QUICK REFERENCE DATA

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{CC}}$ | supply voltage |  | 7.5 | 8.5 | 9.5 | V |
| $\mathrm{I}_{\mathrm{CC}}$ | supply current | $\mathrm{V}_{\mathrm{CC}}=8.5 \mathrm{~V}$ | - | 26 | - | mA |
| $\mathrm{V}_{\mathrm{o}(\mathrm{rms})}$ | maximum output voltage level | $\mathrm{V}_{\mathrm{CC}}=8.5 \mathrm{~V} ; \mathrm{THD} \leq 0.1 \%$ | - | 2000 | - | mV |
| $\mathrm{G}_{\mathrm{v}}$ | voltage gain |  | -86 | - | +20 | dB |
| $\mathrm{G}_{\text {step(vol) }}$ | step resolution (volume) |  | - | 1 | - | dB |
| $\mathrm{G}_{\text {bass }}$ | bass control |  | -15 | - | +15 | dB |
| $\mathrm{G}_{\text {treble }}$ | treble control |  | -12 | - | +12 | dB |
| $\mathrm{G}_{\text {step(treble })}$ | step resolution (bass, treble) |  | - | 1.5 | - | dB |
| $(\mathrm{S}+\mathrm{N}) / \mathrm{N}$ | signal-plus-noise to noise ratio | $\mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V} ; \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB} ;$ <br> unweighted | - | 105 | - | dB |
| $\mathrm{RR}_{100}$ | ripple rejection | $\mathrm{V}_{\mathrm{r}(\mathrm{rms})}<200 \mathrm{mV} ; \mathrm{f}=100 \mathrm{~Hz} ;$ <br> $\mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB}$ | - | 76 | - | dB |
| $\alpha_{\mathrm{cs}}$ | channel separation | $250 \mathrm{~Hz} \leq \mathrm{f} \leq 10 \mathrm{kHz;} \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB}$ | 90 | 96 | - | dB |

ORDERING INFORMATION

| TYPE <br> NUMBER | PACKAGE |  |  |
| :--- | :---: | :--- | :---: |
|  | NAME | DESCRIPTION | VERSION |
| TEA6320 | SDIP32 | plastic shrink dual in-line package; 32 leads (400 mil) | SOT232-1 |
| TEA6320T | SO32 | plastic small outline package; 32 leads; body width 7.5 mm | SOT287-1 |

## BLOCK DIAGRAM



Fig. 1 Block diagram.

PINNING

| SYMBOL | PIN | DESCRIPTION |
| :--- | :---: | :--- |
| SDA | 1 | serial data input/output |
| GND | 2 | ground |
| OUTLR | 3 | output left rear |
| OUTLF | 4 | output left front |
| TL | 5 | treble control capacitor left channel or <br> input from an external equalizer |
| B2L | 6 | bass control capacitor left channel or <br> output to an external equalizer |
| B1L | 7 | bass control capacitor, left channel |
| IVL | 8 | input volume I, left control part |
| ILL | 9 | input loudness, left control part |
| QSL | 10 | output source selector, left channel |
| IDL | 11 | input D left source |
| MUTE | 12 | mute control |
| ICL | 13 | input C left source |
| IMO | 14 | input mono source |
| IBL | 15 | input B left source |
| IAL | 16 | input A left source |
| IAR | 17 | input A right source |
| IBR | 18 | input B right source |
| CAP | 19 | electronic filtering for supply |
| ICR | 20 | input C right source |
| $V_{\text {ref }}$ | 21 | reference voltage (0.5V CC) |
| IDR | 22 | input D right source |
| QSR | 23 | output source selector right channel |
| ILR | 24 | input loudness right channel |
| IVR | 25 | input volume I, right control part |
| B1R | 26 | bass control capacitor right channel |
| B2R | 27 | bass control capacitor right channel or <br> output to an external equalizer |
| VCC | 31 | supply voltage |
| SCL | 32 | serial clock input |
| OUTRF | 29 | \begin{tabular}{l}
\end{tabular} |
| oUTRR |  |  |
| or input from an external equalizer |  |  |$|$



Fig. 2 Pin configuration.

## FUNCTIONAL DESCRIPTION

The source selector selects one of 4 stereo inputs or the mono input. The maximum input signal voltage is $\mathrm{V}_{\mathrm{i}(\mathrm{ms})}=2 \mathrm{~V}$. The outputs of the source selector and the inputs of the following volume control parts are available at pins 8 and 10 for the left channel and pins 23 and 25 for the right channel. This offers the possibility of interfacing a noise reduction system.

The volume control function is split into two sections: volume I control block and volume II control block.

The control range of volume $I$ is between +20 dB and -31 dB in steps of 1 dB . The volume II control range is between 0 dB and -55 dB in steps of 1 dB . Although the theoretical possible control range is 106 dB
(+20 to -86 dB), in practice a range of $86 \mathrm{~dB}(+20$ to -66 dB ) is recommended. The gain/attenuation setting of the volume I control block is common for both channels.

The volume I control block operates in combination with the loudness control. The filter is linear when the maximum gain for the volume I control ( +20 dB ) is selected. The filter characteristic increases automatically over a range of 32 dB down to a setting of -12 dB . That means the maximum filter characteristic is obtained at -12 dB setting of volume I. Further reduction of the volume does not further influence the filter characteristic (see Fig.5). The maximum selected filter characteristic is determined by external components. The proposed application gives a maximum boost of 17 dB for bass and 4.5 dB for treble. The loudness may be switched on or off via $1^{2} \mathrm{C}$-bus control (see Table 7).

The volume I control block is followed by the bass control block. A single external capacitor of 33 nF for each channel in combination with internal resistors, provides the frequency response of the bass control (see Fig.3). The adjustable range is between -15 and +15 dB at 40 Hz .

Both loudness and bass control result in a maximum bass boost of 32 dB for low volume settings.
The treble control block offers a control range between -12 and +12 dB in steps of 1.5 dB at 15 kHz . The filter characteristic is determined by a single capacitor of 5.6 nF for each channel in combination with internal resistors (see Fig.4).

The basic step width of bass and treble control is 3 dB . The intermediate steps are obtained by switching 1.5 dB boost and 1.5 dB attenuation steps.
The bass and treble control functions can be switched off via $\mathrm{I}^{2} \mathrm{C}$-bus. In this event the internal signal flow is disconnected. The connections B2L and B2R are outputs
and TL and TR are inputs for inserting an external equalizer.

The last section of the circuit is the volume II block. The balance and fader functions are performed using the same control blocks. This is realized by 4 independently controllable attenuators, one for each output. The control range of these attenuators is 55 dB in steps of 1 dB with an additional mute step.

The circuit provides 3 mute modes:

1. Zero crossing mode mute via $\mathrm{I}^{2} \mathrm{C}$-bus using 2 independent zero crossing detectors (ZCM, see Tables 2 and 9 and Fig.16).
2. Fast mute via MUTE pin (see Fig.10).
3. Fast mute via $\mathrm{I}^{2} \mathrm{C}$-bus either by general mute (GMU, see Tables 2 and 9 ) or volume II block setting (see Table 4).
The mute function is performed immediately if ZCM is cleared $(Z C M=0)$. If the bit is set $(Z C M=1)$ the mute is activated after changing the GMU bit. The actual mute switching is delayed until the next zero crossing of the audio frequency signal. As the two audio channels (left and right) are independent, two comparators are built-in to control independent mute switches.

To avoid a large delay of mute switching when very low frequencies are processed, the maximum delay time is limited to typically 100 ms by an integrated timing circuit and an external capacitor ( $\mathrm{C}_{\mathrm{m}}=10 \mathrm{nF}$, see Fig.10). This timing circuit is triggered by reception of a new data word for the switch function which includes the GMU bit. After a discharge and charge period of an external capacitor the muting switch follows the GMU bit if no zero crossing was detected during that time.

The mute function can also be controlled externally. If the mute pin is switched to ground all outputs are muted immediately (hardware mute). This mute request overwrites all mute controls via the $\mathrm{I}^{2} \mathrm{C}$-bus for the time the pin is held LOW. The hardware mute position is not stored in the TEA6320.

For the turn on/off behaviour the following explanation is generally valid. To avoid AF output caused by the input signal coming from preceding stages, which produces output during drop of $\mathrm{V}_{\mathrm{Cc}}$, the mute has to be set, before the $\mathrm{V}_{\mathrm{Cc}}$ will drop. This can be achieved by $\mathrm{I}^{2} \mathrm{C}$-bus control or by grounding the MUTE pin.
For use where is no mute in the application before turn off, a supply voltage drop of more than $1 \times \mathrm{V}_{\mathrm{BE}}$ will result in a mute during the voltage drop.

The power supply should include a $\mathrm{V}_{\mathrm{CC}}$ buffer capacitor, which provides a discharging time constant. If the input signal does not disappear after turn off the input will become audible after certain time. A $4.7 \mathrm{k} \Omega$ resistor discharges the $\mathrm{V}_{\mathrm{CC}}$ buffer capacitor, because the internal current of the IC does not discharge it completely.

The hardware mute function is favourable for use in Radio Data System (RDS) applications. The zero crossing mute avoids modulation plops. This feature is an advantage for mute during changing presets and/or sources (e.g. traffic announcement during cassette playback).

## LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

| SYMBOL | PARAMETER | CONDITIONS | MIN. | MAX. | UNIT |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{CC}}$ | supply voltage |  | 0 | 10 | V |
| $\mathrm{~V}_{\mathrm{n}}$ | voltage at all pins except pin 2 <br> referenced to GND (pin 2) |  | 0 | $\mathrm{~V}_{\mathrm{CC}}$ | V |
| $\mathrm{T}_{\mathrm{amb}}$ | operating ambient temperature |  | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{stg}}$ | storage temperature |  | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{\mathrm{es}}$ | electrostatic handling | note 1 |  |  |  |

## Note

1. Human body model: $\mathrm{C}=100 \mathrm{pF} ; \mathrm{R}=1.5 \mathrm{k} \Omega ; \mathrm{V} \geq 2 \mathrm{kV}$. Charge device model: $\mathrm{C}=200 \mathrm{pF} ; \mathrm{R}=0 \Omega ; \mathrm{V} \geq 500 \mathrm{~V}$.

## CHARACTERISTICS

$\mathrm{V}_{\mathrm{CC}}=8.5 \mathrm{~V} ; \mathrm{R}_{\mathrm{S}}=600 \Omega ; \mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega ; \mathrm{C}_{\mathrm{L}}=2.5 \mathrm{nF} ; \mathrm{AC}$ coupled; $\mathrm{f}=1 \mathrm{kHz} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$; gain control $\mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB}$; bass linear; treble linear; fader off; balance in mid position; loudness off; unless otherwise specified.

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | supply voltage |  | 7.5 | 8.5 | 9.5 | V |
| $\mathrm{I}_{\text {CC }}$ | supply current |  | - | 26 | 33 | mA |
| $\mathrm{V}_{\mathrm{DC}}$ | internal DC voltage at inputs and outputs |  | 3.83 | 4.25 | 4.68 | V |
| $\mathrm{V}_{\text {ref }}$ | internal reference voltage at pin 21 |  | - | 4.25 | - | V |
| $\mathrm{G}_{\mathrm{V}(\text { max })}$ | maximum voltage gain | $\mathrm{R}_{\mathrm{S}}=0 \Omega ; \mathrm{R}_{\mathrm{L}}=\infty$ | 19 | 20 | 21 | dB |
| $\mathrm{V}_{\text {orms }}$ | output voltage level for $P_{\text {max }}$ at the power output stage start of clipping | THD $\leq 0.5 \%$; see Fig. 11 | - | 2000 | - | mV |
|  |  | THD $=1 \%$ | 2300 | - | - | mV |
|  |  | $\begin{aligned} & \mathrm{R}_{\mathrm{L}}=2 \mathrm{k} \Omega ; \mathrm{C}_{\mathrm{L}}=10 \mathrm{nF} ; \\ & \mathrm{THD}=1 \% \end{aligned}$ | 2000 | - | - | mV |
| $\mathrm{V}_{\mathrm{i} \text { (rms) }}$ | input sensitivity | $\mathrm{V}_{0}=2000 \mathrm{mV} ; \mathrm{G}_{\mathrm{v}}=20 \mathrm{~dB}$ | - | 200 | - | mV |
| $\mathrm{fr}_{\text {ro }}$ | roll-off frequency | $\begin{aligned} & \hline \mathrm{C}_{\text {KIN }}=220 \mathrm{nF} ; \\ & \mathrm{C}_{\mathrm{KVL}}=220 \mathrm{nF} ; \mathrm{Z}_{\mathrm{i}}=\mathrm{Z}_{\mathrm{i}(\mathrm{~min})} \\ & \text { low frequency }(-1 \mathrm{~dB}) \\ & \text { low frequency }(-3 \mathrm{~dB}) \\ & \text { high frequency }(-1 \mathrm{~dB}) \end{aligned}$ | $\begin{array}{\|l\|} \hline 60 \\ 30 \\ 20000 \end{array}$ |  | $\left.\right\|_{-} ^{-}$ | $\begin{aligned} & \mathrm{Hz} \\ & \mathrm{~Hz} \\ & \mathrm{~Hz} \end{aligned}$ |
|  |  | $\begin{aligned} & \mathrm{C}_{\text {KIN }}=470 \mathrm{nF} ; \\ & \mathrm{C}_{\text {KVL }}=100 \mathrm{nF} ; \mathrm{Z}_{\mathrm{i}}=\mathrm{Z}_{\mathrm{i}} \text { (typ) } \\ & \text { low frequency }(-3 \mathrm{~dB}) \end{aligned}$ | 17 | - | - | Hz |
| $\alpha_{\text {cs }}$ | channel separation | $\mathrm{V}_{\mathrm{i}}=2 \mathrm{~V}$; frequency range 250 Hz to 10 kHz | 90 | 96 | - | dB |
| THD | total harmonic distortion | frequency range <br> 20 Hz to 12.5 kHz $\begin{aligned} & \mathrm{V}_{\mathrm{i}}=100 \mathrm{mV} ; \mathrm{G}_{\mathrm{v}}=20 \mathrm{~dB} \\ & \mathrm{~V}_{\mathrm{i}}=1 \mathrm{~V} ; \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB} \\ & \mathrm{~V}_{\mathrm{i}}=2 \mathrm{~V} ; \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB} \\ & \mathrm{~V}_{\mathrm{i}}=2 \mathrm{~V} ; \mathrm{G}_{\mathrm{v}}=-10 \mathrm{~dB} \end{aligned}$ |  | $\begin{array}{\|l\|} \hline 0.1 \\ 0.05 \\ 0.1 \\ 0.1 \\ \hline \end{array}$ | $\left\lvert\, \begin{array}{\|l} - \\ 0.15 \\ - \\ - \end{array}\right.$ | $\begin{array}{\|l} \hline \% \\ \% \\ \% \\ \% \\ \% \\ \hline \end{array}$ |
| RR | ripple rejection | $\mathrm{V}_{\mathrm{r}(\mathrm{mss})}<200 \mathrm{mV}$ |  |  |  |  |
|  |  | $\mathrm{f}=100 \mathrm{~Hz}$ | 70 | 76 | - | dB |
|  |  | $\mathrm{f}=40 \mathrm{~Hz}$ to 12.5 kHz | - | 66 | - | dB |
| (S+N)/N | signal-plus-noise to noise ratio | unweighted; <br> 20 Hz to 20 kHz (RMS); <br> $\mathrm{V}_{0}=2.0 \mathrm{~V}$; see Figs 6 and 7 | - | 105 | - | dB |
|  |  | CCIR468-2 weighted; quasi peak; $\mathrm{V}_{0}=2.0 \mathrm{~V}$ $\begin{aligned} & \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB} \\ & \mathrm{G}_{\mathrm{v}}=12 \mathrm{~dB} \\ & \mathrm{G}_{\mathrm{v}}=20 \mathrm{~dB} \end{aligned}$ | - | $\begin{array}{\|l} 95 \\ 88 \\ 81 \end{array}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \end{aligned}$ |

Sound fader control circuit

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{P}_{\mathrm{no} \text { (rms) }}$ | noise output power (RMS value) only contribution of TEA6320; power amplifier for 6 W | mute position; note 1 | - | - | 10 | nW |
| $\alpha_{c t}$ | crosstalk $\left(20 \log \frac{V_{\text {bus }(p-p)}}{V_{o(r m s)}}\right)$ between bus inputs and signal outputs | note 2 | - | 110 | - | dB |
| Source selector |  |  |  |  |  |  |
| $\mathrm{Z}_{i}$ | input impedance |  | 25 | 35 | 45 | $\mathrm{k} \Omega$ |
| $\alpha_{S}$ | input isolation of one selected source to any other input | $\mathrm{f}=1 \mathrm{kHz}$ | - | 105 | - | dB |
|  |  | $\mathrm{f}=12.5 \mathrm{kHz}$ | - | 95 | - | dB |
| $\mathrm{V}_{\mathrm{i} \text { (rms) }}$ | maximum input voltage (RMS value) | THD < 0.5\%; $\mathrm{V}_{\mathrm{CC}}=8.5 \mathrm{~V}$ | - | 2.15 | - | V |
|  |  | THD < 0.5\%; $\mathrm{V}_{\mathrm{CC}}=7.5 \mathrm{~V}$ | - | 1.8 | - | V |
| $\mathrm{V}_{\text {offset }}$ | DC offset voltage at source selector output by selection of any inputs |  | - | - | 10 | mV |
| $\mathrm{Z}_{0}$ | output impedance |  | - | 80 | 120 | $\Omega$ |
| $\mathrm{R}_{\mathrm{L}}$ | output load resistance |  | 10 | - | - | $\mathrm{k} \Omega$ |
| $\mathrm{C}_{\mathrm{L}}$ | output load capacity |  | 0 | - | 2500 | pF |
| $\mathrm{G}_{v}$ | voltage gain, source selector |  | - | 0 | - | dB |

Control part (source selector disconnected; source resistance $600 \Omega$ )

| $\mathrm{Z}_{\mathrm{i}}$ | input impedance volume input |  | 100 | 150 | 200 | $\mathrm{k} \Omega$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | input impedance loudness input |  | 25 | 33 | 40 | $\mathrm{k} \Omega$ |
| $\mathrm{Z}_{0}$ | output impedance |  | - | 80 | 120 | $\Omega$ |
| $\mathrm{R}_{\mathrm{L}}$ | output load resistance |  | 2 | - | - | $\mathrm{k} \Omega$ |
| $\mathrm{C}_{\mathrm{L}}$ | output load capacity |  | 0 | - | 10 | nF |
| $\mathrm{R}_{\mathrm{DCL}}$ | DC load resistance at output to ground |  | 4.7 | - | - | $\mathrm{k} \Omega$ |
| $\mathrm{V}_{\mathrm{i}(\mathrm{rms})}$ | maximum input voltage (RMS value) | THD < 0.5\% | - | 2.15 | - | V |
| $\mathrm{V}_{\mathrm{no}}$ | noise output voltage | CCIR468-2 weighted; quasi peak $\begin{aligned} & \mathrm{G}_{\mathrm{v}}=20 \mathrm{~dB} \\ & \mathrm{G}_{\mathrm{v}}=0 \mathrm{~dB} \\ & \mathrm{G}_{\mathrm{v}}=-66 \mathrm{~dB} \end{aligned}$ |  | $\begin{array}{\|l\|} \hline 110 \\ 33 \\ 13 \end{array}$ | $\begin{array}{\|l} 220 \\ 50 \\ 22 \\ \hline \end{array}$ | $\begin{aligned} & \mu \mathrm{V} \\ & \mu \mathrm{~V} \\ & \mu \mathrm{~V} \end{aligned}$ |
|  |  | mute position | - | 10 | - | $\mu \mathrm{V}$ |
| $\mathrm{CR}_{\text {tot }}$ | total continuous control range |  | - | 106 | - | dB |
|  | recommended control range |  | - | 86 | - | dB |
| $\mathrm{G}_{\text {step }}$ | step resolution |  | - | 1 | - | dB |
|  | step error between any adjoining step |  | - | - | 0.5 | dB |


| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\Delta \mathrm{G}_{\mathrm{a}}$ | attenuator set error | $\mathrm{G}_{\mathrm{v}}=+20$ to -50 dB | - | - | 2 | dB |
|  |  | $\mathrm{G}_{\mathrm{v}}=-51$ to -66 dB | - | - | 3 | dB |
| $\Delta \mathrm{G}_{\mathrm{t}}$ | gain tracking error | $\mathrm{G}_{\mathrm{v}}=+20$ to -50 dB | - | - | 2 | dB |
| MUTE $_{\text {att }}$ | mute attenuation | see Fig. 10 | 100 | 110 | - | dB |
| $\mathrm{V}_{\text {offset }}$ | DC step offset between any adjoining step | $\mathrm{G}_{\mathrm{v}}=0$ to -66 dB | - | 0.2 | 10 | mV |
|  |  | $\mathrm{G}_{\mathrm{v}}=20$ to 0 dB | - | 2 | 15 | mV |
|  | DC step offset between any step to mute | $\mathrm{G}_{\mathrm{v}}=0$ to -66 dB | - | - | 10 | mV |
| Volume I control and loudness |  |  |  |  |  |  |
| $\mathrm{CR}_{\text {vol }}$ | continuous volume control range |  | - | 51 | - | dB |
| $\mathrm{G}_{\mathrm{v}}$ | voltage gain |  | -31 | - | +20 | dB |
| $\mathrm{G}_{\text {step }}$ | step resolution |  | - | 1 | - | dB |
| $L_{\text {Bmax }}$ | maximum loudness boost | loudness on; referred to loudness off; boost is determined by external components $\begin{aligned} & f=40 \mathrm{~Hz} \\ & \mathrm{f}=10 \mathrm{kHz} \end{aligned}$ | $\mid-$ | $\begin{aligned} & 17 \\ & 4.5 \end{aligned}$ | - | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ |

## Bass control

| $G_{\text {bass }}$ | bass control, maximum boost | $\mathrm{f}=40 \mathrm{~Hz}$ | 14 | 15 | 16 | dB |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | maximum attenuation | $\mathrm{f}=40 \mathrm{~Hz}$ | 14 | 15 | 16 | dB |
| $\mathrm{G}_{\text {step }}$ | step resolution (toggle switching) | $\mathrm{f}=40 \mathrm{~Hz}$ | - | 1.5 | - | dB |
|  | step error between any adjoining <br>  | $\mathrm{f}=40 \mathrm{~Hz}$ | - | - | 0.5 | dB |
|  | DC step offset in any bass position |  | - | - | 20 | mV |

## Treble control

| $\mathrm{G}_{\text {treble }}$ | treble control, maximum boost | $\mathrm{f}=15 \mathrm{kHz}$ | 11 | 12 | 13 | dB |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | maximum attenuation | $\mathrm{f}=15 \mathrm{kHz}$ | 11 | 12 | 13 | dB |
|  | maximum boost | $\mathrm{f}>15 \mathrm{kHz}$ | - | - | 15 | dB |
| $\mathrm{G}_{\text {step }}$ | step resolution (toggle switching) | $\mathrm{f}=15 \mathrm{kHz}$ | - | 1.5 | - | dB |
|  | step error between any adjoining <br> step | $\mathrm{f}=15 \mathrm{kHz}$ | - | - | 0.5 | dB |
|  | DC step offset in any treble <br> position |  | - | - | 10 | mV |

Volume II, balance and fader control

| CR | continuous attenuation fader and <br> volume control range |  | 53.5 | 55 | 56.5 | dB |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{G}_{\text {step }}$ | step resolution |  | - | 1 | 2 | dB |
|  | attenuation set error |  | - | - | 1.5 | dB |


| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mute function (see Fig.10) |  |  |  |  |  |  |
| Hardware mute |  |  |  |  |  |  |
| $\mathrm{V}_{\text {sw }}$ | mute switch level ( $2 \times \mathrm{V}_{\mathrm{BE}}$ ) |  | - | 1.45 | - | V |
| mute active |  |  |  |  |  |  |
| $\mathrm{V}_{\text {swLow }}$ | input level |  | - | - | 1.0 | V |
| $\mathrm{I}_{\mathrm{i}}$ | input current | $\mathrm{V}_{\text {swLow }}=1 \mathrm{~V}$ | -300 | - | - | $\mu \mathrm{A}$ |
| mute passive: level internally defined |  |  |  |  |  |  |
| $\mathrm{V}_{\text {swHIGH }}$ | saturation voltage |  | - | - | $\mathrm{V}_{\text {CC }}$ | V |
| $\mathrm{t}_{\text {d(mute) }}$ | delay until mute passive |  | - | - | 0.5 | ms |
| Zero crossing mute |  |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{d}}$ | discharge current |  | 0.3 | 0.6 | 1.2 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {ch }}$ | charge current |  | -300 | -150 | - | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {swDEL }}$ | delay switch level ( $3 \times \mathrm{V}_{\mathrm{BE}}$ ) |  | - | 2.2 | - | V |
| $\mathrm{t}_{\mathrm{d}}$ | delay time | $\mathrm{C}_{\mathrm{m}}=10 \mathrm{nF}$ | - | 100 | - | ms |
| $\mathrm{V}_{\text {wind }}$ | window for audio signal zero crossing detection |  | - | 30 | 40 | mV |
| Muting at power supply drop |  |  |  |  |  |  |
| $\mathrm{V}_{\text {CCdrop }}$ | supply drop for mute active |  | - | $\begin{aligned} & \hline \mathrm{V}_{19}- \\ & 0.7 \end{aligned}$ | - | V |
| Power-on reset (when reset is active the GMU-bit (general mute) is set and the $I^{2} \mathrm{C}$-bus receiver is in reset position) |  |  |  |  |  |  |
| $\mathrm{V}_{\text {cc }}$ | increasing supply voltage start of reset |  | - | - | 2.5 | V |
|  | end of reset |  | 5.2 | 6.5 | 7.2 | V |
|  | decreasing supply voltage start of reset |  | 4.2 | 5.5 | 6.2 | V |
| Digital part ( ${ }^{2} \mathbf{C}$-bus pins); note 3 |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{iH}}$ | HIGH level input voltage |  | 3 | - | 9.5 | V |
| $\mathrm{V}_{\mathrm{iL}}$ | LOW level input voltage |  | -0.3 | - | +1.5 | V |
| $\mathrm{l}_{\mathrm{iH}}$ | HIGH level input current |  | -10 | - | +10 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {iL }}$ | LOW level input current |  | -10 | - | +10 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {oL }}$ | LOW level output voltage | $\mathrm{I}_{\mathrm{L}}=3 \mathrm{~mA}$ | - | - | 0.4 | V |

## Notes to the characteristics

1. The indicated values for output power assume a 6 W power amplifier at $4 \Omega$ with 20 dB gain and a fixed attenuator of 12 dB in front of it. Signal-to-noise ratios exclude noise contribution of the power amplifier.
2. The transmission contains: total initialization with MAD and subaddress for volume and 8 data words, see also definition of characteristics, clock frequency $=50 \mathrm{kHz}$, repetition burst rate $=400 \mathrm{~Hz}$, maximum bus signal amplitude $=5 \mathrm{~V}(p-p)$.
3. The AC characteristics are in accordance with the $I^{2} \mathrm{C}$-bus specification. This specification, "The $\mathrm{I}^{2} \mathrm{C}$-bus and how to use it', can be ordered using the code 939839340011.
$1^{2} \mathrm{C}$-BUS PROTOCOL
$1^{2} \mathrm{C}$-bus format

| $S^{(1)}$ | SLAVE ADDRESS | $(2)$ | $A^{(3)}$ | SUBADDRESS | $(4)$ | $A^{(3)}$ | DATA $^{(5)}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $A^{(3)}$ | $P^{(6)}$ |  |  |  |  |  |  |

## Notes

1. $S=$ START condition.
2. SLAVE ADDRESS $(M A D)=10000000$.
3. $A=$ acknowledge, generated by the slave.
4. SUBADDRESS (SAD), see Table 1.
5. DATA, see Table 1 ; if more than 1 byte of DATA is transmitted, then auto-increment of the significant subaddress is performed.
6. $P=S T O P$ condition.

Table 1 Second byte after MAD

| FUNCTION | BIT | MSB |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 7 | 6 | 5 | 4 | 3 | $2^{(1)}$ | $1^{(1)}$ | $0^{(1)}$ |
| Volume/loudness | V | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Fader front right | FFR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| Fader front left | FFL | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| Fader rear right | FRR | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| Fader rear left | FRL | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| Bass | BA | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 |
| Treble | TR | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |
| Switch | S | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |

## Note

1. Significant subaddress.

Table 2 Definition of third byte after MAD and SAD

| FUNCTION | BIT | MSB |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Volume/loudness | V | ZCM ${ }^{(1)}$ | LOFF ${ }^{(2)}$ | V5 ${ }^{(3)}$ | V4 ${ }^{(3)}$ | V3 ${ }^{(3)}$ | $\mathrm{V} 2^{(3)}$ | $\mathrm{V} 1^{(3)}$ | $\mathrm{V} 0^{(3)}$ |
| Fader front right | FFR | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | FFR5 ${ }^{(5)}$ | FFR4 ${ }^{(5)}$ | FFR3 ${ }^{(5)}$ | FFR2 ${ }^{(5)}$ | FFR1 ${ }^{(5)}$ | FFR0 ${ }^{(5)}$ |
| Fader front left | FFL | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | FFL5 ${ }^{(6)}$ | FFL4 ${ }^{(6)}$ | FFL3 ${ }^{6}$ ) | FFL2 ${ }^{(6)}$ | FFL1 ${ }^{(6)}$ | FFL0 ${ }^{6}$ ) |
| Fader rear right | FRR | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | FRR5 ${ }^{(7)}$ | FRR4 ${ }^{(7)}$ | FRR3 ${ }^{(7)}$ | FRR2 ${ }^{(7)}$ | FRR1 ${ }^{(7)}$ | FRR0 ${ }^{(7)}$ |
| Fader rear left | FRL | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | FRL5 ${ }^{(8)}$ | FRL4 ${ }^{(8)}$ | FRL3 ${ }^{(8)}$ | FRL2 ${ }^{(8)}$ | FRL1 ${ }^{(8)}$ | FRLO ${ }^{(8)}$ |
| Bass | BA | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | BA4 ${ }^{(9)}$ | BA3 ${ }^{(9)}$ | BA2 ${ }^{(9)}$ | BA1 ${ }^{(9)}$ | BAO ${ }^{(9)}$ |
| Treble | TR | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | TR4 ${ }^{(10)}$ | TR3 ${ }^{(10)}$ | TR2 ${ }^{(10)}$ | TR1 ${ }^{(10)}$ | TR0 ${ }^{(10)}$ |
| Switch | S | GMU ${ }^{(11)}$ | $\mathrm{X}^{(4)}$ | $\mathrm{X}^{(4)}$ | X ${ }^{(4)}$ | $\mathrm{X}^{(4)}$ | SC2 ${ }^{(12)}$ | SC1 ${ }^{(12)}$ | SC0 ${ }^{(12)}$ |

## Notes

1. Zero crossing mode.
2. Switch loudness on/off.
3. Volume control.
4. Don't care bits (logic 1 during testing).
5. Fader control front right.
6. Fader control front left.
7. Fader control rear right.
8. Fader control rear left.
9. Bass control.
10. Treble control.
11. Mute control for all outputs (general mute).
12. Source selector control.

Table 3 Volume I setting

| $\mathbf{G}_{\mathbf{v}}$ <br> $(\mathrm{dB})$ | DATA |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | V5 | V4 | V3 | V2 | V1 | V0 |  |

Loudness on: the increment of the loudness characteristics is linear at every volume step in the range from +20 to - $\mathbf{1 1} \mathrm{dB}$

| +20 | 1 | 1 | 1 | 1 | 1 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| +19 | 1 | 1 | 1 | 1 | 1 | 0 |
| +18 | 1 | 1 | 1 | 1 | 0 | 1 |
| +17 | 1 | 1 | 1 | 1 | 0 | 0 |
| +16 | 1 | 1 | 1 | 0 | 1 | 1 |
| +15 | 1 | 1 | 1 | 0 | 1 | 0 |
| +14 | 1 | 1 | 1 | 0 | 0 | 1 |
| +13 | 1 | 1 | 1 | 0 | 0 | 0 |
| +12 | 1 | 1 | 0 | 1 | 1 | 1 |
| +11 | 1 | 1 | 0 | 1 | 1 | 0 |
| +10 | 1 | 1 | 0 | 1 | 0 | 1 |
| +9 | 1 | 1 | 0 | 1 | 0 | 0 |
| +8 | 1 | 1 | 0 | 0 | 1 | 1 |
| +7 | 1 | 1 | 0 | 0 | 1 | 0 |
| +6 | 1 | 1 | 0 | 0 | 0 | 1 |
| +5 | 1 | 1 | 0 | 0 | 0 | 0 |
| +4 | 1 | 0 | 1 | 1 | 1 | 1 |
| +3 | 1 | 0 | 1 | 1 | 1 | 0 |
| +2 | 1 | 0 | 1 | 1 | 0 | 1 |
| +1 | 1 | 0 | 1 | 1 | 0 | 0 |
| 0 | 1 | 0 | 1 | 0 | 1 | 1 |
| -1 | 1 | 0 | 1 | 0 | 1 | 0 |
| -2 | 1 | 0 | 1 | 0 | 0 | 1 |
| -3 | 1 | 0 | 1 | 0 | 0 | 0 |
| -4 | 1 | 0 | 0 | 1 | 1 | 1 |
| -5 | 1 | 0 | 0 | 1 | 1 | 0 |
| -6 | 1 | 0 | 0 | 1 | 0 | 1 |
| -7 | 1 | 0 | 0 | 1 | 0 | 0 |
| -8 | 1 | 0 | 0 | 0 | 1 | 1 |
| -9 | 1 | 0 | 0 | 0 | 1 | 0 |
| -10 | 1 | 0 | 0 | 0 | 0 | 1 |
| -11 | 1 | 0 | 0 | 0 | 0 | 0 |


| $\mathrm{G}_{\mathrm{V}}$ <br> (dB) | DATA |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | V5 | V4 | V3 | V2 | V1 | v0 |  |

Loudness characteristic is constant in a range from -11 dB to $\mathbf{- 3 1} \mathrm{dB}$

| -12 | 0 | 1 | 1 | 1 | 1 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| -13 | 0 | 1 | 1 | 1 | 1 | 0 |
| -14 | 0 | 1 | 1 | 1 | 0 | 1 |
| -15 | 0 | 1 | 1 | 1 | 0 | 0 |
| -16 | 0 | 1 | 1 | 0 | 1 | 1 |
| -17 | 0 | 1 | 1 | 0 | 1 | 0 |
| -18 | 0 | 1 | 1 | 0 | 0 | 1 |
| -19 | 0 | 1 | 1 | 0 | 0 | 0 |
| -20 | 0 | 1 | 0 | 1 | 1 | 1 |
| -21 | 0 | 1 | 0 | 1 | 1 | 0 |
| -22 | 0 | 1 | 0 | 1 | 0 | 1 |
| -23 | 0 | 1 | 0 | 1 | 0 | 0 |
| -24 | 0 | 1 | 0 | 0 | 1 | 1 |
| -25 | 0 | 1 | 0 | 0 | 1 | 0 |
| -26 | 0 | 1 | 0 | 0 | 0 | 1 |
| -27 | 0 | 0 | 0 | 0 | 0 | 0 |
| -28 | 0 | 0 | 0 | 1 | 1 | 1 |
| -29 | 0 | 0 | 1 | 1 | 1 | 0 |
| -30 | 0 | 0 |  | 1 | 1 | 0 |
| -31 | 0 | 0 | 1 |  | 1 | 0 |

Repetition of steps in a range from $\mathbf{- 2 8} \mathbf{d B}$ to $\mathbf{- 3 1} \mathrm{dB}$

| -28 | 0 | 0 | 1 | 0 | 1 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| -29 | 0 | 0 | 1 | 0 | 1 | 0 |
| -30 | 0 | 0 | 1 | 0 | 0 | 1 |
| -31 | 0 | 0 | 1 | 0 | 0 | 0 |
| -28 | 0 | 0 | 0 | 1 | 1 | 1 |
| -29 | 0 | 0 | 0 | 1 | 1 | 0 |
| -30 | 0 | 0 | 0 | 1 | 0 | 1 |
| -31 | 0 | 0 | 0 | 1 | 0 | 0 |
| -28 | 0 | 0 | 0 | 0 | 1 | 1 |
| -29 | 0 | 0 | 0 | 0 | 1 | 0 |
| -30 | 0 | 0 | 0 | 0 | 0 | 1 |
| -31 | 0 | 0 | 0 | 0 | 0 | 0 |

Sound fader control circuit

Table 4 Volume II setting (fader and balance); note 1

| $\begin{aligned} & \mathrm{G}_{\mathrm{v}} \\ & (\mathrm{~dB}) \end{aligned}$ | DATA |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FRR5 | FRR4 | FRR3 | FRR2 | FRR1 | FRR0 |
|  | FRL5 | FRL4 | FRL3 | FRL2 | FRL1 | FRLO |
|  | FFL5 | FFL4 | FFL3 | FFL2 | FFL1 | FFLO |
|  | FFR5 | FFR4 | FFR3 | FFR2 | FFR1 | FFR0 |
| 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| -1 | 1 | 1 | 1 | 1 | 1 | 0 |
| -2 | 1 | 1 | 1 | 1 | 0 | 1 |
| -3 | 1 | 1 | 1 | 1 | 0 | 0 |
| -4 | 1 | 1 | 1 | 0 | 1 | 1 |
| -5 | 1 | 1 | 1 | 0 | 1 | 0 |
| -6 | 1 | 1 | 1 | 0 | 0 | 1 |
| -7 | 1 | 1 | 1 | 0 | 0 | 0 |
| -8 | 1 | 1 | 0 | 1 | 1 | 1 |
| -9 | 1 | 1 | 0 | 1 | 1 | 0 |
| -10 | 1 | 1 | 0 | 1 | 0 | 1 |
| -11 | 1 | 1 | 0 | 1 | 0 | 0 |
| -12 | 1 | 1 | 0 | 0 | 1 | 1 |
| -13 | 1 | 1 | 0 | 0 | 1 | 0 |
| -14 | 1 | 1 | 0 | 0 | 0 | 1 |
| -15 | 1 | 1 | 0 | 0 | 0 | 0 |
| -16 | 1 | 0 | 1 | 1 | 1 | 1 |
| -17 | 1 | 0 | 1 | 1 | 1 | 0 |
| -18 | 1 | 0 | 1 | 1 | 0 | 1 |
| -19 | 1 | 0 | 1 | 1 | 0 | 0 |
| -20 | 1 | 0 | 1 | 0 | 1 | 1 |
| -21 | 1 | 0 | 1 | 0 | 1 | 0 |
| -22 | 1 | 0 | 1 | 0 | 0 | 1 |
| -23 | 1 | 0 | 1 | 0 | 0 | 0 |
| -24 | 1 | 0 | 0 | 1 | 1 | 1 |
| -25 | 1 | 0 | 0 | 1 | 1 | 0 |
| -26 | 1 | 0 | 0 | 1 | 0 | 1 |
| -27 | 1 | 0 | 0 | 1 | 0 | 0 |
| -28 | 1 | 0 | 0 | 0 | 1 | 1 |
| -29 | 1 | 0 | 0 | 0 | 1 | 0 |
| -30 | 1 | 0 | 0 | 0 | 0 | 1 |
| -31 | 1 | 0 | 0 | 0 | 0 | 0 |
| -32 | 0 | 1 | 1 | 1 | 1 | 1 |
| -33 | 0 | 1 | 1 | 1 | 1 | 0 |
| -34 | 0 | 1 | 1 | 1 | 0 | 1 |

Sound fader control circuit

| $\begin{aligned} & \mathrm{G}_{\mathrm{v}} \\ & (\mathrm{~dB}) \end{aligned}$ | DATA |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FRR5 | FRR4 | FRR3 | FRR2 | FRR1 | FRRO |
|  | FRL5 | FRL4 | FRL3 | FRL2 | FRL1 | FRLO |
|  | FFL5 | FFL4 | FFL3 | FFL2 | FFL1 | FFLO |
|  | FFR5 | FFR4 | FFR3 | FFR2 | FFR1 | FFRO |
| -35 | 0 | 1 | 1 | 1 | 0 | 0 |
| -36 | 0 | 1 | 1 | 0 | 1 | 1 |
| -37 | 0 | 1 | 1 | 0 | 1 | 0 |
| -38 | 0 | 1 | 1 | 0 | 0 | 1 |
| -39 | 0 | 1 | 1 | 0 | 0 | 0 |
| -40 | 0 | 1 | 0 | 1 | 1 | 1 |
| -41 | 0 | 1 | 0 | 1 | 1 | 0 |
| -42 | 0 | 1 | 0 | 1 | 0 | 1 |
| -43 | 0 | 1 | 0 | 1 | 0 | 0 |
| -44 | 0 | 1 | 0 | 0 | 1 | 1 |
| -45 | 0 | 1 | 0 | 0 | 1 | 0 |
| -46 | 0 | 1 | 0 | 0 | 0 | 1 |
| -47 | 0 | 1 | 0 | 0 | 0 | 0 |
| -48 | 0 | 0 | 1 | 1 | 1 | 1 |
| -49 | 0 | 0 | 1 | 1 | 1 | 0 |
| -50 | 0 | 0 | 1 | 1 | 0 | 1 |
| -51 | 0 | 0 | 1 | 1 | 0 | 0 |
| -52 | 0 | 0 | 1 | 0 | 1 | 1 |
| -53 | 0 | 0 | 1 | 0 | 1 | 0 |
| -54 | 0 | 0 | 1 | 0 | 0 | 1 |
| -55 | 0 | 0 | 1 | 0 | 0 | 0 |
| mute | 0 | 0 | 0 | 1 | 1 | 1 |
| mute | 0 | 0 | 0 | 1 | 1 | 0 |
| mute | 0 | 0 | 0 | 1 | 0 | 1 |
| mute | 0 | 0 | 0 | 1 | 0 | 0 |
| mute | 0 | 0 | 0 | 0 | 1 | 1 |
| mute | 0 | 0 | 0 | 0 | 1 | 0 |
| mute | 0 | 0 | 0 | 0 | 0 | 1 |
| mute | 0 | 0 | 0 | 0 | 0 | 0 |

## Note

1. For a particular range the data is always the same, only the subaddress changes.

Table 5 Bass setting

| $G_{\text {bass }}$ <br> (dB) | DATA |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | BA4 | BA3 | BA2 | BA1 | BAO |
| +15.0 | 1 | 1 | 1 | 1 | 1 |
| +13.5 | 1 | 1 | 1 | 1 | 0 |
| +15.0 | 1 | 1 | 1 | 0 | 1 |
| +13.5 | 1 | 1 | 1 | 0 | 0 |
| +15.0 | 1 | 1 | 0 | 1 | 1 |
| +13.5 | 1 | 1 | 0 | 1 | 0 |
| +12.0 | 1 | 1 | 0 | 0 | 1 |
| +10.5 | 1 | 1 | 0 | 0 | 0 |
| +9.0 | 1 | 0 | 1 | 1 | 1 |
| +7.5 | 1 | 0 | 1 | 1 | 0 |
| +6.0 | 1 | 0 | 1 | 0 | 1 |
| +4.5 | 1 | 0 | 1 | 0 | 0 |
| +3.0 | 1 | 0 | 0 | 1 | 1 |
| +1.5 | 1 | 0 | 0 | 1 | 0 |
| $0^{(1)}$ | 1 | 0 | 0 | 0 | 1 |
| $0^{(2)}$ | 1 | 0 | 0 | 0 | 0 |
| -1.5 | 0 | 1 | 1 | 1 | 1 |
| -3.0 | 0 | 1 | 1 | 1 | 0 |
| -4.5 | 0 | 1 | 1 | 0 | 1 |
| -6.0 | 0 | 1 | 1 | 0 | 0 |
| -7.5 | 0 | 1 | 0 | 1 | 1 |
| -9.0 | 0 | 1 | 0 | 1 | 0 |
| -10.5 | 0 | 1 | 0 | 0 | 1 |
| -12.0 | 0 | 1 | 0 | 0 | 0 |
| -13.5 | 0 | 0 | 1 | 1 | 1 |
| -15.0 | 0 | 0 | 1 | 1 | 0 |
| -13.5 | 0 | 0 | 1 | 0 | 1 |
| -15.0 | 0 | 0 | 1 | 0 | 0 |
| note 3 | 0 | 0 | 0 | 1 | 1 |
| note 3 | 0 | 0 | 0 | 1 | 0 |
| note 3 | 0 | 0 | 0 | 0 | 1 |
| notes 3 and 4 | 0 | 0 | 0 | 0 | 0 |

## Notes

1. Recommended data word for step 0 dB .
2. Result of 1.5 dB boost and 1.5 dB attenuation.
3. The last four bass control data words mute the bass response.
4. The last bass control and treble control data words (00000) enable the external equalizer connection.

Table 6 Treble setting

| $\mathrm{G}_{\text {treble }}$ <br> (dB) | DATA |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | TR4 | TR3 | TR2 | TR1 | TRO |
| +12.0 | 1 | 1 | 1 | 1 | 1 |
| +10.5 | 1 | 1 | 1 | 1 | 0 |
| +12.0 | 1 | 1 | 1 | 0 | 1 |
| +10.5 | 1 | 1 | 1 | 0 | 0 |
| +12.0 | 1 | 1 | 0 | 1 | 1 |
| +10.5 | 1 | 1 | 0 | 1 | 0 |
| +12.0 | 1 | 1 | 0 | 0 | 1 |
| +10.5 | 1 | 1 | 0 | 0 | 0 |
| +9.0 | 1 | 0 | 1 | 1 | 1 |
| +7.5 | 1 | 0 | 1 | 1 | 0 |
| +6.0 | 1 | 0 | 1 | 0 | 1 |
| +4.5 | 1 | 0 | 1 | 0 | 0 |
| +3.0 | 1 | 0 | 0 | 1 | 1 |
| +1.5 | 1 | 0 | 0 | 1 | 0 |
| $0^{(1)}$ | 1 | 0 | 0 | 0 | 1 |
| $0^{(2)}$ | 1 | 0 | 0 | 0 | 0 |
| -1.5 | 0 | 1 | 1 | 1 | 1 |
| -3.0 | 0 | 1 | 1 | 1 | 0 |
| -4.5 | 0 | 1 | 1 | 0 | 1 |
| -6.0 | 0 | 1 | 1 | 0 | 0 |
| -7.5 | 0 | 1 | 0 | 1 | 1 |
| -9.0 | 0 | 1 | 0 | 1 | 0 |
| -10.5 | 0 | 1 | 0 | 0 | 1 |
| -12.0 | 0 | 1 | 0 | 0 | 0 |
| note 3 | 0 | 0 | 1 | 1 | 1 |
| note 3 | 0 | 0 | 1 | 1 | 0 |
| note 3 | 0 | 0 | 1 | 0 | 1 |
| note 3 | 0 | 0 | 1 | 0 | 0 |
| note 3 | 0 | 0 | 0 | 1 | 1 |
| note 3 | 0 | 0 | 0 | 1 | 0 |
| note 3 | 0 | 0 | 0 | 0 | 1 |
| notes 3 and 4 | 0 | 0 | 0 | 0 | 0 |

## Notes

1. Recommended data word for step 0 dB .
2. Result of 1.5 dB boost and 1.5 dB attenuation.
3. The last eight treble control data words select treble output.
4. The last treble control and bass control data words (00000) enable the external equalizer connection.

Table 7 Loudness setting

| CHARACTERISTIC | DATA LOFF |
| :--- | :---: |
| With loudness | 0 |
| Linear | 1 |

Table 8 Selected input

| FUNCTION | DATA |  |  |
| :--- | :---: | :---: | :---: |
|  | SC2 | SC1 | SC0 |
| Stereo inputs IAL and IAR | 1 | 1 | 1 |
| Stereo inputs IBL and IBR | 1 | 1 | 0 |
| Stereo inputs ICL and ICR | 1 | 0 | 1 |
| Stereo inputs IDL and IDR | 1 | 0 | 0 |
| Mono input IMO | 0 | $\mathrm{X}^{(1)}$ | $\mathrm{X}^{(1)}$ |

## Note

1. $X=$ don't care bits (logic 1 during testing).

Table 9 Mute mode

| FUNCTION | DATA |  |
| :--- | :---: | :---: |
|  | GMU | ZCM |
| Direct mute off | 0 | 0 |
| Mute off delayed until the next zero <br> crossing | 0 | 1 |
| Direct mute | 1 | 0 |
| Mute delayed until the next zero <br> crossing | 1 | 1 |

Sound fader control circuit




Fig. 5 Volume control with loudness (including low roll-off frequency).






Fig. 10 Mute function diagram.

If the 20 dB gain is not required for the maximum volume position, it will be an advantage to use the maximum boost gain and then increased attenuation in the last section, Volume II.

Therefore the loudness will be at the correct place and a lower noise and offset voltage will be achieved.

a.

b.
a. Gain volume $\mathrm{I}=20 \mathrm{~dB}\left(\mathrm{G}_{\mathrm{v}(\max )}\right)$; gain volume $\mathrm{II}=0 \mathrm{~dB}$; fader and balance range $=55 \mathrm{~dB}$.
b. Gain volume $\mathrm{I}=20 \mathrm{~dB}\left(\mathrm{G}_{\mathrm{v}(\max )}\right)$; gain volume $\mathrm{II}=-6 \mathrm{~dB}$ global setting; fader and balance range now 49 dB , previously 55 dB .

Fig. 11 Level diagram.


Fig. 12 Turn-on/off power supply circuit diagram.

(1) $\mathrm{V}_{\mathrm{CC}}$
(2) $\mathrm{V}_{\mathrm{O}}$.

Fig. 13 Turn-on/off behaviour.


Fig. 14 Test circuit for power supply ripple rejection (RR).


Fig. 15 Test circuit for channel separation ( $\alpha_{\text {cs }}$ ).

## Selection of input signals by using the zero crossing mute mode

A selection from input $A$ (IAL) to input $B$ (IBL) left sources produces a modulation click depending on the difference of the signal values at the time of switching.

At $t_{1}$ the maximum possible difference between signals is $7 \mathrm{~V}(\mathrm{p}-\mathrm{p})$ (see Fig. 16) and gives a large click. Using the cross detector no modulation click is audible.

For example: The selection is enabled at $t_{1}$, the microcontroller sets the zero cross bit (ZCM = 1) and then the mute bit $(G M U=1)$ via the ${ }^{2} \mathrm{C}$-bus. The output signal
follows the input A signal, until the next zero crossing occurs and then activates mute.

After a fixed delay time at $t_{2}$, the microcontroller sends the bits for input switching and mute inactive.

The output signal remains muted until the next signal zero crossing of input $B$ (IBL) occurs, and then follows that signal.

The delay time $\mathrm{t}_{2}-\mathrm{t}_{1}$ is e.g. 40 ms . Therefore the capacity $\mathrm{C}_{\mathrm{m}}=3.3 \mathrm{nF}$. The zero cross function is working at the lowest frequency of 40 Hz determined by the $\mathrm{C}_{\mathrm{m}}$ capacitor.

(1) Input A (IAL).
(2) Output.
(3) Input B (IBL).

Fig. 16 Zero cross function; only one channel shown.

## Loudness filter calculation example

Figure 17 shows the basic loudness circuit with an external low-pass filter application. R1 allows an attenuation range of 21 dB while the boost is determined by the gain stage $\mathrm{V}_{2}$. Both result in a loudness control range of +20 to -12 dB .

Defining $f_{\text {ref }}$ as the frequency where the level does not change while switching loudness on/off. The external resistor R3 for $f_{\text {ref }} \rightarrow \infty$ can be calculated as:

$\mathrm{R} 3=3.2 \mathrm{k} \Omega$ is generated.
For the low-pass filter characteristic the value of the external capacitor C1 can be determined by setting a specific boost for a defined frequency and referring the gain to $G_{v}$ at $f_{\text {ref }}$ as indicated above.

$$
\left|\frac{1}{j(\omega C 1)}\right|=\frac{(R 1+R 3) \times 10^{\frac{G_{v}}{20}}-R 3}{1-10^{\frac{G_{v}}{20}}}
$$

For example: 3 dB boost at $\mathrm{f}=1 \mathrm{kHz}$
$\mathrm{G}_{\mathrm{v}}=\mathrm{G}_{\mathrm{v}(\mathrm{ref})}+3 \mathrm{~dB}=-18 \mathrm{~dB} ; \mathrm{f}=1 \mathrm{kHz}$ and $\mathrm{C} 1=100 \mathrm{nF}$.
If a loudness characteristic with additional high frequency boost is desired, an additional high-pass section has to be included in the external filter circuit as indicated in the block diagram. A filter configuration that provides AC coupling avoids offset voltage problems.

Sound fader control circuit
TEA6320

## INTERNAL PIN CONFIGURATIONS

Values shown in Figs 18 to 30 are typical DC values;
$\mathrm{V}_{\mathrm{CC}}=8.5 \mathrm{~V}$.


Fig. 18 Pin 1: SDA ( $I^{2} \mathrm{C}$-bus data).

Fig. 20 Pins 5 and 28: treble control capacitors.

Fig. 19 Pins 3, 4, 29, 30: output signals.


Fig. 21 Pins 6 and 27: bass control capacitor outputs.



Fig. 22 Pins 7 and 26: bass control capacitor inputs.


Fig. 23 Pins 8 and 25: input volume 1, control part.


Fig. 25 Pins 10 and 23 : output source selector.

Sound fader control circuit


Fig. 26 Pins 11, 13 to 18, 20, 22: inputs.


Fig. 27 Pin 12: mute control.


Fig. 28 Pin 19: filtering for supply; pin 21: reference voltage.


Fig. 29 Pin 31: supply voltage.


Fig. 30 Pin 32: SCL ( ${ }^{2} \mathrm{C}$-bus clock).

## PACKAGE OUTLINES

SDIP32: plastic shrink dual in-line package; 32 leads ( 400 mil )


DIMENSIONS ( mm are the original dimensions)

| UNIT | $\mathbf{A}$ <br> max. | $\mathbf{A}_{\mathbf{1}}$ <br> $\boldsymbol{m i n}$. | $\mathbf{A}_{\mathbf{2}}$ <br> max. | $\mathbf{b}$ | $\mathbf{b}_{\mathbf{1}}$ | $\mathbf{c}$ | $\mathbf{D}^{(\mathbf{1})}$ | $\mathbf{E}^{(\mathbf{1})}$ | $\mathbf{e}$ | $\mathbf{e}_{\mathbf{1}}$ | $\mathbf{L}$ | $\mathbf{M}_{\mathbf{E}}$ | $\mathbf{M}_{\mathbf{H}}$ | $\mathbf{w}$ | $\mathbf{Z}^{(\mathbf{1})}$ <br> $\mathbf{m a x}$. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| mm | 4.7 | 0.51 | 3.8 | 1.3 | 0.53 | 0.32 | 29.4 | 9.1 | 1.778 | 10.16 | 3.2 | 10.7 | 12.2 | 0.18 | 1.6 |

Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

| OUTLINE <br> VERSION | REFERENCES |  |  |  | EUROPEAN <br> PROJECTION | ISSUE DATE |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | IEC | JEDEC | EIAJ |  |  |  |
|  |  |  |  |  | - |  |



DIMENSIONS (inch dimensions are derived from the original mm dimensions)

| UNIT | A max. | $\mathrm{A}_{1}$ | $\mathrm{A}_{2}$ | $\mathrm{A}_{3}$ | $\mathrm{b}_{\mathrm{p}}$ | c | $D^{(1)}$ | $E^{(1)}$ | e | $\mathrm{H}_{\mathrm{E}}$ | L | $\mathrm{L}_{\mathrm{p}}$ | Q | v | w | y | $\mathrm{Z}^{(1)}$ | $\theta$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| mm | 2.65 | $\begin{aligned} & 0.3 \\ & 0.1 \end{aligned}$ | $\begin{aligned} & 2.45 \\ & 2.25 \end{aligned}$ | 0.25 | $\begin{aligned} & 0.49 \\ & 0.36 \end{aligned}$ | $\begin{aligned} & 0.27 \\ & 0.18 \end{aligned}$ | $\begin{aligned} & \hline 20.7 \\ & 20.3 \end{aligned}$ | $\begin{aligned} & 7.6 \\ & 7.4 \end{aligned}$ | 1.27 | $\begin{aligned} & 10.65 \\ & 10.00 \end{aligned}$ | 1.4 | $\begin{aligned} & 1.1 \\ & 0.4 \end{aligned}$ | $\begin{aligned} & 1.2 \\ & 1.0 \end{aligned}$ | 0.25 | 0.25 | 0.1 | $\begin{aligned} & 0.95 \\ & 0.55 \end{aligned}$ | $\begin{aligned} & 8^{\circ} \\ & 0^{\circ} \end{aligned}$ |
| inches | 0.10 | $\begin{aligned} & 0.012 \\ & 0.004 \end{aligned}$ | $\begin{aligned} & 0.096 \\ & 0.086 \end{aligned}$ | 0.01 | $\begin{aligned} & 0.02 \\ & 0.01 \end{aligned}$ | $\begin{aligned} & 0.011 \\ & 0.007 \end{aligned}$ | $\begin{aligned} & 0.81 \\ & 0.80 \end{aligned}$ | $\begin{aligned} & 0.30 \\ & 0.29 \end{aligned}$ | 0.050 | $\begin{aligned} & 0.419 \\ & 0.394 \end{aligned}$ | 0.055 | $\begin{aligned} & 0.043 \\ & 0.016 \end{aligned}$ | $\begin{aligned} & 0.047 \\ & 0.039 \end{aligned}$ | 0.01 | 0.01 | 0.004 | $\begin{aligned} & 0.037 \\ & 0.022 \end{aligned}$ |  |

Note

1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.

| OUTLINE VERSION | REFERENCES |  |  | EUROPEAN PROJECTION | ISSUE DATE |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | IEC | JEDEC | EIAJ |  |  |
| SOT287-1 |  |  |  | $\square$ | $\begin{aligned} & 95-01-25 \\ & 97-05-22 \end{aligned}$ |

## SOLDERING

## Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC Package Databook" (order code 9398652 90011).

## SDIP

## Soldering by dipping or by wave

The maximum permissible temperature of the solder is $260^{\circ} \mathrm{C}$; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ( $\mathrm{T}_{\text {stg max }}$ ). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

## Repairing soldered joints

Apply a low voltage soldering iron (less than 24 V ) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than $300^{\circ} \mathrm{C}$ it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and $400^{\circ} \mathrm{C}$, contact may be up to 5 seconds.

## SO

## Reflow soldering

Reflow soldering techniques are suitable for all SO packages.
Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to $250^{\circ} \mathrm{C}$.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at $45^{\circ} \mathrm{C}$.

## Wave soldering

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is $260^{\circ} \mathrm{C}$, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than $150^{\circ} \mathrm{C}$ within 6 seconds. Typical dwell time is 4 seconds at $250^{\circ} \mathrm{C}$.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

## Repairing soldered joints

Fix the component by first soldering two diagonallyopposite end leads. Use only a low voltage soldering iron (less than 24 V ) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to $300^{\circ} \mathrm{C}$. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and $320^{\circ} \mathrm{C}$.

## DEFINITIONS

| Data sheet status |  |
| :--- | :--- |
| Objective specification | This data sheet contains target or goal specifications for product development. |
| Preliminary specification | This data sheet contains preliminary data; supplementary data may be published later. |
| Product specification | This data sheet contains final product specifications. |
| Limiting values | Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or <br> more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation <br> of the device at these or at any other conditions above those given in the Characteristics sections of the specification <br> is not implied. Exposure to limiting values for extended periods may affect device reliability. |

## Application information

Where application information is given, it is advisory and does not form part of the specification.

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## NOTES

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