PROJECTS = THEORY = APPLICATIONS = CIRCUITS = TECHNOLOGY

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ve Action

EVERYTHING FOR ELECTRONICS

For Your Model Railroad

n L

0

This season, try an LED Christmas Tree as your table centerpiece.



Meet the new PICAXE M2-Class

crocontroller

control of your

to CB radio

home thermostal

Whatever happened

eb browser

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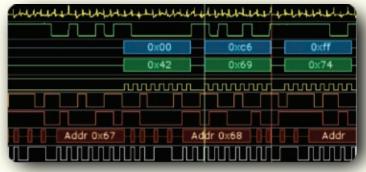
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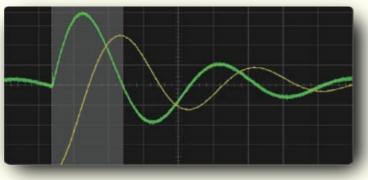
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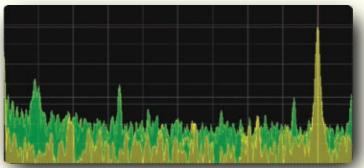
Protocol Analyzer



Digital Oscilloscope



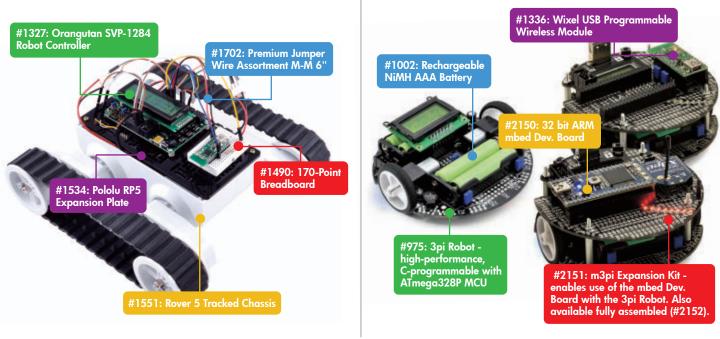
Spectrum Analyzer



bitscope.com/nv

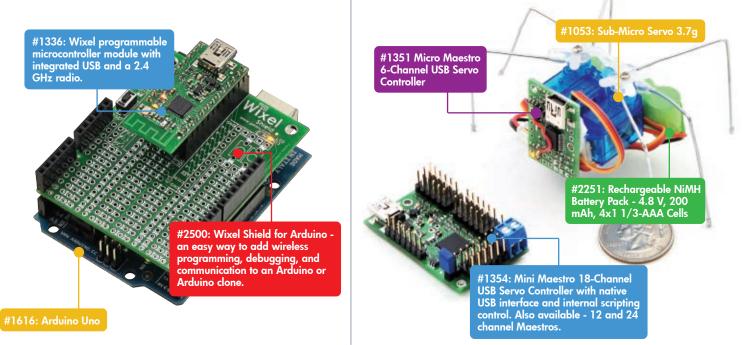
DIY Projects: Simple Tracked Robot

Robots and Robot Kits: Pololu 3pi and m3pi



Programmable Controllers: Wixel and Wixel Shield

Hobby/RC Servo Controllers: Micro and Mini Maestros



Finding the right parts for your robot can be difficult, but you also don't want to spend all your time reinventing the wheel (or motor controller). That's where we come in: Pololu has the unique products - from actuators to wireless modules - that can help you take your robot from idea to reality.



Find these products and more at **www.pololu.com**



Projects & Features

30 Web Browser Control of Your Home Thermostat – Part 2

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You Asked and We Listened — **New Forums are ONLINE!**

This issue marks the first time that Nuts & Volts content will be directly linked to our online presence! We have completely rearchitected the forums so they now closely reflect the content of the magazine. Every regular column now has its own dedicated forum with the column author as your host! In addition, there are also forums to support project articles for reader projects and even an area where you can learn about what it takes to become a writer for the magazine! This new forum design invites you to interact with the author(s), read about corrections or modifications to projects, post questions, or just find and interface with lots of other folks who enjoy the same things you do. As an added bonus, we have populated the forums with example articles by each author so you can read them there if you don't have time to read them here. We hope you find the new forum layout exciting and useful. Please come by, have a look around, and introduce yourself! To explore the new forums, just point your browser to: http://forum.servomagazine.com.

Hope to see you online soon! :)

Vern Graner, Forum Moderator, Nuts & Volts

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Digital Multimeter Pro-50A



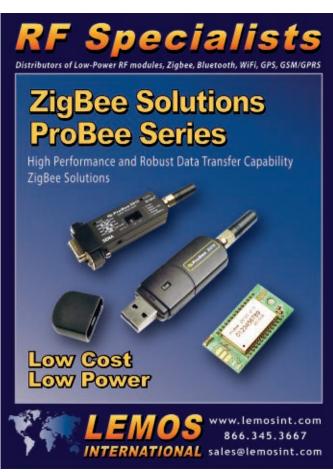
Model **PRO-50A** is a highly versatile, economically-priced, hand-held digital multimeter. The New Model PRO-50A's versatility & multi-functional measuring capabilities makes is perfect for use by engineers, technicians, electricians and students.



Biology and Electronics

f you recall the experiment from your high school biology, Luigi Galvani found that electricity could animate the legs of a dead frog. This simple experiment is worth noting because the intersection of biology and electricity has come a long way since Galvani's experiments in 1791. Today, medical and physiological electronics are multi-billion dollar industries, and the tools for experimentation have never been as affordable or easily obtainable.

One of the easiest projects to start with is an electronic stethoscope. You can buy a commercial version on eBay for a few dollars if you're not looking for the thinnest, lightest model available. Or, you can build your own. You'll need an inexpensive, acoustic stethoscope head, a microphone element, and an audio amplifier. Once you've learned what your heart, lungs, and intestines sound like, you can use an analog-to-digital converter and your favorite microcontroller



to create a heart rate or breath rate display. Another option is to pick up and teardown an inexpensive blood pressure monitor from your local drug store. Can you figure out how they're determining pressure? You may have to search the Web for details, such as how the sound of blood flowing through arteries changes as external pressure is applied.

Then, there's the kit approach. I've seen several electrocardiogram (ECG) kits on the market. Ramsey (**www.ramsey.com**) sells one for about \$50. If you're familiar with op-amps, want a great way to learn how to work with op-amps and low level, low frequency signals, then build your own ECG monitor. I'd take the precaution of NEVER using an AC powered supply, even one of those low voltage bricks. In addition, use an optical isolator to minimize the chance that a voltage could be applied to your body as you're testing the circuit.

Of course, there are experiments you can perform in which signals are intentionally sent to your body, as in Galvani's experiment. Again, have a look on eBay for one of those abdominal toners that send pulses of electricity to your ab muscles, causing them to contract. At the other extreme – which I don't recommend – are high voltage tasers designed to temporarily stun.

If you're into exercising and fitness, then you probably know about the heart rate monitor pickups that work with Polar watches, and many of the treadmills on the market. This is another source of 'teardown' material that's relatively inexpensive — especially if you're willing to tear open one of the sealed sensors and replace a battery.

If you're into optics and LEDs, there are dozens of experiments that you can perform with an op-amp, microcontroller, LED, phototransistor, and a handful of discrete components. The simplest is a heart rate monitor based on the reflection or transmission of red light through your fingertip or earlobe. With each contraction of your heart, red blood cells rush through the capillaries in your fingertip. A phototransistor/op-amp circuit can detect the resulting change in transmission/reflection of red light. Once you've mastered heart rate, you can try your hand at blood oxygen saturation (the more oxygen carried by a red blood cell, the more red the cell). You'll have to do a little research to determine which wavelengths of light to use.

I've been working with a few simple sensors — a piezo microphone element and a few resistive pressure sensors and an Arduino on a number of experiments. I use the piezo element in my shoe to determine each time my shoe strikes the ground — similar to the method used by the popular Apple/Nike running sensor. The pressure sensors mounted at different locations under my running shoe insert



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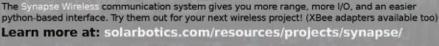
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– record the changes in pressure distribution as I run. My goal is to record changes in pressure distribution at different running paces. Runners at the Olympic Village use similar systems, but these sell for around \$50K. I've amassed most of my sensor collection from teardowns of devices I've picked up from Ebay. I've also purchased a few sensors from SparkFun Electronics and Parallax. Amazon is a convenient source for ECG electrodes, but you'll have to buy in bulk. If you want to experiment with a Polar heart rate monitor sensor strap, then consider the Polar interface from SparkFun. A Polar chest strap sensor is about \$50 on Amazon. As a final note, think about what you're doing. Electricity is potentially dangerous, in part because the salty fluid in your body is an excellent conductor. Don't use AC power and DC – at significant current – can burn. If you're not familiar with adhesive electrodes and other 'direct' connections, start out safe with optical sensors. Even then, consider that you could damage your eyes if you stare long enough into an active LED. Have fun on your journey.

WARNING: SPORTS ANALOGY AHEAD.

Think of XBee like a pitcher, throwing bits of wireless data to a catcher. Synapse is like a pitcher that has a giant mecha arm cannon that shoots data to robo t-rex catcher on a skateboard.







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BY JEFF ECKERT

ADVANCED TECHNOLOGY

SPLIT MAGNET SETS RECORD

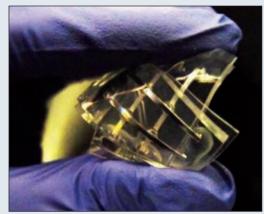
A t 25 tesla (T), it's not the absolute most powerful magnetic device ever built, but apparently it's top dog in the "split magnet" category, i.e., a magnet that is fabricated in two halves, with holes to allow the observation of experiments conducted at its core. Earlier this year, the custom-built, \$2.5 million system, located at the National High Magnetic Field Laboratory at Florida State University (**www.magnet.fsu.edu**), easily broke the previous record of 17.5 T, set by the French in 1991. In addition, it has 1,500 times as much room for



Interior portion of the split coil magnet.

experiments in its "bore," where four large, elliptical ports provide direct horizontal access to the central experimental space. Building a magnet system with ports strong enough to withstand this physical stress was once considered physically impossible. After all, the structure has to withstand 500 tons of pressure pulling the two halves together while allowing 160,000 A of electrical current and 3,500 gallons of cooling water per minute to flow through it. However, through some innovative design work, extensive testing, and locating people who could actually build the thing, it was achieved. Not much specific information was provided about how the magnet will be employed, but it was revealed that "researchers in chemistry, physics, and biology are poised to conduct research using the split magnet, while others are optimistic about the potential for breakthroughs in nanoscience and semiconductor research."

The lab also holds the world's record for magnetic strength at 45 T, achieved by a hybrid (i.e., part resistive and part superconducting) device. The 11,000 sq ft lab houses a variety of research magnets, but only one or two can be operated at a time – the power system can handle only a measly 56 MW. Even so, the lab consumes seven percent of all the electricity used in its hometown of Tallahassee, FL. There's an open house at the lab every February, so check the website for the exact date if you expect to be in the neighborhood.



■ NCSU's soft memristor-type device functions in wet environments.

MEMORY GOING SOFT

For the most part, when your memory starts to go soft, it is not a good thing. Take my word for it. But researchers from North Carolina State University (**www.ncsu.edu**) have developed a memory device "with the physical properties of Jell-O" that seems to have some useful properties. Apparently, it functions well in wet environments which is "opening the door to a new generation of biocompatible electronic devices."

We're talking about a "memristor" type of device which — in the better-known Hewlett-Packard rendition — is composed of two layers of titanium dioxide connected by wire. In the NCSU version, the device is made using a liquid gallium/indium alloy set into a water-based gel. When you expose the alloy electrode to a positive charge, it creates an oxidized skin that resists the flow of electricity equivalent to a binary 0. When you expose it to a negative charge, it becomes conductive, giving us a 1. As explained in an NCSU press release, "Normally, whenever a

negative charge is applied to one side of the electrode, the positive charge would move to the other side and create another oxidized skin – meaning the electrode would always be resistive. To solve that problem, the researchers 'doped' one side of the gel slab with a polymer that prevents the formation of a stable oxidized skin. That way, one electrode is always conducive – giving the device the 1s and 0s it needs for electronic memory.

At this point, the device is just a prototype that admittedly is not yet capable of holding significant amounts of memory. However, because the gels used in the technology offer a high level of biocompatibility, it offers the promise of interfacing electronics directly with biological systems, e.g., living cells or tissues.

COMPUTERS AND NETWORKING

SPEEDIER GRAPHICS CARD

If you're not a hard-core graphics user or serious gamer, the graphics card industry probably looks a little confusing. And, with a huge range of products with similar names coming from several manufacturers, it actually is. So, first a little history. Back in 2000, ATI Technologies (formerly Array Technologies, Inc.) began producing the Radeon line of graphics cards as a successor to its Rage products. In 2006, AMD (www.amd.com) acquired the company. However, AMD no longer manufacturers the cards. Instead, it sells graphics processing units (GPUs) to third-party manufacturers who sell their own versions of AMD Radeon-branded products. Therefore, if you want to buy, let's say, an AMD Radeon HD 6950, it could actually be a product of one of about a dozen companies – even if you buy it through the AMD website. One of those companies is Sapphire Technology (not to be confused with Sapphire Technologies, an IT staffing group soon to be rebranded Randstad Technologies), and its latest contribution is the Sapphire HD 6950 Toxic Edition card. The company claims that it's the fastest production card in the class,



based on clock speeds of 880 MHz for the core and 1,300 MHz (5.2 Gb/sec effective) for the memory. It also offers a dual BIOS feature that allows users to play around with alternative BIOS profiles and settings, and performance can be boosted with the company's TriXX overclocking tool (a free download). Using TriXX, you can modify memory and clock core speeds and voltages, and monitor the results. If you're interested in things like texture processing units and stream processors, you can get details at **www.sapphiretech.com**. As of this writing, price and availability are unknown, but earlier models sell for about \$275.

WIN \$200,000!

t's not all that unusual for software companies to pay "bug bounties" to people who find and fix vulnerabilities, but Microsoft has upped the ante. As announced at this year's Black Hat security conference, the company is sponsoring the \$250,000 BlueHat contest, created to "generate new ideas for defensive approaches to support computer security ... The object of this contest is to design a novel runtime mitigation technology solution that is capable of preventing the exploitation of memory safety vulnerabilities." First prize is \$200,000, second prize is \$50,000, and third prize is a Microsoft Developer Network (MSDN) Universal subscription, valued at \$10,000. All you have to do is design a runtime solution that can prevent the exploitation of memory safety vulnerabilities. Rules are available at www.microsoft.com/security/bluehatprize/. Just make sure your entry is received by April 1, 2012.

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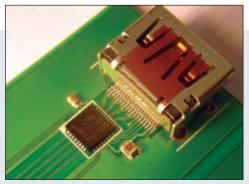
If you have ever wondered if there is a sneaky way to download your favorite videos from YouTube, wonder no more. There are several ways, but the sneakiest is also the easiest. Just go to the URL window in your browser and add the word "kiss" after the "www." part. For example, let's say you're listening to Phil Keaggy doing a typically amazing guitar solo at **www.youtube.com/watch?v=0T3 to3DHLIE**. Change "youtube" to "kissyoutube" and hit return. You'll be taken to **saveyoutube.com** and given several download options. Nothing to it. Just remember that if you do this too often, you're going to need a bigger hard drive.



CIRCUITS AND DEVICES

CHIP OFFERS HDMI PROTECTION

The next generation of home entertainment systems will be able to offer improved signals, as well as protection against electrostatic discharge (ESD) strikes by incorporating a new IC from NXP Semiconductors (www.nxp.com). The company's IP4786CZ32 chip – said to be the industry's only complete HDMI 1.4 transmitter signal-conditioning and protection device – features a transmission line clamping architecture for reduced peak voltages during a strike plus impedance-matching inductors that provide matched 100-differential impedance through the device. By eliminating up to 29 discrete components, the IP4786CZ32 is claimed to offer the fewest components and a 60 percent smaller footprint compared to any other HDMI conditioning and protection scheme. The IC is



A new NXP chip offers both ESD protection and signal conditioning for HDMI signals.

compatible with HDMI 1.3a and 1.4, 340 MHz pixel clock, deep color, and HDMI Ethernet and Audio return Channel (HEAC). It's available now from NXP and distributors, with recommended pricing of \$1.20 in quantities of 1,000.



STUDIO IN YOUR SHIRT POCKET

Once upon a time, a multi-track recording studio was a huge and wondrous thing, filled with expensive mixing boards, monitor speakers, and recording machines with reels of magnetic tape the size of hubcaps. As you recorded your masterpieces, the reels would turn, needles on the VU meters would flip around, colored lights would flash, and Satan would dance around the room in a pink tutu. (Well, that last part may have been just a hallucination.) A lot has changed now, and you can get your own studio for a few hundred bucks and carry it around in your pocket. An example is the Boss Micro BR, model BR-80, a new eight-track micro studio from Roland (www.roland.com). It provides three recording modes: an eight-track MTR (multi-track recorder) mode, an "eBand" mode for onstage backing tracks and phrase training, and a "live rec" mode for making

instant stereo recordings via the built-in condenser mics. It records directly to an SD/SDHC memory card (up to 32 GB) and provides a user interface that basically follows a standard console layout, offering equalization, reverb, and a range of mastering tools. You even get a library of backing and rhythm patterns, and composite object sound modeling (COSM) amps and effects for guitar, bass, and vocals. It is, of course, USB connectable, so you can import and export files from a computer. Not bad for \$300.

GO PLUCK YOURSELF

hat happens if you get excited and buy the Boss BR-80 only to suddenly remember that you can't play guitar, don't want to put in a few thousand hours to learn, and don't have any musical aptitude in the first place? Not to worry! The lunatics at ThinkGeek (**www.thinkgeek.com**) have a solution: the Electronic Rock Guitar Shirt. Just put it on, place your finger in the right spot on the neck, and strum the string with one of the included magnetic picks. Out comes any of 15 power chords, recorded from a real electric guitar. The shirt includes a mini amp that clips onto your belt and features both a tone control and a volume knob that goes all the way up to 11. Not just 10, like your typical Marshall stack, but 11! All of the electronic components can be removed, so you'll have no trouble washing it after it becomes soaked with rancid Budweiser. For a demo by "L. Hadron and the Colliders," just log onto www.youtube.com/watch?v=V6JjY-AKPIY. And, yes, that's the Electronic Drum Kit Shirt on the other guy. Not shown is the Electronic Music Synthesizer Shirt. You can slip on any of them for \$29.99. Batteries not included, of course. NV



The Electronic Rock Guitar Shirt from ThinkGeek.

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- ☑ Vertical sensitivity 1 mV...5 V/div. Lowest noise fan
- ☑ 12 div. x-axis display range
- 20 div. y-axis display range with VirtualScreen function
- $\overline{\mathbf{N}}$ Trigger modes: slope, video, pulsewidth, logic, delayed, event
- \checkmark Component tester, 6 digit counter, Autoset, automeasurement, formula editor, ratiocursor, FFT for spectral analysis

PROGR. 2/3/4 CHANNEL HIGH-PERFORMANCE POWER SUPPLY HMP SERIES



- \checkmark HMP2030: 3x0...32V/0...5A 188W
- \checkmark HMP4030: 3 x 0...32 V/0...10 A, max. 384 W
- \checkmark HMP4040: 4 x 0...32 V/0...10 A, max. 384 W
- \checkmark 188/384W output power realized by intelligent power management
- \checkmark Low residual ripple: < $150\,\mu V_{\rm rms}$ due to linear post regulators
- \checkmark High setting- and read-back resolution of up to 1 mV/0.2 mA
- \checkmark HMP4030/HMP4040: Keypad for direct parameter entry
- \checkmark Galvanically isolated, earth-free and short circuit protected output channels
- \checkmark Advanced parallel- and serial operation via V/I tracking
- \square EasvArb function for free definable V/I characteristics
- \checkmark FuseLink: individual channel combination of electronic fuses
- \checkmark Free adjustable overvoltage protection (OVP) for all outputs
- $\overline{\mathbf{A}}$ All parameters clearly displayed via LCD/glowing buttons

LCR-BRIDGE HM8118



- \checkmark Transformer parameter measurement
- \checkmark External capacitor bias up to 40 V
- \checkmark Kelvin cable and 4 wire SMD Test adapter included in delivery
- \checkmark Galvanically isolated USB/RS-232 Interface, optional IEEE-488

1 GHZ SPECTRUM ANALYZER HMS 1000/HMS 1010

educational

discount



- Frequency range 100 kHz...1 GHz
- $\overline{\mathbf{V}}$ Amplitude measurement range -114...+20 dBm
 - DANL -135dBm with Preamp. Option H03011
- $\overline{\mathbf{A}}$ Sweep time 20 ms...1000 s
- Resolution bandwidth 100 Hz...1 MHz in 1–3 steps, 200 kHz (-3 dB) additional 200 Hz, 9 kHz, 120 kHz, 1 MHz (-6 dB)
- $\overline{\mathbf{A}}$ Spectral purity < -100 dBc / Hz (@100 kHz)
- \checkmark Video bandwidth 10 Hz...1 MHz in 1-3 steps
- \square Tracking Generator (HMS 1010) -20 dBm/0 dBm
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INTRODUCING THE NEW PICAXE M2-CLASS MICROCONTROLLERS

Back in July when I was just beginning to think about a topic for this month's Primer, Revolution Education released three new M2-class processors: the 08M2, 14M2, and 20M2. That settled the issue for me; I immediately ordered a bunch of them, and impatiently waited for the package to arrive at my doorstep. If you prefer to not wait as long as I did, all the M2 processors are now available here in the USA from Peter Anderson (**www.phanderson.com/picaxe/index.html**). In case you have been unable to locate a source for the AXE401 Shield Base, I should mention that Professor Anderson also carries the kit version of the AXE401 on his site.

BRIEF OVERVIEW OF SELECTED FEATURES OF THE M2-CLASS PROCESSORS

We don't have near enough space in this installment of the Primer to discuss all the significant new features of the M2 processors, so I'm going to limit our discussion to just a few of the most significant improvements to the older M-class devices. Naturally, the three new M2 processors have much in common with their older sibling — the PICAXE-18M2 — but they also add a few new tricks of their own, including a built-in internal temperature sensor and two To begin with, all M2 devices are able to operate with a supply voltage as low as 1.8V which means that battery-powered projects now only require a two-cell alkaline battery pack. In addition, the range of the internal clock frequency has been greatly expanded; all M2 processors can operate at nine different frequencies, from 31 kHz to 32 MHz.

Since power consumption is directly related to clock frequency, the new lower frequencies can be used to significantly extend the life of battery-powered projects. Of course, the maximum operating frequency of 32 MHz is four times faster than the 8 MHz maximum of the older Mclass processors, which greatly increases the range of possible M2 projects.

There are also several significant improvements in the memory capacity of the M2 processors. First, the program memory has been increased to 2,028 bytes which is eight times the capacity of the older M-class processors. Also, the 256byte data (EEPROM) memory is now completely separate, so using it doesn't decrease the amount of available program memory. This means that all the M2 processors including the tiny 08M2 - can runprograms containing as many as 1,800 lines of BASIC code, which again greatly increases the range of possible M2 projects.

powerful new commands (*rfin* and *rfout*) that will greatly simplify working with inexpensive RF wireless transmitters and receivers. (I'm sure we'll be exploring that capability before long!)

Figure 1 presents a summary of what I consider to be the most significant new features of the M2 processors.

Feature	08M2	14M2	18M2	20M2
Voltage Range	1.8V - 5.5V	1.8V - 5.5V	1.8V - 5.5V	1.8V - 5.5V
Min. Internal Freq	31kHz	31kHz	31kHz	31kHz
Max. Internal Freq	32MHz	32MHz	32MHz	32MHz
Program Memory	2048 bytes	2048 bytes	2048 bytes	2048 bytes
General-Purpose Variables (b0É b27)	28 bytes	28 bytes	28 bytes	28 bytes
Storage Variables	100 bytes	484 bytes	228 bytes	484 bytes
Total Variables	128 bytes	512 bytes	256 bytes	512 bytes
Pwmout channels	1	4	2	4

The number of general-purpose (GP) variables has been doubled from 14 to 28 (i.e., b0...b27) which also makes more complex projects possible. Finally, the amount of memory space available for "storage variables" has also been greatly

■ FIGURE 1. Selected features of the PICAXE M2 processors.

PICAXE PRIMER

increased. (The M-Class processors supported 48 storage variables.)

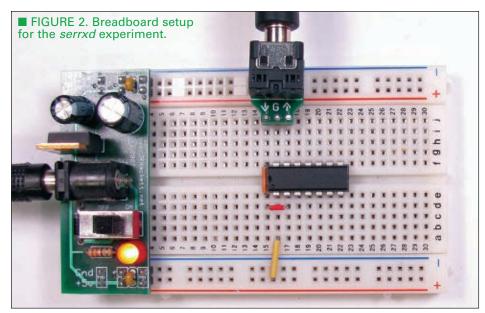
This may not seem like such an important improvement at first. (How often have you needed more than 48 storage variables?) However, the M2 processors also now support the X2class feature of "indirect addressing" of the storage variable memory. In case you're not familiar with indirect addressing, we're going to take a close look at it shortly because we can accomplish some really powerful tasks with it.

If you're interested in microcontroller-based robotics, you will definitely want to check out the 14M2 and 20M2 processors which both include a total of four PWMOUT channels; two of them are entirely independent, and the other two share the "servo" timer. As a result, either processor can control two independent DC motors and several servomotors at the same time (or three independent DC motors). That's an amazing amount of processing power for a small robot controller.

That's about all the space we have this month to introduce the most significant new features of the M2 processors. If you're interested in a more complete overview, RevEd has released a comprehensive M2 datasheet (PICAXE-M2 Product Briefing, available at **www.reved.co.uk/docs/picaxem2.pdf**). At this point, we're going to turn our attention to a project that's been on my mind for some time, and the 14M2 is just the processor for the job.

IMPROVING OUR SERIAL LCD PROJECT

Back in June '09, we began our serial LCD project, and in August of that year we developed a major software driver for it. As you know, we used a PICAXE-14M processor which we pushed to its memory and speed limits. Even so, we had to make some compromises, especially in the requirement of sending data to the LCD in fixed-length packets. This



was necessary because the 14M's *serin* command is "blocking" which means that whenever the 14M executes a *serin* command, the program stops running until the specified number of bytes have been received. (That's why we needed to know how many bytes were going to be transmitted each time.)

One of the many software improvements in the M2-class processors is the inclusion of a *timeout* option for the *serin* and *serrxd* commands. The timeout option enables us to specify how long (in milliseconds) an M2 program should wait at a *serin* or *serrxd* command before "giving up" and moving on if no data is received. (We aren't going to discuss it this month, but you may be interested in knowing that the M2 *irin* command now also has a timeout option.)

As long as we provide more than enough variables in which to store the incoming serial data, the new timeout option will enable our program to move on to processing the data, no matter how many bytes are actually received. That's exactly the capability we need to improve our original serial LCD project.

With that in mind, as soon as I received my new M2 processors in the mail, I removed the 14M from one of my serial LCDs, replaced it with a 14M2, and began modifying the software. I spent more than two

days working on the program, and was getting nowhere. I just couldn't get it to work at all, and the more I tried, the more frustrated I became. (Sound familiar?)

When I couldn't stand it any more, I decided to take a break from it for a day or two, which is a strategy that has helped me in the past. When I returned to it, I simplified my approach, temporarily eliminating the LCD and using the terminal window for output. My goal was to develop a simple serrxd routine that would be able to accept a variable number of bytes as input. My breadboard setup for this experiment is shown in Figure 2. The programming adapter in the photo is the Prog-03 adapter. It's designed specifically for the new M2 processors. As you can see, it minimizes the number of necessary connections (because it also directly connects to the ground rail).

The program we'll be using (*SerrxdFastSimple.bas*) is included in the download files on the *N&V* website, along with the other two programs we'll be using this month. We'll use the following abridged version of the program to discuss the issues I encountered.

、 ***** SerrxdFastSimple.bas)

#com 6
#picaxe 14M2
#no_data
#terminal 9600
October 2011 NUTS!VOLTS 15

```
setfreq m8
```

In line 1, we're initializing the first four GP variables to "*" (ASCII char 42). Don't forget, it's okay to put multiple statements on the same line if you separate them with a colon. Line 2 simply displays a prompt in the terminal window. Line 3 is a traditional blocking serrxd command. Its purpose it to make the compiler wait at this point until a serial character is received, and then move on to the non-blocking serrxd command in line 4. This command includes a timeout option that instructs the compiler to "give up" after a timeout of 500 mS (at 8 MHz) if no additional serial input is received.

The *reconnect* command in line 5 requires a bit of explanation. We have often sent data from a PICAXE program to the terminal window for display, but I don't think we have done the reverse, i.e., sent data from the terminal window to a PICAXE program. It's easy to do; you just type the data into the *Output Buffer* area at the bottom of the terminal window and click the *Send* button. However, there is one important little complication.

Whenever a PICAXE program is running, the processor repetitively scans the *serin* line to see if the Programming Editor (or AxPad) wants to initiate a new program download. The scanning of the *serin* line would disrupt any serial input we want to send to the running program. To avoid this problem, the PICAXE compiler automatically issues a *disconnect* command in the background (see the *disconnect* documentation in Section 2 of the PICAXE manual). Since the processor is now disconnected from the programming software, it's no longer possible to download a new program without first carrying out a *hard-reset* procedure (i.e., disconnecting the power to the processor, initiating the download, and then quickly restoring the power).

A simpler alternative is to include a *reconnect* command immediately after the *serrxd* commands so that program updates can again be downloaded to the processor without necessitating a hard-reset. Finally, line 6 simply transmits the four characters that are currently stored in the first four GP variables to the terminal window.

So, how should this program behave? Since the four GP variables are each initialized to "*" each time through the loop, entering each of the following strings in the *Output Buffer* and clicking *Send* should result in the indicated output, right?

```
Entering "a" should result in a
display of "a***"
Entering "as" should result in
a display of "as**"
Entering "asd" should result in
a display of "asd*"
Entering "asdf" should result
in a display of "asdf"
```

I imagine you can guess the outcome — it doesn't work! What's a frustrated programmer to do?

THE PICAXE FORUM TO THE RESCUE (AGAIN)

I don't know about you, but I hate to admit when I can't solve a problem. However, I was beginning to think that there might be an issue with the new serrxd command, so I put my pride in my back pocket and posted a cry for help on the PICAXE Forum (www.picaxeforum.co.uk/ forum.php). In about an hour, I received a reply from "Technical" (one member of RevEd's technical support team) confirming that there is, in fact, a bug in the current version of the Programming Editor software (v5.4.0, as of August '11). Ironically, this bug doesn't involve the new

timeout option of the *serrxd* command; it affects the older blocking version, and results in the command failing to wait for a serial input. The program immediately moves on to the next command even in the absence of any serial input!

Technical reassured me that the bug would be corrected in the next update of the Programming Editor software. Fortunately, Technical also provided a simple work-around that solves the problem until the software update is released. All I had to do was to replace line 3 in the above program with the following code:

bugfix:

serrxd [60000,bugfix], b0

In order to fully understand why this work-around is effective, you may want to read the serrxd documentation in Section 2 of the PICAXE manual, but I'll briefly explain how it works. The timeout of 60000 forces the serrxd command to wait about 30 seconds (at 8 MHz) for the first serial character to arrive. If a character fails to arrive in that amount of time, the second optional parameter (bugfix) instructs the compiler to jump to the *bugfix* address. Since this address immediately precedes the serrxd command, the effect is to keep the compiler from advancing beyond this point in the program until the first character has been received.

If you modify the program to include Technical's work-around, you will see that it behaves exactly as we would predict. By the time you are reading this, version 5.4.1 of the Programming Editor will probably have been released, so you can just install it on your PC and test the program without the "bug fix" — it should work correctly. If not, let me know and we'll figure it out.

In addition to the *serrxd* issue we just discussed, I had two other minor problems as I was experimenting this month. First, in some of the programs I tested the terminal window failed to display the first line of text that was sent to it. Also, even when the first line did show up in the terminal window, there was frequently a single "garbage" character preceding what I wanted to display.

Since I had already reached my maximum frustration-tolerance level, I didn't hesitate to again ask for help on the PICAXE Forum. Within 20 minutes, I received a reply from "Hippy" (another member of RevEd's technical support team) which explained that a newly downloaded PICAXE program starts running before the terminal window has time to open, so if the program sends data immediately, it will not display. The solution is simple: just put a delay at the beginning of the program.

A couple of hours later, Technical responded to my garbage character issue, asking me which download cable I was using. At the time, I was using a reconfigured FTDI cable with one of my new AxMate-MS programming adapters which – like the Prog-03 in **Figure 2** – is specifically designed for the new M2 processors. (I just can't stop making programming adapters!)

Technical's second reply directed my attention to the schematic diagram for the AXE027 PICAXE USB cable (www.rev-ed.co.uk/docs/ axe027.pdf). In the schematic, I saw a 10K pull-down resistor on the serin line which I know isn't included in the FTDI cable. In fact, that cable includes a 10K pull-up resistor on the serin line. I wasn't sure what to do about that, but the first thing I tried (just adding a 10K pull-down) worked - the garbage character disappeared. Two problems and two solutions within a couple of hours, and it all transpired on a Saturday - that's outstanding technical support!

Ultimately, I decided to permanently install the pull-down resistor on the bottom of my AxMate-MS adapter (see **Figure 3**). If you are using any FTDI-based USB adapter and getting the extra garbage character, you may also want to consider installing the 10K pull-down resistor.

Once my *serrxd* routine was functioning correctly, I was ready to tackle the more complicated part of my serial LCD update project. One of the goals I had for the update was to be able to change what's displayed on both lines of the 16x2 LCD with one serial string. Since there could be a few LCD commands mixed in with the 32 characters that the LCD can display, I was determined to be able to receive and process strings as large as 40 characters.

As a result, the 28 GP variables that are now available on the M2 processors wouldn't be enough to hold all the incoming data in my grandiose little scheme, so I decided to use the storage variables instead. As I mentioned earlier, the 14M2 contains 484 bytes of storage variables, so I was sure it could handle the largest LCD I could possibly find. In addition, the storage area in all M2 processors can be accessed using what's referred to as indirect addressing which is much faster that the usual approach (i.e., direct addressing), so that's the next topic we need to discuss.

DIRECT ADDRESSING OF VARIABLES

As we saw earlier in **Figure 1**, the M2 processors have 28 GP variables and many more storage variables. The values of these variables are stored in an area of RAM that's known as the *byte scratchpad*. The M2 GP variables *b0* through *b27* are assigned to RAM locations 0 through 27 in the byte scratchpad, and the storage variable area begins at RAM location 28 and continues consecutively for however many storage variables the processor contains.

For example, the 14M2 contains a total of 512 bytes of RAM in the byte scratchpad (28 bytes for its GP variables and 484 bytes for storage variables). Since the GP variables occupy locations 0 through 27, the storage variable area in the 14M2 extends from location 28 to location 511.

The standard type of

■ FIGURE 3. Pull-down resistor added to the AxMate-MS.

addressing that we use when we work with the GP variables is referred to as *direct* addressing because we directly read or write the value that is stored at any given location in the byte scratchpad. For example, the value of the first GP variable (b0)resides at RAM location 0, so when we write b0 = 5, we're directly storing the value 5 at location 0 in the byte scratchpad. If we define the name myVar for the b0 variable (symbol myVar = b0), then we can also write $m_V Var = 5$ to directly store the value 5 at location 0 in the byte scratchpad.

PICAXE BASIC includes two commands (peek and poke) that provide a second method of directly addressing the processor's variables. For example, assuming we have already written (symbol myVar = b0), poke myVar, 3 accomplishes the exact same thing as myVar = 3; the value 3 is stored in memory location 0. Of course, you probably would never use the *poke* command for that purpose, because it's much simpler to write $m_V Var = 3$. However, peek and poke can also be used to directly access the storage variables as well, and the simpler method can't because there are no names assigned



to the storage variables, and no way of assigning names to them.

For example, even if you write symbol myStoreVar = 28, you can't write myStorVar = 7 because you have defined a constant, not a variable. However, you can write poke myStoreVar, 7 when you want to store the value 7 at the (constant) location 28.

The most important point to remember about the storage variables is that they can only be used to store and retrieve data; we can't assign convenient names to them or use them in any sort of calculations. For those purposes, we need the GP variables. In spite of these limitations, the storage variables are a very powerful feature of the M2 processors. To fully appreciate the power of the storage variables, we need to understand the concept of indirect addressing.

INDIRECT ADDRESSING OF VARIABLES

Indirect addressing is more complicated (and much more powerful) than *direct* addressing. Rather than directly addressing the memory location we want to access, indirect addressing uses the concept of a *pointer* variable which "points to" the desired address. PICAXE BASIC includes the following four built-in special function variables that we can use to implement indirect addressing (I'll explain each one shortly):

bptr – (pronounced bee pointer); the byte scratchpad pointer.

@bptr – (pronounced *at bee pointer*); the byte scratchpad value pointed to by bptr.

@bptrinc – (pronounced *at bee pointer inc*); the byte scratchpad value pointed to by bptr (post increment).

@bptrdec – (pronounced *at bee pointer dec*); the byte scratchpad value pointed to by bptr (post decrement).

The byte scratchpad pointer **18 NUTSVOLTS** October 2011 (*bptr*) is used to point to the location we want to access, and the other three variables automatically contain the value that's stored at the location pointed to by *bptr*. I know that sounds confusing, so let's take a look at a simple program to clarify all this. (the line numbers are included for the following discussion):

<pre>* **StorageVarDemo.bas #com 6 #picaxe 14M2 #no_data #terminal 9600</pre>	**
setfreq m8	
bptr = 28	` [1]
for b0 = 65 to 90 @bptrinc = b0 next b0	`[2] `[3] `[4]
@bptr = 0	` [5]
bptr = 28	` [6]
do until @bptr = 0 sertxd (@bptrinc) loop	`[7] `[8] `[9]

In line 1, *bptr* is initialized to 28 (the first storage variable location). In lines 2-4, we're executing a *for...next* loop 26 times. The start and end values for the loop (65 and 90) correspond to the ASCII values for "A" and "Z," so the first time line 3 is executed, the ASCII character "A" is stored at the location pointed to by *bptr*, and then *bptr* is automatically incremented in preparation for the next iteration of the loop. (This automatic incrementing of *bptr* is what makes indirect addressing so powerful.)

The second time through the loop, "B" is stored at location 29, and *bptr* is again automatically incremented. When the loop has finished executing, the characters "A" through "Z" have been sequentially stored at locations 28 through 53. At this point, *bptr* = 54 because it started at 28 and was automatically incremented 26 times. In line 5, we're storing the value zero (which is a non-printing ASCII character) at location 54. We're going to use the zero as an "end of string" marker, as we're about to see.

In line 6, we need to reset *bptr* back to its initial value of 28 in

preparation for sending the received characters to the terminal window. Here's where our end of string marker comes into play. In lines 7-9, we execute a *do until*... loop that sends each stored character to the terminal window and automatically increments *bptr* in preparation for the next iteration of the loop. When we reach the zero, the loop immediately terminates, so the zero is not sent to the terminal window.

The program we just examined is also included in this month's downloads. If you don't yet have them, it would be a good idea to download all three programs at this point, and spend some time experimenting with *StorageVarDemo.bas* so that you have a good understanding of indirect addressing. When you're ready, we'll move on to our next program.

USING INDIRECT ADDRESSING TO SPEED UP RECEPTION OF SERIAL DATA

This program (SerrxdFast14M2.bas) combines the benefits of the serrxd timeout option with the speed of indirect addressing to efficiently process large, variable-length serial data strings. Before you run SerrxdFast14M2.bas, there are a couple of points I want to mention, so you may want to print out a copy for reference. First, there's no *reconnect* statement in the program; just a *disconnect* statement near the beginning. Reconnect can't be used in this program because it loops back around to the first serrxd statement (which - as you remember includes an automatic disconnect) so quickly that the program is hardly ever scanning for a new download. Therefore, the only way to initiate a new download in situations like this is to carry out the *hard-reset* procedure mentioned earlier (i.e., disconnect the power to the processor, initiate the download, and then guickly restore the power).

SerrxdFast14M2.bas uses the

same bugfix work-around that was included in SerrxdFastSimple.bas, except now we're storing the first data byte in the storage area rather than in a GP variable. The second serrxd statement also uses the storage area to hold as many as 49 additional bytes of incoming data. The statement may look a little strange due to its unusual length, but this approach is much faster than trying to accomplish the same thing in a loop.

I assume there is some preset maximum length for a single line in the Programming Editor, but just for fun, I ran the program with twice as many @bptrinc parameters in this line and it worked perfectly. Of course, this suggests that we could use the same approach to fill a 20x4 LCD with one serial command, but that's a project for another time!

When you have read through the program and understand how it should work, download SerrxdFast14M2.bas to your 14M2 processor and try it out. If you have any problems with it, email me (Ron@JRHackett.net) and I'll try to help.

UPDATING OUR SERIAL LCD "CUSTOM CHARACTER" DRIVER

The SerrxdFastSimple.bas demonstrates all the programming techniques we need to update the LCD16x2-CustCharDriver.bas program (June '09) so that it can handle variable-length serial input without missing a character. I'm happy to say that I was able to do that without encountering any additional problems along the way. Unfortunately, there isn't enough space to discuss the updated software this month.

However, I also don't want you to have to wait another two months for the denouement of this month's project, so I'll post the updated software, along with an explanation of its features on my website (see www.jrhackett.net/LCD16.shtml).

See you next time ... NV





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You've probably seen a laser show at con-

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Laser Light Show Kit 110VAC Power Supply 1151 AC125

\$49.95 \$9.95



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The technical applications are endless. Use it to detect radiation from monitors The technical applications are endless. Use it to detect radiation from monitors and TV's, electrical discharges from appliances, RF emissions from unknown or hid-den transmitters and RF sources, and a whole lot more! If you're wondering whether your wireless project or even your cell phone is working, you can easily check for RF! A 3-position switch in the center allows you to select electric, mag-netic, or RF fields. A front panel "zero adjust" allows you to set the sensors and displays to a known clean "starting point."

If the TFM3C looks familiar, it's probably because you saw it in use on the CBS show Ghost Whisperer! It was used throughout one episode (#78, 02-27-2009) to detect the presence of ghosts!

The concept is simple, it is believed (by the believers!) that ghosts give off an elec-tric field that can be detected with the appropriate equipment. In the electric mode, the TFM3C's displays will wander away from zero even though there isn't a clear reason for it (not scientifically explainable, aka paranormal!). This would mean something has begun to give off an electric field. What it was a friendly ghost. What it will be in your house... who knows! Makes a GREAT leaning project besides! Requires 4 AA batteries.

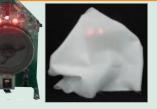


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Who says ghosts are make believe? Once your friends come upon this one they'll think differently! The unique circuit board design includes two ominous



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25 bright LED's!

- Random flash simulates flickering candle! Super bright LED illuminates entire pumpkin!
- Simple & safe 9V battery operation

The perfect "starter" kit with a terrific Halloween theme! You

transistor circuit provides a neat random flash pattern that looks just like a flickering candle. Then a super bright LED illuminates the entire pumpkin with a spooky glow!

The pumpkin face is the actual PC board, and assembly is easy through-hole soldering of all components and LED's. Your pumpkin is powered by a standard 9V battery (not included) which snaps to the back of the pumpkin. An on/off switch is also included. Create a new kind of pumpkin this year, and learn about LED's and electronics at the same time!

MK145 **Electronic Halloween Pumpkin Kit**

Portable EL Electroluminescence

- ✓ 3.3 feet long!
- Low power consumption!
 Highly visible brilliant colors

Electroluminescence illuminated flexible wire sets can be used for a lot of things but when they're battery powered they're perfect for Halloween and Trick or Treat safety concerns! Each

thin, flexible EL wire set is 3.3' long and runs on two standard AA batteries (not included). Current con-sumption is less than 100mA for long life



Mode settings include steady glow and slow or fast flash! Make it part of a bril-liantly lit, custom designed costume or simply add it for illuminated safety while Trick or Treating in the dark.

NWRR15 EL Illumination Wire Set, Red NWRG15 EL Illumination Wire Set, Green NWRB15 EL Illumination Wire Set, Blue

Spark Generating HV Plasma Generator

Generate 2" sparks to a handheld screwdriver!

Light fluorescent tubes without wires!

Build your own Plasma Balls! Generates up to 25kV @ 20 kHz from a solid state circuit!



This popular kit was conceived by one of our engineers who likes to play with things that can generate large, loud sparks, and other frightening devices! And at Halloween there's no better effect than high voltage sparks flying through the air! The PG13 Plasma Generator creates a very impressive 25,000 volts at 20 kHz, to provide a stunning display of high voltage! It will draw a cool looking 2" spark to hand held screwdriver, or light fluores-cent tubes without any connection!



It produces stunning lighting displays, drawing big sparks, to perform lots of high voltage experiments. In the picture, we took a regular clear "Decora" style light bulb and connected it to the PG13 - WOW! A storm of sparks, light tracers and plasma filled the bulb. Holding your hand on the bulb doesn't hurt a bit and you can control the discharge! It can also be used for powering other experiments; let your imagination be your guide! Operates on 16VAC/VDC for max-imum output. Can also be run from 5-16VAC/VDC to reduce the output voltage.

PG13





\$19.95

\$19.95

\$19.95



rámsey amazon.com 📻 🔂

rices, availability, and specifications are subject to change. Not responsible for typos, stupids, printer's bleed, or really weird Halloween tricks! Visit www.ramseykits.com for the latest pricing, specials, terms and conditions. hanks Robin... for reminding me of this ad deadline and that Halloween is approaching! Therefore today's colo.



WITH RUSSELL KINCAID

WHAT'S UP:

Join us as we delve into the basics of electronics as applied to every day problems, like:

In this column, I answer questions about all aspects of electronics, including computer hardware, software, circuits, electronic theory, troubleshooting, and anything else of interest to the hobbyist. Feel free to participate with your questions, comments, or suggestions. **Send all questions and comments to: Q&A@nutsvolts.com**

VFD Filament Driver



Gel Cell Charging Voltage

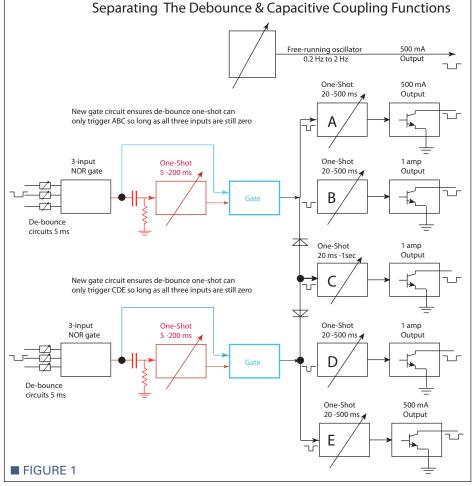
CONVERT RELAY LOGIC

I am relatively fluent in relay circuit configurations, but feel incompetent when it comes to IC design and PCB fabrication.

My goal is to replace two threerelay gate circuits with two IC counterparts. Each set of three relays has its N/O contacts connected in series to detect simultaneous occurrence of three independent phenomena.

The relays operate from separate negative-going 12V signal inputs.

My exploration thus far suggests that two of the three three-input NOR gates on a CD4025BEE4 IC will provide the starting point for the



replacement circuit.

One drawback of the simple IC gate is that it will respond to momentary coincidence of all three inputs. In contrast, the inherent inertia of the relays means that all three inputs must be present for upwards of 10 milliseconds before all three relay's contacts close.

This need to ensure that all three inputs have been present for 10 milliseconds or more adds a layer of complexity to the simple IC configuration.

An R/C de-bounce circuit ahead of each input to the NOR gate seems like a good starting point. Would an R/C de-bounce circuit with a trimpot for each input be practical?

I need each of the two threeinput gate circuits to feed its own buffer, capable of delivering 1A for additional circuit activation.

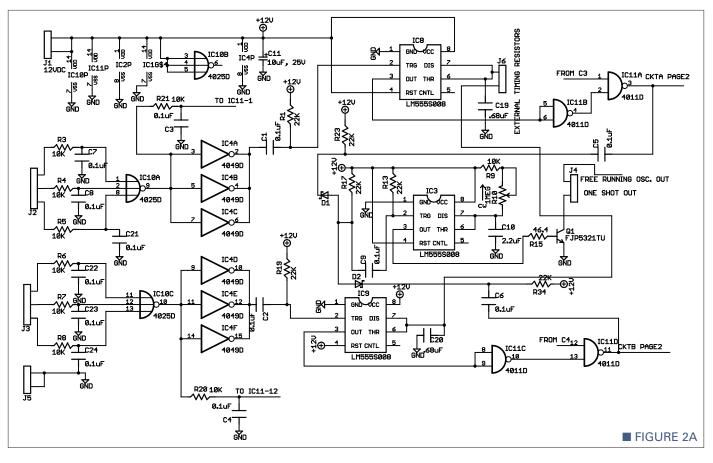
Besides driving separate 1A buffers, the two gates will also drive a common ONE SHOT that delivers an output whose duration is trimpotadjustable between 150 milliseconds and one second. (Isolation diodes will separate the two gates.)

The ONE SHOT, in turn, drives a 1A rated buffer whose output is a negative 150 ms - 1 sec 12V pulse.

That's the story in a nutshell. I hope this provides enough information to ask further questions. I'd like to up the ante and add two more identical ONE SHOTs to the PCB. Is this okay?

Another tweak. The "central" ONE SHOT drives a 24V stepper and

QUESTIONS & ANSWERS



requires a high voltage 1A output transistor.

In the revised diagram, the center ONE SHOT is triggered by the firing of either gate. In addition, two ONE SHOTs are now triggered when either of the gates is fired. The center ONE SHOT has a maximum cycle time of one second; the other four have half-second timing.

Instead of using a single resistor-capacitor circuit to provide both de-bounce and capacitive coupling, I'd like to up the ante again and separate these two functions. Can you please make these additions?

- 1. Add the two ONE SHOTs to the circuit.
- 2. Add a variable frequency oscillator 0.2 Hz to 2 Hz with potentiometer control.

Figure 1 conveys the general two-step idea. Add two new ONE SHOTs to provide the de-bounce delay, and capacitive-couple the three-input gate outputs to the two

new ONE SHOTs.

I plan to use two (ganged) 10-position selector switches to select various de-bounce timing resistor values. Can you please include terminals for selector switch connections?

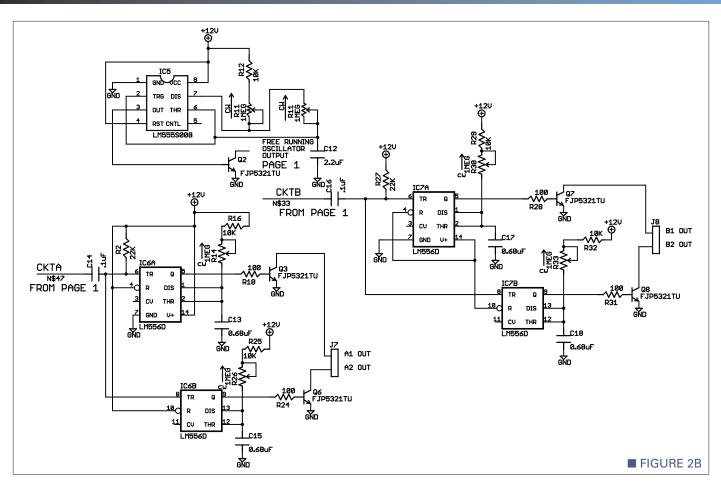
- Stanley Froud

Now that we have a final block diagram (Figure 1), take a look at the schematic before committing to hardware. The input NOR gate acts as an inverted AND. The output is inverted to trigger the ONE SHOT. The rest of the circuit (see Figure 2) is all ONE SHOTs, so not much of a challenge. I used some 556 dual ONE SHOTs for a slight savings in space and cost, but in a production environment I might choose all 555s to maximize purchasing power and minimize stocking. The use of surface-mount parts is the biggest savings because the cost of the printed circuit board is based on the area and number of holes.

VFD FILAMENT DRIVER

I enjoy the look of Vacuum Fluorescent Displays, and use them wherever I can. My challenge is driving the filament properly. It is traditionally driven with an AC source, taken from a special winding on the power supply transformer. The filament is also the cathode of the display. The transformer winding usually has a center tap which is used as the display cathode connection to which the grid and anode are biased.

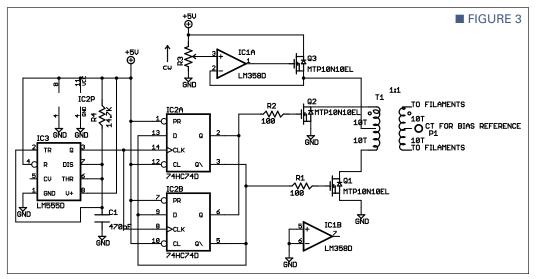
This is done to ensure even lighting of the VFD, especially in tubes with multiple segments. In this age of power bricks, etc., a transformer is often not used (as well as unavailable), and the filament is driven by a square wave with a 50% duty cycle in the pos and negative direction (essentially, a square AC). National Semiconductor made a chip to do just that — the LM9022. It is now obsolete, and has all but disappeared



from the market.

My question is this: Could you propose a small schematic to substitute for this chip? The output voltage is usually in the range of one to four volts (to be adjustable for different tubes), and current draw is around 40 to 50 mA per tube. It would be great if the circuit could drive at least six tubes (parallel filaments) for a current around 300 mA as a minimum. The input voltage could be either a fixed 5V or something between 12 and 50V, since both are usually available in the VFD design. Also, Noritake proposes to use a filament frequency for this type of drive of 10 kHz or more to prevent any flickering effect.





I started out considering a variable pulse width circuit which would allow operation from either +5 or +12 volts, but then thought about the problem of determining the RMS value such that the filament is not overheated. The RMS of a square wave is just the amplitude, so that would make it easy to get the

right value. Noritake advocates an AC center tap configuration, so that is what I will do.

In **Figure 3**, the 555 oscillator runs at 100 kHz and is divided by two in IC2. I paralleled the outputs of IC2 in order to increase the drive to the MOSFETs. IC1 and Q3 are a DC supply for the transformer. If you want 3V RMS output, just set the emitter of Q3 to 3V.

The transformer is 1:1, center-tapped. Since the output is to be 300 mA

QUESTIONS & ANSWERS

max, the magnetizing current at the input should be no more than 30 mA. I will try 10 mA which means the inductive reactance will be 5V/.01 = 500 ohms. Since the frequency is 50 kHz, the inductance has to be L = XI/2/PI/F = 1.59 mH.

Calculating the turns from N = $(L^{1}e6/Al)^{.5}$, then N = .74. That's not reasonable. I can use more turns. Twenty turns will fit in one layer on the bobbin, so use 10 turns to the center-tap, then L = 230 mH. Now, check to see if the core is big enough: WaAc for the core is given as .162. WaAc is the product of the core cross-section and the window area. The equation is:

where

k = .00528 for an E core Po = 5V*0.3A = 1.5 watt B = 1000 gaussF = 50 KHz

Then:

WaAc = .0016 and the core is much larger than needed.

The wire size can be calculated from: AWG = -4.31*ln(1.889*l/Cd), where Cd is the current density in amperes per square centimeter. The Micrometals manual recommends 400 A/sq cm.

That yields an AWG of #28; anything bigger is okay.

CURRENT SENSE FOR A DATA LOGGER

I have a DATAQ EL-USB-5 data logger. I want to use it to track when 110V appliances turn on and off; for instance, a refrigerator's compressor. I wish to use a leg of an extension cord to sense current flow.

The data logger can be set to be triggered by a contact closure or to a pre-set voltage (three volts min). Can you design a circuit that can provide a satisfactory voltage

trigger (using a 9V battery) from a current sensor (toroid)? The sensor would have to be adjustable such that the current draw from the control circuit of the appliance will not interfere with the sensing of the compressor's current draw. - Charlie Young

draw of my small freezer to

be four amps: if you use a

1,000:1 transformer, the

output current will be 4 mA; 4 mA *

1K = 4V which is enough to trigger

don't work well when the load is

the data logger. Current transformers

greater than 100 ohms, so I am using

(essentially zero ohms); see Figure 4.

I am rectifying the output because I

suspect that the data logger prefers

DC input. The gain pot can be 2K or

10K, depending on the current range

Mouser part number 553-CST-

I have a 12V flashlight with

need to replace the internal

a 12V 7AH gel battery. I

charging circuit. I will be

1020 is rated at 20 amps and costs

CHARGING VOLTAGE

using an external 12V DC supply.

recommended charging voltage

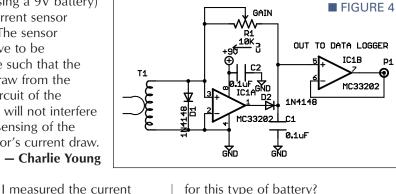
What would be the

you expect to encounter.

\$6.11.

GEL CELL

the input impedance of an op-amp

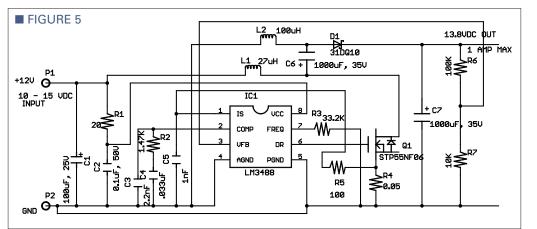


for this type of battery?

- Ken Bartone

A lead-acid type battery is fully charged at 13.8 VDC. A 10 hour charge is recommended so the charging current should be 0.7 amps. You should measure the output voltage of your power supply because it may be high enough to charge the battery without any additional circuitry. My RadioShack power supply output is 14V no load which is adequate; I would put two or three ohms in series to limit the current. If the power supply voltage is greater than 15V, you will not be able to leave the battery connected for more than 10 hours or it will overcharge (the GEL cell has some overcharge protection, but don't depend on it). Another solution would be to plug it into the cigarette lighter socket in the car. The voltage is 13.8 VDC when the motor is running.

A boost circuit is needed to get 13.8V from a 12V source; I used National Semiconductor's WebBench (www.national.com/en/webench/



SEPIC PARTS LIST						
PART	DESCRIPTION	PKG	PART #	PRICE		
C1	100 µF, 25V, ALUM	6.3MM	647-UPW1E101MED	0.28		
C2	0.1 µF, 50V, CERAMIC	5X3.5MM	81-RPER71H104K2P1A03	0.18		
C3	2.2 nF, 50V, CERAMIC	5X3.5MM	81-RPER71222K2P1A03B	0.24		
C4	.033 μF, 50V, CERAMIC	5X3.5MM	81-RPER71333K2P1A03B	0.23		
C5	1 nF, 50V, CERAMIC	5X3.5MM	81-RPER71102K2P1A03B	0.24		
C6, C7	1000 µF, 35V, ALUM	12.5x5MM	871-B41044A7108M000	0.54		
L1	27 μH, 6.4A TOROID	.86DIA	542-2100HT-270H-RC	2.61		
L2	100 µH, 4.6A TOROID	.86DIA	542-2100HT-101H-RC	2.61		
D1	3A, 100V SHOTTKEY	AXIAL	844-31DQ10	1.06		
R4	0.05 OHM, 5%, 1W	TO-220	652-PWR221T-30-R050J	2.60		
ALL OTH	ALL OTHER RESISTORS 1/8W, 1%					
IC1	LM3488 CUR. MODE	MSOP-8	LM3488QMMTR-ND	3.66		

power.html) to design the circuit of **Figure 5**. National's design used all surface-mount parts but I have listed through hole parts in the parts list (except for the LM3488 which is only available as surface-mount). The part numbers in the parts list (**Figure 6**) are Mouser except the LM3488 is from Digi-Key. The LM3488 is also available at Newark.

This is a SEPIC design which

means that the output can be higher or lower than the input. The operating frequency is around 500 kHz, so a good layout is necessary. The LM3488 is a current mode device because the inductor current is sensed by R4 and fed to pin 1 which will shuts off Q1 when the sensed voltage is 0.156. When Q1 shuts off, the energy stored in L1 is transferred to the

output through C6 and D1. L2 provides a DC path to ground. Otherwise, C6 would keep charging to higher voltage.

Voltage regulation is obtained by feeding part of the output voltage back to pin 3 which has a threshold of 1.26V. National estimates the efficiency to be 88% but they sacrificed some efficiency for size so this design should be better.



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BASIC ON BOARD



BASIC ON BOARD is a Basic interpreter that is preprogrammed on many of the ForeRunner microcontroller modules from ATRIA Technologies. BASIC ON BOARD is a derivative of the popular StickOSTM Basic developed by software engineer Richard Testardi.

BASIC ON BOARD is a perfect tool to learn programming, explore microcontrollers, design prototypes, build home projects, and even small products. To get started, all you need is a terminal emulator. BASIC ON BOARD enables access to many of the features of a microcontroller using built-in commands, pin, and register variables.

- Write text to a display: lcd 1, "Hello World."
- Read a 4x4 keypad: on keychar do gosub KYPD.
- Create a timer interrupt: on timer 1 do gosub clock.
- Write to an I²C device: i2c start 0x68, i2c write t, i2c stop.

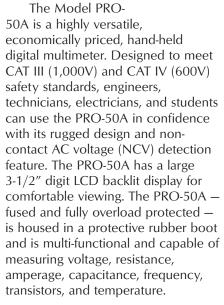
Common features such as console I/O and string handling are available. With the HD44780 interface, a character LCD can be directly controlled with a single command. A 4 x 4 keypad may be read as an interrupt. Developing programs is easy — type it in and run it. ForeRunner microcontroller modules with BASIC ON BOARD start at \$24.

For more information, contact: **ATRIA Technologies, Inc.** PO Box 896 Webster, NY 14580 Tel: **585-671-4390** Web: www.AtriaTechnologies.com

THE MODEL PRO-50A

Global Specialties introduced a new low cost, high quality, three year Digital DMM.

The Model PRO-50A has a number of features and functions not offered by any similarly priced products on the market today.



For more information, contact: **Global Specialties** Tel: **800-572-1028** Web: www.globalspecialties.com

SirMORPH IR SENSOR



M ikronauts has just released the SirMorph – a versatile IR sensor for robotics and industrial use. (SirMorph is featured in the

"Upgrading the Boe-Bot" series in the September/October issues of *SERVO Magazine*.) SirMorph is designed to be used for/as:

- Short range distance measurement (2-40 mm to 2-150 mm, depending on LED drive current).
- Servo mounted wheel encoder for Solarbotics wheels.
- Virtual bumper.
- Line sensor when mounted facing down.

A pair of SirMorphs can even be used as a combination servo mount, wheel encoder, and side virtual bumper. SirMorph can be configured for different uses and ranges simply by using different resistor values for the IR LED current limiting resistor. The product documentation includes a table of suggested resistor values with their current draw and approximate detection range. The default configuration is 11 mA at 5 VDC. SirMorph is available now both as a kit and as assembled and tested sensors. Reseller and bulk educational discounts are available.

ROBOAXE AND SCOOTER

Scooter prototype with RoboProp advanced controller and ultrasonic range sensor head.

RoboAxe DC Motor Edition.



introduces RoboAxe and

Scooter. RoboAxe is an intermediate level robot controller board based on the PICAXE 20m2 processor with an optional dual DC motor driver and two generous prototyping areas matching the layout of the optional dual 170 point solderless breadboards. RoboAxe was designed to meet the needs of the educational market, making it ideal for use in developing STEM educational courses, FIRST competitions, and it can be used as a platform when trying for the new Boy Scouts of America robotics merit badge. The nature of the resulting design also meets the need of robotics enthusiasts and hobbvists for an affordable intermediate level controller. Some of RoboAxe's many features include:

- PICAXE 20m2 microcontroller.
- Free Basic development environment and books are available from PICAXE.
- 4.00" x 3.05" inch high quality printed circuit board with mounting hole; pattern matches other Mikronauts (and Parallax) boards.
- Two large prototyping areas with matching optional solderless breadboards.
- Standard DB9F connector for serial programming.
- Optional USB-serial programming cable.
- 5V voltage regulation.

- Reverse battery voltage protection diode.
- Two solderless breadboards with 170 holes each.
 - Some of Scooter the Robot

features include:

- RoboAxe DC Motor edition.
- Two solderless breadboards. Five SirMorph sensors for virtual
- bumpers and line following.Blue Magician chassis.
- Two gear motors.
- Two yellow wheels with black tires.

- Optional ultrasonic scanning sensor head with servo.
- Optional infrared scanning sensor head with servo.

GADGETS

RoboAxe and Scooter are also available now. Reseller and bulk educational discounts are available.

For more information, contact: **Mikronauts** Email: **mikronauts@gmail.com** Web: http://mikronauts.com/sirmorph



expresspcb.com



WEB BROWSER CONTROL OF YOUR HOME

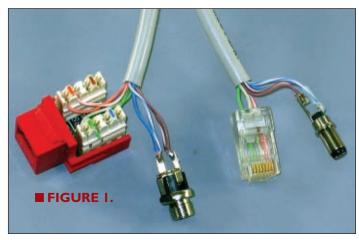
By Jürgen G. Schmidt

Last month, I described the basic hardware for implementing a thermostat with an embedded web

server so it can be controlled from a browser. This consists of a TCP/IP Base board using a Microchip 16-bit PIC24 processor to provide a general-purpose web server, with a daughterboard that implements the thermostat hardware and interface to the home heating, ventilation, and air conditioning (HVAC) equipment. In this article, I will describe the physical connections, some construction details, and the software that makes it all work.

Network, Power, and HVAC Connections

While I had the foresight to install CAT-5 cable to the thermostats when I built my house, I did not run any power lines. I assumed at the time that I would get power from the thermostat itself. Alas, while some thermostats have a fifth wire to provide power, mine did not. Instead, power for the new thermostat is provided via the unused pairs of the CAT-5 networking cable. Only pins 1, 2, 3, and 6 are used for the network connection. If you look at the Wikipedia article on CAT-5 wiring (see **References**), you will see there are two ways to arrange the wires in the connectors. I used the T568B arrangement, since that



matched my existing cables. The Wikipedia article on Power over Ethernet shows which wires to use for the power connection. The table at the end of that article shows that pins 4 and 5 are used for the positive terminal and pins 7 and 8 are used for the negative terminal.

Part 2

When I made my cables, I separated the wires and fed only the network-related pairs into the RJ45 connectors and soldered the other pairs to coaxial 2.1 mm power connectors. To verify that I had everything figured out correctly, I made a short test cable, the two ends of which are shown in **Figure 1**.

The red connector mounts on my patch panel, from which it is connected via a jumper to my network switch. A regulated five volt power supply plugs into the 2.1 mm jack. The male RJ-45 connector and the 2.1 mm plug connect to the TCP/IP base board. The live thermostat end of this can be seen in **Figure 2.**

The green connector coming out of the wall is attached to the HVAC control cable. This is a Phoenix pluggable terminal block that plugs onto a header on the thermostat board (**J1** on the circuit diagram). These connectors are handy because they can be oriented three different ways when plugged into the header and are convenient for testing and installation. If you mount screw terminals directly on the thermostat board, you have to juggle the thermostat, screwdriver, and cable while you attach and detach the wires. Instead, you attach the terminal block directly to the cable and then plug and unplug as needed.

Construction Issues

Eventually, the thermostat would be mounted in an enclosure and attached to the wall in place of the original thermostat. To simplify the overall assembly, I wanted to limit the connections to the circuit boards to power, network, and HVAC control. All the switches, the display, and connectors should be on the circuit boards. Finding the right switches was a challenge, along with mounting the LCD so that everything could be mounted at the right height behind and through the faceplate. The connectors between the circuit boards needed to be the right length to leave just enough clearance for the MagJack.

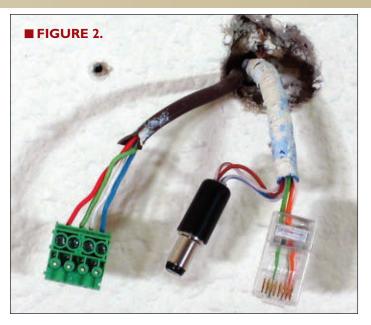
Plastic enclosures do not come in an infinite array of sizes and colors. The BUD CU-389 enclosure was just the right size but only available in black. After drilling the ventilation holes in the base, I spray-painted it off-white to match the walls. I custom-made my own faceplates from some scrap plastic. You can see these in **Figure 3**. The translucent gray cover is made from acrylic and the almond colored cover is made from an old ABS plastic enclosure. These were made on a ZenBot CNC router. The routed lettering is filled in with oil pastel.

Some of the components for this project are only available in surface-mount (SMD) format. That and size limitations dictated a SMD implementation for all but the connectors, LCD, relays, and switches. Some of the connectors are available in SMD format, but I prefer to use the through-hole version for mechanical sturdiness. This was my largest SMD project to date. I've used a finetipped soldering iron in the past, but for this project I decided to invest in a compressed air-powered solder paste dispenser and got a convection toaster oven to use as a reflow oven.

Initially, I used a thermocouple attached to my multimeter to monitor the oven temperature, but now I do so much surface-mount work that I built an automated reflow oven controller, based on the thermostat daughterboard. It provides the display, pushbuttons, and relay driver. (See the **sidebar**: Flexible Designs.) My website – **www.jgscraft.com** – has additional information about my surface-mount soldering process and how these boards were made.

Everything worked well, including soldering some of the bypass capacitors on the underside of the TCP/IP board. I moved these to the back to reduce clutter on the top of the PCB. As the flux heats up, there is enough surface tension to hold the parts in place, so only a single baking cycle is needed. I used 805-sized components for the resistors, LEDs, and most of the capacitors. The bypass 0.1 µf capacitors are 603s.

After baking the boards, the only cleanup I needed to do was around the PIC 44-pin TQFP part. There were some solder bridges that were easy to fix by running a fine-tipped soldering iron between the leads. The corollary to solder bridges is unsoldered pins which again can be fixed with a fine-tipped soldering iron and .025 inch solder. I test all the connections on the multi-

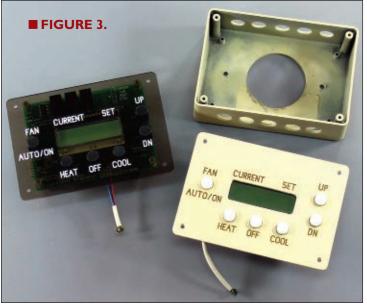


pin parts for shorts and opens just to avoid headaches later on.

For the fine-pitched parts, there is a fine line between applying enough solder paste to avoid dry connections and too much that results in bridged connections. I also found that some parts had slightly bent pins that did not make contact with their pads, so now I check the parts on the flat side of an old CPU heatsink to be sure all the pins touch the surface.

Software

Developing the software for the thermostat requires Microchip's C30 compiler which is now called "MPLAB C Compiler for PIC24 MCUs" and the TCP/IP stack version 5.25 (or later) which is part of the Microchip Application Libraries. You also need the MPLAB IDE to manage the project and its many files. The student or demo version of



Flexible Designs

The initial design work for the browser-controlled thermostat started out as a single, monolithic circuit and board design. Space limitations suggested a double-decker

design which led to a separation of functions. This ultimately led to the design of two boards which could be applied to much more than just a thermostat application.

I isolated the hardware for the network connection and embedded webserver on one board and added some additional support circuitry to make it more flexible. The original design using a 28-pin processor had just enough I/O pins to support the thermostat application. Switching to a 44-pin part allows flexibility with other applications. I was able to develop the TCP/IP stack, bootloader, and webserver software without needing the thermostat board. All the unused I/O lines, power, and UART2 connections are available on PCB pads that will support standard 0.1" pitch headers.

The thermostat interface board has an LCD, some pushbuttons, relay driver circuitry, and some relays. Many of the projects I have seen in *Nuts & Volts* have similar interface requirements. The board does have dedicated circuitry for

the C30 compiler works fine, so all your software development tools are free. The TCP/IP stack is fairly easy to use, provided you know the basics of the TCP/IP protocols and you read the documentation for the stack. Studying the source code and posts on the Microchip forums related to Ethernet and the processor family you are using is also a good idea. The forums are well supported and provide a wealth of information that might be hard to find elsewhere.

The first step in using the stack is to take one of Microchip's sample applications as a starting point and modify it for your hardware. This is usually the "TCP/IP Demo App" and is referred to regularly in forum postings. This implements a webserver, as well as most of the common TCP/IP components such as UDP, TCP, Client/Server, email, DHCP, and more. The Demo App is configured to run on Microchip's evaluation and demonstration boards but modifications to the hardware configuration files will adapt it to run with your own TCP/IP hardware or – in this case – the thermostat project. The TCP/IP base board comes with detailed instructions for configuring Microchip's TCP/IP stack version 5.31 to run the Demo App.

This is not my first embedded TCP/IP project and I don't recommend anyone start a project like this from scratch. Get a development kit from Microchip or other hardware that comes with working software and start experimenting from there. I started out working with TCP/IP on a small development board, a sample application, and a series of tutorials that were known to work. Over time, my software and hardware evolved from this in gradual steps. If something failed, I could go back a step to find out if the software or the hardware was at

attaching the temperature and humidity sensors. However, these – as well as all the other connections – are just links via headers to I/O lines on the processor – they could be connected to anything.

There are lots of projects in my backlog that need a browser interface to hardware systems. The TCP/IP base

board will work well for those. Other projects require a display and pushbutton interface. I have already repurposed one of the thermostat daughterboards as a reflow oven controller. Instead of the twoline LCD, I used a three-line version from the same manufacturer. I only needed one heavy-duty relay and three of the pushbuttons. The headers attach to a perfboard that mounts the PIC16F690 controller I used for this, a thermocouple interface, and power regulator. All of this I mounted in the same enclosure I used for the thermostat. I just deleted some of the holes from the existing CNC router program to give me the right faceplate. The results can be seen in the photos.

The added benefit of flexible designs is that I make use of the extra PCBs I end up with. Some boards I order have design errors, but if the design is flexible, they can still be used for other projects. Even for the final boards there is usually a minimum area per order, resulting in extra boards. Ideally I would sell them, but if I can't, I will eventually find another use for them.

fault. One incorrectly assigned pin in the software or one loose solder joint is all it takes to hang up the system, and it's hard to tell if hardware or software is the culprit. So, before adding my own code, I make sure any new hardware will work with the demo application.

The files that typically need to be modified are HardwareProfile.h and TCPIPConfig.h. For the TCP/IP base board, I also needed to modify ENC28J60.c to enable the clockout signal that drives the CPU clock. The main.c file contains some hardware initialization routines that need to be modified to match the hardware, as well. Once the board tests okay for the basic TCP/IP applications, I can add support modules for any additional hardware. Then, I add modules to main.c for my specific application.

The HVAC control software consists of a loop that: 1) Pauses a while and checks the system clock to see if certain timeout events have occurred; 2) Reads the sensor(s) and updates variables with the results; 3) Checks if a button has been pressed and updates variables corresponding to the thermostat settings; 4) Checks to see if any of the relays need to be turned on or off; and 5) Updates the display. There is no rocket science here, but there is a control detail that needs attention.

If a heater is set to turn on when the temperature gets below 60 degrees F, it will do so when the sensor is at that temperature. However, once the sensor is at 60 or above again, the heater will turn off and when the temperature falls again, the heater will turn on. This can result in the heater trying to turn on and off rapidly as the temperature fluctuates. To avoid this, we include a deadband or hysteresis into the control loop.

Once the heater turns on, it will not turn off again until the heat is a certain amount, maybe one degree



```
switch( myAppState )
  case smMY_IDLE:
                                   //----- don't work too hard ------
         check if we need to turn off backlight
    if( TickGet() - ticks3 >= TICK_SECOND * 5uL ) LEDB_OFF();
    //-- check if we need to turn off alternate LCD display
if( TickGet() - ticks4 >= TICK_SECOND * 3uL ) showLCD = LCD_MAIN;
    if( TickGet() - ticks1 >= TICK_SECOND/20u1 )
                                                                   LISTING 2.
      ticks1 = TickGet();
      myAppState = smMY_STEP1;
    break;
  case smMY STEP1:
                              //----- get temp and humidity ------
    //-- get temp and humdity periodically
if( TickGet() - ticks2 >= TICK MINUTE/12ul )
      ticks2 = TickGet();
       sht_rd(&tempC, &humidity);
                                           // SHT11 libraru
    humidity = (int)(humidity);
                                           // round to integer value
    tempF = (int)(tempC * 9 / 5 + 32);
myAppState = smMY_STEP2;
    break;
  case smMY STEP2:
                             //----- check for button press -----
    myAppState = smMY_STEP3;
    break:
                              //----- check if we need to turn anything on or off
  case smMY STEP3:
    myAppState = smMY_STEP9;
    break;
  case smMY_STEP9:
                              //----- check for any configuration changes and save them
    myAppState = smMY_IDLE;
    hreak
} //switch( myAppState )
```

higher than the set point; in this case, 61 degrees. This temperature difference is the dead-band in which no changes will take place. In the thermostat software, there is a dead-band of two degrees Fahrenheit. The heater will turn on at one degree below the set point and will turn off at one degree above the set point. You can see the details of this in the thermostat.c file that is included with this article's downloads.

Multitasking

For a webserver to work properly, the TCP/IP software needs to monitor the network connection for incoming traffic. While the incoming data is processed, it still needs to continue checking for incoming data so none is lost. In order to accomplish this, the TCP/IP stack is implemented as a series of nested loops. Each of these loops has multiple steps in it. Each step performs a little bit of work, exits the loop to let the outer loop do a little work, and then the inner loop resumes with the next little step, exits, and so on. This is called "cooperative multitasking" and is described in detail in the TCP/IP stack documentation. This is important in the context of the thermostat application because it must also follow this pattern. Listing 1 shows the case statement that runs the TCP/IP stack in the demo application. This is running in an infinite while() loop. The blue code highlights where the user application, MyTasks(), has been added. This gets called every fifth time through the loop.

The MyTasks() function in the file thermostat.c implements the thermostat as described above and is

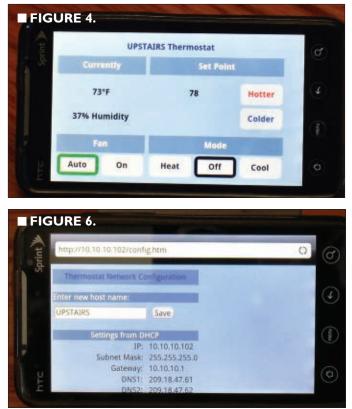
```
while(1)
    switch( smSysState )
      £
        //-----
        //----- begin of main TCP/IP processing loop
        case RUN IDLE:
           //-- toggle heartbeat LED
           if( TickGet() - t1 >= TICK_SECOND/2ul )
                                                             LISTING I.
             t1 = TickGet();
            LED1_IO ^= 1;
          3
           smSysState = RUN_NET_TASKS;
          break:
        case RUN NET TASKS:
          StackTask();
smSysState = RUN_NET_APPS;
                                         // HTTPServer runs here as another state machine
          break;
        case RUN NET APPS:
          StackApplications()
          smSusState = RUN MY APPS:
           break;
        case RUN_MY_APPS:
          MuTasks():
           smSysState = RUN UPDATES;
          break;
        case RUN UPDATES:
           if( AppConfig.MyIPAddr.Val != eth0_ip_addr )
             eth0_ip_addr = AppConfig.MyIPAddr.Val;
IPAddressIoString( &AppConfig.NuIPAddr, buffer);
printf("New IP Address = *s\n\r", buffer);
#if defined(STACK_USE_ANNOUNCE)
             AnnounceIP():
#endif
           smSusState = RUN IDLE:
           break;
      //----- end of main processing loop ------
      case SYS ERROR:
        printf("ERROR STATE\n\r");
         while(1);
      case SYS RESET:
        printf("%s System Reset\n\r", PRODUCT_NAME);
        Reset():
    } //switch( smSysState )
  ₹//while(1)
```

shown in abbreviated form in **Listing 2**. Whenever it is called, the state variable, myAppState, is checked and the corresponding portion of the case statement is processed. Then, the state variable is updated with the next value and the function exits to allow the main loop to do some work. Five iterations of the main loop later, we're back in MyTasks() to process the next section of code, and so on. The application programmer is responsible for making sure nothing is interrupted for long – or worse – blocked while waiting for something to happen such as a button press or serial communication.

The Embedded WebServer

The usual software process for microcontroller development involves compiling a program, loading the HEX file, and testing the result. With an embedded webserver, there are some additional steps. You need to develop the web pages you want displayed and then you need to load them onto the server.

Web pages for the TCP/IP stack are created just like any other website. You can include pictures, JavaScript, CSS, and any other files that your target browser will support. The webserver stores the files in the serial Flash memory, up to four megabytes worth. The primary job of the webserver is to just deliver the files to the browser. Dynamic variables in the web files will be filled in by the



webserver before the files are sent to the browser. This is fine for relatively static content, but for an interactive application such as a thermostat, we need something more. We want to be able to see the temperature change without having to hit the refresh button on the browser. This is accomplished via an AJAX interface.

AJAX is basically a library of JavaScript functions that leverage certain browser features for interactive applications. The AJAX routines communicate with the corresponding functions in CustomHTTPApp.c which is part of the TCP/IP stack. A detailed explanation of how this works would take at least another article on its own. For now, you'll have to settle for reading about it in the documentation and sample code from Microchip. The JavaScript code runs in the user's browser, periodically sending updates from the web page to the webserver and retrieving data from the webserver for updating the browser display.



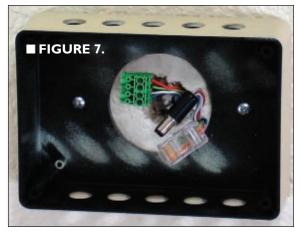
For example, the thermostat web page has a button for lowering the set point. Clicking this button activates a piece of lavaScript code that sends a message to the server that the decrease button has been clicked. On the server, the set point value is decremented and transmitted back to the web browser so the new set point can be displayed.

Another piece of code updates the LCD display. Even if there aren't any userinitiated changes on the web page, the JavaScript code periodically queries

the webserver for updates so that the web page is kept current. That will happen if the decrease pushbutton on the physical thermostat is pressed. The set point value is decremented and displayed on the LCD. The JavaScript code will retrieve the updated value for display on the web page.

The web page design can be changed and uploaded to the webserver without making any changes to the application code. The only time you would need to change the application and recompile it is if you want to exchange more data between the web page and the application. If you don't like the way the thermostat web page is laid out, change it and upload it using the MPFS2.EXE utility. You can even add additional pages, as long as they fit into the memory.

Figure 4 shows the default thermostat page. **Figure 5** shows an alternative page that is accessed by changing the web address. **Figure 6** shows a configuration page that



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changes the host name for the thermostat. Most computer-based browsers allow you to access the web page by the host name. Unfortunately, the browser on my EVO doesn't do that and I have to type in the IP address and save it as a bookmark.



The hardest part of



the installation is probably arranging for a network cable to come out where you plan to mount the physical thermostat. In most cases, that should be a matter of dropping the cable down inside the wall from the attic or feeding it up from the basement or crawl space. Attach the connectors as described in Wikipedia articles and here.

You do need a DHCP server on your network to provide the thermostat with an IP address. Most cable or DSL routers provide this by default. Disconnect the old thermostat from the wall and identify the function of the wires. The diagram at http://wiki.xtronics .com/index.php/

Image:Thermostat.gif should help you in identifying them if it's not obvious. Connect the thermostat wires to the Phoenix terminal block in the sequence shown on the thermostat PCB. For now, connect only the network cable and power. The heartbeat LED should blink once per second and the LCD should display the current status.

Test the various buttons to be sure they change something. If you did not connect a terminal to see the debug output, you can press the "UP" and "DN" buttons at the same time to display the IP address and host name. If all this is working, connect the

thermostat connector, making sure it is oriented correctly.

Sequencing the final assembly of the pieces on the wall was a puzzle in itself. The enclosure needs to be attached to the wall. Wires need to be plugged in. The PCB assembly needs to be mounted in the enclosure and then the faceplate goes on. After playing with some mockups, I decided to attach the thermostat daughterboard to the faceplate with some standoffs. The TCP/IP base is socketed to the daughterboard, held firmly in place by the four eight-pin headers.

After plugging in all the connectors, you can attach the faceplate to the enclosure that has already been attached to the wall. It's a tight fit so you need to be sure you can push any excess wire into the wall. **Figures 7**, **8**, and **9** show this sequence. You are now ready to control your thermostat from your recliner. **NV**

Thermostat Circuit

http://wiki.xtronics.com/index.php/

Image:Thermostat.gif

Brush Electronics

http://brushelectronics.com/

Microchip Application Libraries www.microchip.com/mal

Microchip Bootloader

Application Note

http://ww1.microchip.com/down

loads/en/AppNotes/01094a.pdf

REFERENCES

CAT-5 Wiring http://en.wikipedia.org/wiki/Cat-5

Power Over Ethernet http://en.wikipedia.org/wiki/ Power_over_Ethernet

ZenBot CNC Router

Hysteresis Explained http://wiki.xtronics.com/index.php/ Dead-band_hysteresis

AJAX http://en.wikipedia.org/wiki/Ajax_%2 8programming%29

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BUILD THE LED CENTERPIECE GRRSTMAS TREE

Since the time candles were first used for decorating Christmas trees, a well-lit tree has always been an important part of the holiday season. The Christmas tree described here makes a great table centerpiece, or with a simple modification can be used in a front yard display that your neighbors will envy! If you have read the previous articles in *Nuts & Volts*, you have been introduced to the Spectrum-ACE family of single board controllers with the ALEC® operating system. (If not, check out the March '11, April '11, and July '11 issues because this article specifies these platforms exclusively.)

To help get into the holiday spirit, this will be a project that everyone can enjoy. Using either the Spectrum-Lite or the Spectrum-ACE 2a, we are going to build our Christmas tree from LEDs. Once that's done, we will show you how to easily modify the design to run traditional 110 volt AC strings of lights. This little tree has 20 different routines controlling 256 LEDs and will bring non-stop enjoyment to your holiday season. First up, let's talk about the control hardware for the tree.

To make it easy for everyone to build, the main hardware can go in one of two directions. The first version of the hardware is the simplest way. Using the Spectrum-

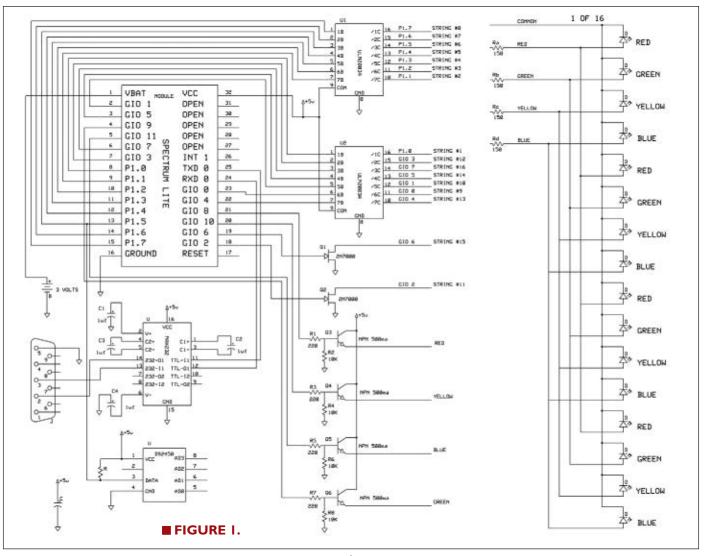


Lite CPU with a small evaluation PCB (printed circuit board), all we add are two ULN2003a seven-channel low side N channel FET drivers, two 2N7000 N channel FETS, four NPN transistors, and a handful of resistors.

As seen in **Photo 1**, the hardware is very straightforward and can easily be built-up in an hour or two. During testing, a simple hand-wired board was added with a few LEDs to display what the software is doing. The **Parts List** for the Spectrum-Lite version is shown later in the article.

So, let's get down to it and have some fun. First, let's take a closer look at the schematic shown in **Figure 1**.

The Spectrum-Lite has one dedicated UART, a real time clock, and 20 general-purpose I/O (GPIO) lines, which is great because for our Christmas tree we need – you guessed it – 20 I/Olines. As can be seen in the schematic, 16 I/O lines are used to drive the 16 FETs for 16 common lines of the tree and four I/O lines are used for color selection on each of the 16 strings that make up the tree. The proto area on the small evaluation board has plenty of area to support all the circuitry and more.

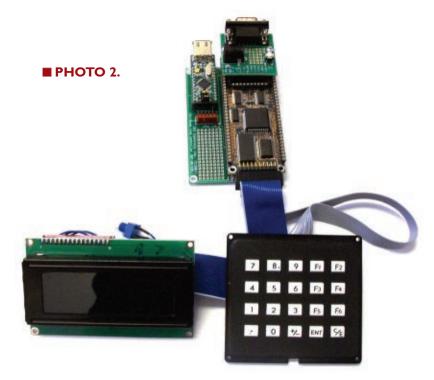


The evaluation board also has a DS2450 A/D (analogto-digital) converter open if the user wants to take advantage of it. Port 1 of the CPU is used to control strings 1 through 8 (spokes 0-7). The lower eight GPIO lines of the real time clock control strings 9 through 16 (spokes 8-15), with the last four GPIO lines controlling the color of the tree. The control of the strings of lights is done in a simple matrix type of format. To turn on a string of lights, select the appropriate I/O line for the string, set it high, select the color control I/O line, set it high, and away we go! Once you have your hardware up and running, it is very easy to write your own routines and add to the library of routines in the main program.

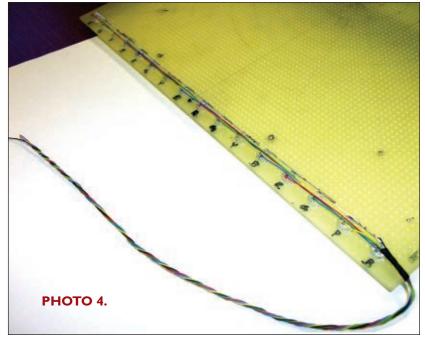
The second version of the hardware uses the Spectrum-ACE 2a with the addition of a 20-key keypad (the keypad shown here was just one out of the junk box), a four line 20 character LCD, a USB drive, and the option for a Maxim[™] 7301 I/O expander for even more lighting control. Even with these additions, this is still a project that can easily be up and running for this holiday season. At the core of both designs, the control of the tree is the same. This way, you can start with a simple design to get it up and running, and then go all out and really impress friends and family. For a complete schematic of the expanded hardware version, you can visit our website at **www.i2icontrols.com** and download it. It is also included

PARTS LIST

OTV	ITEM		
	Spectrum-Lite CPU module		
1			
	Small evaluation board for Spectrum-Lite		
2 2	ULN2003A Darlington transistor arrays		
2	2N7000 N channel FETs		
4	NPN transistors (any NPN will work, but must be		
	good for 500 ma of current)		
4	220 ohm 1/8 watt resistors		
4	10K ohm 1/8 watt resistors		
64	150 ohm 1/8 watt resistors		
64	Red LEDs		
64	Green LEDs		
64	Yellow LEDs		
64	Blue LEDs		
8	36 inch lengths of 1/16 inch brass rod		
160 inch	160 inches of 1/16 inch black heat shrink		
36 feet of multi-color four-conductor wire with ground			
For 110 volt AC operation, the I2I Controls four-channel			
AC switch is available.			
AC SWIL			







at the article link shown here.

With the second version of the hardware and software, the user can control morning and evening timers, and program the running sequence and speed of the routines. In addition to these programming features, the user can also test the different strings on the tree, as well as the colors. As seen in **Photo 2**, the Spectrum-ACE 2a, interface board, LCD, and keypad all plug together very easily, and then can go into whatever box the user wants. The last sub-assembly for the tree is a junction board for plugging in the 16 plugs from the 16 strings of LEDs. For the first prototype, this board was handwired, but a quick-turn PCB was much nicer for the finished project.

BVILDING THE TREE

Construction is pretty simple, but it does take some time. I built a simple jig from a piece of 12" x 12" perforated board for making up each string. Once I got the procedure down, a string could be built and tested in about 20 minutes. In **Photo 3**, the fixture is shown with a string being built. **Photo 4** is a completed string. Each string is made up of a total of 16 LEDs: four red, four green, four yellow, and four blue. I chose to use clear LEDs so when the tree was off, it would appear to just be a centerpiece on the table. I used fivepin Molex connectors on the end of each string.

Once all the strings are built and tested, the next step is to build the framework for the tree. The framework was made from 1/16 inch brass rod. The top of the tree is made from a small round piece of PCB material with 16 small holes for the rods to go into and then be soldered into place. (The same piece of perforated board used for the jig is used to build the frame of the tree.) Once the frame is built, heat shrink tubing is put over each piece of the brass rods to insulate the strings from the rods. Then, with the frame built, a final piece of brass rod is soldered completely around the base to help support the frame.

The next step is to mount each string on the frame. This is done with a combination of hot glue and small zip-ties. Once the tree is completely assembled, all that is needed is to load the software and you are good to go. **Photo 5** shows a completed tree. The final touch will be to add garland around it to give a finished look. The operating tree only consumes about three to four watts, depending on how much current you run through the LEDs. Once your tree is up and running, set back and enjoy the show!

WHAT'S NEXT?

Now that you have a great little centerpiece for the table, let's think bigger — you are going to enter the neighborhood lighting contest and you are driven to win! Well, this little tree can be a great start. All that is needed is to convert the drive signals from our little table-top tree to 110 volts AC (please use extreme caution in doing this!!). The easiest way is to use an SSR (solid-state relay). There are a number of companies that make SSRs. You can pick them up as surplus, make them yourself, or buy them new off the shelf. I decided to make them myself.

I knew that each string would need four channels and that there would be 16 strings. **Photo 6** is a prototype of the four-channel SSR, complete with heatsink and fuses. This SSR module is good for 400 watts per channel or 1,600 watts total. With this much power, you can control 25,600 watts total for 16 strings of lights. I wanted to see how big of a tree I could make but my power panel had other ideas, so I am limited to about 100 watts total.

The next step is to use the small junction board from the tree and replace the 150 ohm resistors with solder bridges across where the resistors go. Then, plug in one SSR module to each plug and you now have a beacon to guide an incoming sleigh and reindeer!

HEY, WHAT ABOUT THE SOFTWARE!

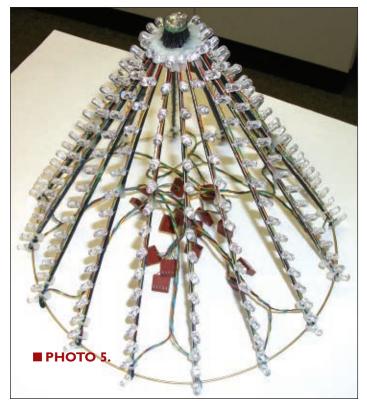
Okay, so now that we have covered the hardware, let's talk software.

The first bits (or bytes) of knowledge you need is what to get and where to get it. We can't print the source code for the entire project because of space limitations, so it's available on our website, along with all the workings of the bigger 2a project which includes a 4x20 character LCD and a 20-key keypad for controlling the tree while it's in operation. (The source code is also available at the article link.) That's way cool when running outdoors!

What can be done is a general overview of the source code modules and how to install them into your CPU board, some in-depth examination of what is in the functions that drive the project, and several prime examples of the routines that light the tree. The modules are not that large, and are fairly well documented. We'll also talk more about the programming environment that makes ALEC® a really great RAD (Rapid Application Development) tool.

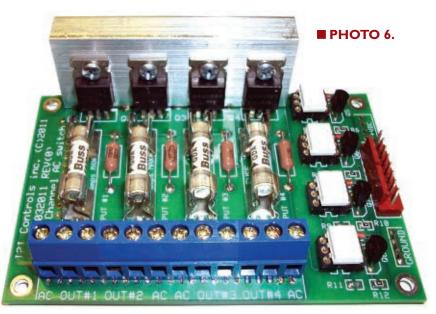
To start, download the source code and unzip it to a working directory. There are three files that make up the Spectrum Lite version in ALEC®:

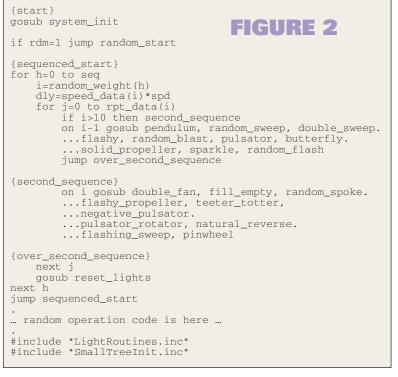
- 1. SmallTree.bas ~4 Kb
- 2. SmallTreeInit.inc ~4 Kb
- 3. LightRoutines.inc ~16 Kb



The SmallTree application is only about 560 lines of onboard, microcompiler-ready code after compacting during the load-in using the Lexical Preprocessor in our Spectrum PortMaster terminal interface (a free application available on our website). The file, SmallTree.bas.lst (~21 kb), along with a reference file, SmallTree.bas.ref (~22 kb), are created during this load-in process, but are supplied for those using other terminal software.

To load the app, reset the CPU and initiate the conversation with ALEC® by pressing the space bar. Then, drag-n-drop (or send via the file dialog menu) the SmallTree.bas file (please note the common control settings provided in this file). The .lst, .ref, and .err files are re-created during the Lexical Preprocess and the app is





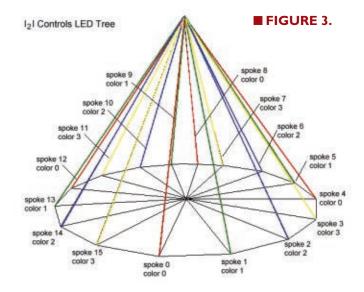


FIGURE 5 {sparkle} SPARKLE (#9) output "SPARKLE (#9)" for m=3 to -2 step -1 rtc(c)=color_data(abs(m)) cb=1 : cc=1 for p=1 to 11 step 2 for e=1 to nat_rev(p) port1=nat_revb(cb) cb++ if cb>15 then cb=0 for t=1 to dly : next t next e for e=1 to nat_rev(p+1) rtc(a) =nat_revc(cc) CC++ if cc>15 then cc=0 for t=1 to dly : next t next e next p next m return

{random_blast} RANDOM BLAST (#5) output "RANDOM BLAST (#5)" rtc(c)=int((1000*rnd)/4)+1 p=int((1000*rnd)/4) : port1=p : rtc(a)=p for t=1 to dly : next t return ; *** end random_blast (#5) *** FIGURE 4

loaded. Press the Enter key for a new prompt. Type Prog at the prompt to compile and save to Flash. Type Exe to run SmallTree.

Loading the program using HyperTerminal or other serial communications software, reset the CPU and initiate the conversation with ALEC by pressing the space bar. Then, send the file SmallTree.bas.lst to the CPU. Press the Enter key for a new prompt. Type Prog at the prompt to compile and save to Flash. Type Exe to run SmallTree.

That was easy. Now, open the SmallTree.bas file in your favorite editor. The first half of the file is comment data, but read it anyway. Find the {start} label. This is where the fun begins. If you haven't already done so, please download the ALEC manual. It will be helpful along the way.

The labels for jump points and subroutines are enclosed in curly braces {}. So, the first call is to system_init, which is in the include file SmallTreeInit.inc as shown at the bottom of **Figure 2**. This sets up the hardware and loads some useful data. The next line checks for the start-up type, either running the light effects in random or sequential order (see the include file). The next label encountered is {sequenced_start} and 22 lines down is the jump back; contained between is the code for running the lighting routines in sequential order.

For-Next loops are common, and the variable named 'h' was chosen at random. Being a primary control loop variable, h is not re-used elsewhere. The loop is 0 to 19, as referenced by variable 'seq' in the include file.

Now that all the ground work is done, let's get on with it and set the effects in motion! The variable 'j' is another loop control but is within the primary loop. It is used to control how many repetitions each routine gets before moving to the next. This data is in the rpt_data array. This array is a copy of the original array data read in, repeat_in, and may be modified on the fly, referenced, and then restored from the original data.

The **ON – GOSUB** instruction in the j loop provides a simple way to call various subroutines, depending on the value of the expression (see the ALEC manual). The expression basically evaluates to an index into the list of routines following the GOSUB.

Reset the lights by calling the subroutine reset_lights, and do it all over again. That's it for sequential operation. Random operation simply calls the **RND** function (see the manual), changes it to an integer value, and substitutes the outcome to the variable 'i'.

Now, let's examine a couple of routines that light up the tree. Remember that the tree can be from a desktop to rooftop in size, but the controls are the same: Colors 0-3 are the real time clock I/O at 0x121h, or RTC(C). Spokes 0-7 are on PORT1 and Spokes 8-15 are at 0x120h, or RTC(A). Refer to **Figure 3**.

Open LightRoutines.inc in your editor. Let's start with the most simple routine: RANDOM BLAST, routine #5. This routine simply takes two random numbers and applies them to the color and spokes. Since a random number from ALEC can be anywhere from zero to one (ex: .709514), it must be upgraded to integers to be useful.

In Figure 4, the formula rtc(c) = int((1000*rnd)/4)+ 1 is for the color. An example result is: rtc(c) = int(

 $(1000 \times .709514) / 4) + 1$, or int((709.514 / 4) + 1, or 178 decimal (0xB2h). Since the color range is zero to three and the register RTC(C) is a nibble, it evaluates to two and the rest of the data is superfluous. The spokes formula is similar – **int((1000*rnd) /4)** – and is easily applied to both control ports using a trash variable 'p.' A time delay is added so that you actually see the effect, and the routine is repeated according to the data in the repeat routine data array, rpt_data(i) with i=4.

The next routine we'll examine is a fairly complicated piece of code that we call SPARKLE, routine #9. It steps through the individual colors while cycling the spoke controls from off to fully on and then off again. Outdoors, the kids really love it! As explained before, the color and spoke controls are eight-bit values either as I/O registers in the real time clock or port 1 of the CPU. So, the intent is to set up cycling color loops, and then turn the various spokes within on and off.

The routine starts with the color loop, using a FOR-NEXT loop from 3 to -2 with a step of -1, using the variable 'm' (see **Figure 5**). The color data is applied by using the absolute value function (ABS) within ALEC. This ensures an eight-bit value is written to the control. Two trash index variables ('cb' and 'cc') are initialized with the value of zero. The FOR-NEXT loop using the variable p reads the natural reverse array ('nat_rev') using a step increment of two.

Embedded within this loop are two more loops (using the variable 'e'): one that sets the control for the lower eight spokes, and the other is for the upper eight spokes with an index value of +1. These reference data from the second and third natural reverse arrays 'b' and 'c' ('nat_revb' and 'nat_revc'), and are applied to the controls at port 1 and rtc(a), respectively. Both rash index variables (cb and cc) are incremented after use to keep indexes current. Standard delays are applied as each control is updated.

Well, that's all we have room for here. Additional help for all the routines, plus programming tips and tricks are available on the I2I Controls website. When you visit, please sign the guestbook or register at the forum.



A FLASHER, GATE, AND BELL FOR YOUR MODEL RALLROADEL 537

The intersection of a train track with a road provides a natural focal point of interest. Why not take advantage of it with a crossing signal that flashes and a gate that lowers when a train passes? This simple PIC microcontroller circuit even enhances the action with a realistic bell sound!

To activate a signal, you need a way to detect approaching trains. Model railroaders have developed several methods over the years. For larger, heavier trains various configurations of mechanical switches can be used. However, most train layouts rely on more complicated methods, such as photodetectors or circuits that sense the current drawn by an engine in an electrically isolated section of track. The method described here uses magnetic reed switches. They're simple, inexpensive, and easy to install and conceal. Their one drawback might be that they require magnets mounted on the engine and caboose (or other tail-end car).

When an engine approaches a road crossing, its magnet closes the contacts of a reed switch hidden in the track. This sends a signal to a PIC that alternately flashes a pair of LEDs, pulses a servo motor that slowly lowers a gate, and sends digitized samples of a ringing bell to a speaker. The action continues until a second reed switch buried at the road crossing detects the passing of the caboose.

Circuit Board

The schematic and board layout in **Figures 1** and **2** reveal the simple circuit. It's basically a PIC with a bunch of I/O connectors and a power supply. A PIC16F684 was chosen because it's inexpensive, comes in an easy to handle package (DIP), and has support for pulse-width modulation (PWM) which is used for the bell sound. The circuit will drive two sets of flashers and two servos, so automobile traffic is protected in both directions. Of course, you can use any combination of the outputs. You might, for instance,

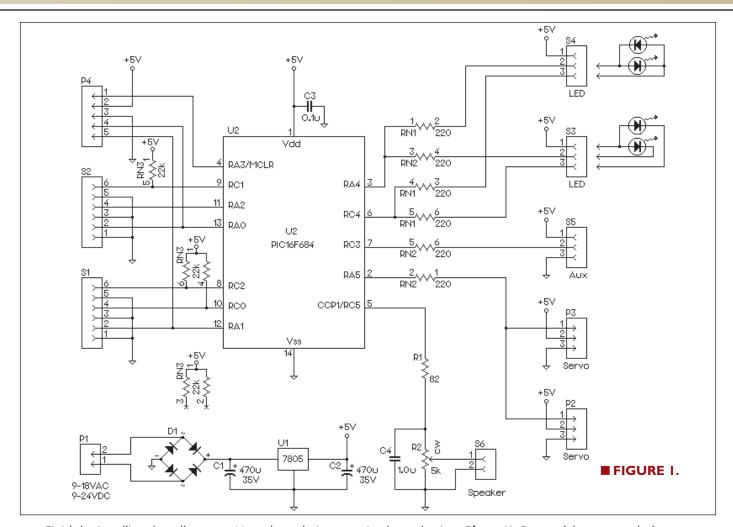


want to start with just one flashing signal and add the gates and speaker later. Please note that there is only enough power to operate LEDs and not incandescent bulbs.

A 7805 regulator provides the required five volts. It accepts a wide variety of input voltages – both AC and DC – but is typically driven from the 18 VAC accessory power provided by most train transformers. The relatively large 470 μ F capacitor on the output provides momentary surge power for the servo motors. The 220 ohm resistors on the PIC's output lines not only limit the current for the LEDs, but also protect the PIC from short circuits if connectors are plugged in wrong or wires on the layout are shorted. The 22K pull-up resistors are required for inputs on port *C*, but not port A since it has built-in pull-ups. The 82 ohm resistor limits the maximum current drawn by the 100 ohm speaker. The 5K pot provides a volume control. The 1.0 μ F cap filters out most of the aliasing noise in the bell waveform which is sampled at 8 kHz.

Assembly

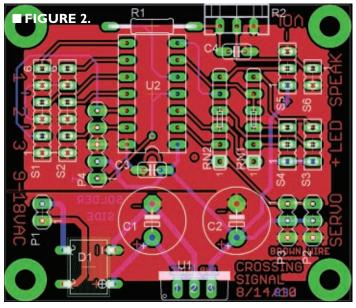
Begin by gathering all the parts and necessary tools. Solder in the shorter parts first. Install the 14-pin IC socket for U2, being sure to orient it correctly. Install the two 220 ohm resistor networks RN1 and RN2. Don't confuse their locations with the six-pin locations used for sockets S1 and S2. Tape is useful to hold the connectors in place while soldering. Mount the 1.0 μ F capacitor, C4, on-end.



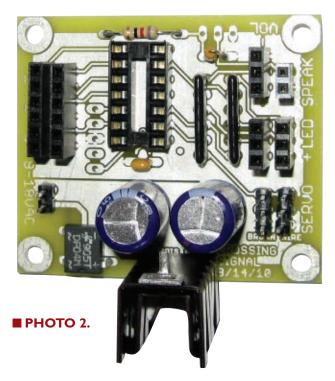
Finish by installing the taller parts. Note the polarity on the electrolytic capacitors C1 and C2. Solder in the 7805 regulator with its metal tab facing C1 and C2. Apply about a pinhead's worth of heatsink compound to the heatsink, then – with its silver tab up – slide it onto the 7805 until it's even with the bottom of the regulator's case. There should be a slight air gap between the heatsink and C1 and C2. Make a power cable with a length of two-conductor wire soldered to a two-hole connector. Plug it onto the two pins extending up from the circuit board, and connect the other end to a nine to 18 volt power source. If the five volt output of the 7805 checks out okay, remove power and plug in the PIC, being sure to orient it properly. Refer to **Photo 2**.

Reed Switches

At least two reed switches must be mounted on your layout: one to detect approaching engines and one to detect when the last car (caboose) has cleared the road. If trains run in both directions, then a third switch must be mounted. The circuit supports up to six reed switches which will detect trains on double tracks traveling in both directions. Cables for the reed switches are made as follows. Bend the leads on a reed switch using needle-nose pliers so that they both point in the same direction, parallel to the glass tube (see **Photo 3**). Be careful not to apply force against the tube because it can break easily (believe me, I know). Trim the lead with the 180 degree bend so that its solder joint will end up alongside the glass tube where it can't short against the other lead. Trim this other, straight



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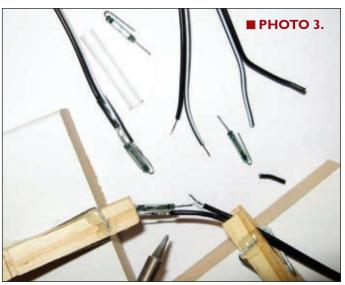
lead about 1/4" from the tube. Trim the leads on a one foot length of 24 gauge two-conductor wire to match the leads on the reed switch, and solder them onto the switch. This is easier if you tin all four leads first. Make cables for the other reed switches, but use lengths of wire needed for the approach distance (plus a little extra for "engineering margin"). Encapsulate each reed switch in a 1-1/2" length of 3/16" heat shrink tubing. The heat shrink should not extend beyond the end of the glass tube (again see Photo 3). Attach a two-pin connector to the leads at the opposite end of the reed switch cable. An easy way to do this is to strip 1/8" insulation off both leads (simultaneously), then align the stripped ends with a two-pin connector as shown in Photo 4. The short leads of the connector are the soldered end. Slightly spread the wire leads if necessary using a screwdriver blade to get a precise alignment with the connector. Solder and insulate the joint with a dab of hot melt glue.



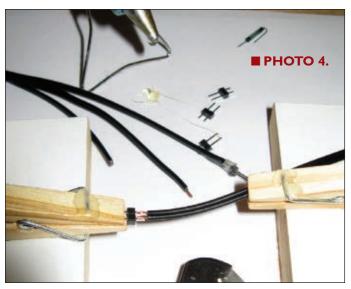
Several different brands of signals can be used. Photo 1 shows a pair of realistic ones (in both HO and N scales) made by N.J. International. These come with LEDs and a gate arm. Much less expensive signals are made by Bachmann (also available in HO and N scales), but you'll need to install the LEDs yourself and purchase the gates separately. **Photo 5** shows a pair of Life-Like signals which are more realistic than Bachmann's, but they also require the LEDs to be installed. Photos 6 and 7 show how to do this. The original lenses are removed by placing them over an 1/8'' hole (such as on the circuit board) and punching them out with a nail set. The alignment jig prevents melting the delicate plastic signal while soldering. Connecting the LEDs back to back (rather than common anode) requires only two leads, and fine magnet wire is easy to conceal. A small connector attached to the opposite end facilitates mounting (and especially un-mounting) on your layout.

Initial Checkout

Plug the reed switch cables into S1 - the six-hole socket at the edge of the board. The cable from the center of the road goes into the center two holes marked "2." The order of the other reed cables doesn't matter, but they must be on S1. Plug a signal into the three-hole LED socket. Note that two kinds of signals are supported: the Life-Like ones use a two-pin connector, while most commercial signals such as N.J. International's use a three-pin connector. The black wire (common anode) on the three-pin N.J. International connector goes in the hole marked "+." The two-pin Life-Like connector goes in the other two holes. If the signals are plugged in wrong, it won't damage anything. At this point, you might also want to connect the speaker and servos, but we're mainly interested in testing the signal LEDs. Power on the circuit board. Both LEDs on the signal should flash momentarily. Use a hand magnet to activate one of the reed switches. The signal should alternately flash right and left. It might stop after 30 seconds depending on which reed you activate, but this is normal and explained



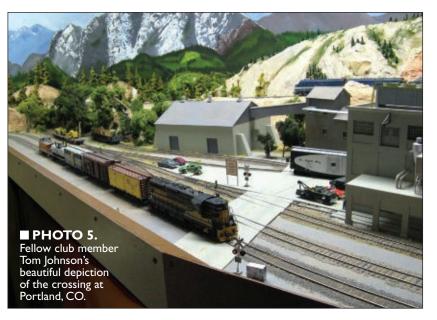
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below. As with any electronic device, it's a good idea to use a surge suppressor on the 120 VAC power line.

Mounting

Drill 3/16'' holes in the center of the track to mount the reed switches. Insert them from the underside of the layout and glue (Goop) them in place. The switches should protrude up about 1/32'' below the tops of the rails as shown in **Photo 8**. Attach a $1/8'' \times 1/8''$ super magnet with its poles aligned up and down, under an engine in its center near the front. It might not be necessary to glue the magnet if it simply attaches to some metal. Similarly, attach a magnet under a caboose (or other rear-end car) at its center near the rear. The magnets should clear the tops of the rails by about 1/32''. If it turns out the magnets aren't strong



enough to activate the reed switches, additional magnets can be stacked onto the originals. The magnetic fields will add. Mount your signals on the layout following the manufacturer's recommendations. The way a speaker is mounted can make a big difference in its volume and sound quality. Ideally, it should be mounted against a 1-1/2" hole in the surface of the layout. **Photo 8** shows the top of the hole covered with nylon screen which can be concealed by scenery materials. An array of 1/4" holes drilled in the layout works almost as well. Another possibility is to mount the speaker on a small board under the layout as shown in **Photo 9**.

Servo Mounting

Installing the servo motor that operates the crossing gate requires some skill and dexterity, especially for N scale and smaller. Begin by mounting your gate on the layout. If it's designed to be operated — like the N.J. International gate — it will have a linkage wire already attached. Otherwise, you'll have to devise this linkage yourself. (A paper clip works well for a Bachmann gate.) Drill an oversized hole (about 1/8") in the layout to allow for some lateral movement of the linkage.

Be aware that — especially for N scale the servo and reed switch at the road are mounted near each other and could interfere with each other. Also, the linkage on an N-scale N.J. International signal must be extended if the layout board is 3/4" thick or more.

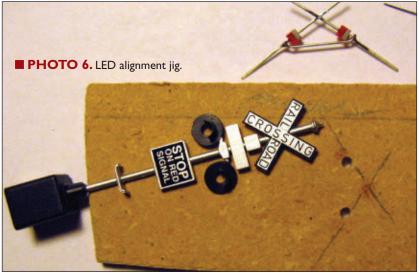
The servo must be rotated to the down position of the gate. Plug it into the circuit board (noting the orientation of the brown wire on its cable) and power up the board. The servo will immediately move to its up position. Move it to its down position by activating one of the reed switches. When the servo is in the down position, remove power. Attach the small nylon arm to the servo at the approximate angle shown in **Photo 10**. This will make its average position horizontal as it swings up and down. Determine how the servo will be oriented in its bracket and on the layout. The servo arm rotates clockwise to move the gate up.

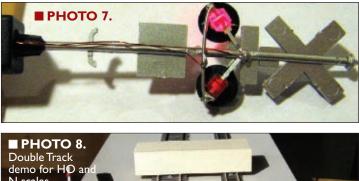
Loosely attach the servo to the mounting bracket with 2x12 mm machine screws. Mark the position where the bracket and servo will be mounted on the layout. For the HO N.J. International gate, the linkage moves about 1/8" to move the gate through its range. (The precise range is adjusted later.) This roughly corresponds to the movement of the innermost hole in the servo arm. Therefore, this hole should be centered under the linkage hole in the layout.

Be sure that any wires from the signal are routed such that they don't interfere with the motion of the servo. Make two holes in the layout to mount the bracket by its (adjustable) slots, then attach it with wood screws.

Bend about 1/4" of the linkage wire 90 degrees so it engages in the hole in the servo arm. Insert it in the hole, then bend the last 1/8" of the linkage another 90 degrees to keep it from slipping back out of the hole.

Slide the servo up or down in the bracket slots to move the gate to its down position, then tighten the screws that







hold the servo onto the bracket. You may want to put a dab of Loctite (nail polish or glue) on the nuts.

Cycle power on, then off. This will move the gate to its default up position. This position is adjusted as follows. With the power off, put magnets on reed switches 1 and 2 so they close. (These reeds are on positions 1 and 2 on connector S1.) Turn the power on. One of the signal LEDs will remain on indicating that the microcontroller is in its gate-adjustment mode.

Gently tap the magnet on reed switch 1 to move the



gate upward a step at a time. Tapping on reed 2 moves the gate downward. Using these controls, adjust the desired up position. When power is cycled, the gate will operate from this position, because it's recorded in the PIC's internal EEPROM. If the servo makes a buzzing noise, it's probably straining against a physical limit. It should be readjusted to prevent it from drawing excessive power.

Software

Since I'm a programmer by trade, I like to think that most of the magic of this project is in the software. It's what enables us to get by with such a simple and inexpensive circuit. Because the bell waveform is sampled by an interrupt routine 8,000 times per second, and because the

ſ			DESCRIPTION	SUPPLIER PART #
	Parts 1	for crossif	ng signal circuit board (in recommended order of a Printed circuit board	Advanced Circuits
	1	R1	Resistor, 82 ohm	Mouser 291-82-RC
	1	D1	Bridge rectifier, DF04M, 1A, DIP	Jameco 102971
	1	DI	Socket, IC, 14-pin DIP	Jameco 112214
		RN1,2	Resistor networks, 220 ohm, six-pin SIP	Mouser 267-220-RC
	2	S1,S2	Sockets, 0.100" six-hole	Digi-Key S7004-ND
	5	P2,P3	Headers, 0.100" three-pin	Digi-Key WM6003-ND
	3	S3-S5	Sockets, 0.100" three-hole	Digi-Key S7001-ND
	2 2 5 3 7 2	P1	Headers, 0.100" two-pin	Digi-Key WM6002-ND
	2	S6	Sockets, 0.100" two-hole	Digi-Key S7000-ND
	1	C4	Capacitor, 1.0 µF axial ceramic, "105"	Jameco 546345
	2	C1,C2	Capacitors, 470 µF 35V radial	Mouser 667-ECA-1VM471B
	1	U1	LM7805AC, 5V regulator, 1.5A	Jameco 924570
	1	01	Heatsink TO-220	Jameco 2077327
	1	C3	Capacitor, 0.1 µF radial ceramic, "104"	Jameco 544868
	1	RN3	Resistor network, 22K ohm, four-pin SIP	Mouser 264-22K-RC
	1	U2	PIC16F684, microcontroller, programmed, DIP-14	
	30		Feet wire, 24 gauge, two-conductor, stranded	Jameco 100299
	6		Magnetic reed switches	Digi-Key HE503-ND
	6		3/16" heat shrink tubes, clear	5 .,
	4		LEDs, 1.9 mm red diffused, Avago	Jameco 1555219
	4		Spacers, mounting, 5/32" #4	
	4		Wood screws, #4 1/2" Phillips	
	6		Rare earth magnets, rod 1/8" x 1/8"	Magcraft.com NSN0658
	Parts	for gate se		-
	1		Servo motor, Dynam 9g	HobbyPartz.com
	1		Mounting bracket	
	2		Machine screws & nuts, 2x12mm	
	2		Wood screws, #4 1/2" Phillips	PARTS LIST
	2		Linkage wires (paper cirps)	
		for bell so		
	1	De	Speaker, 1.75" square, 100 ohm, 0.2W	Mouser 25SP023
	1	R2	5K pot	Mouser 652-3352W-1-502LF
	14		Inches wire, 24 gauge, two-conductor, stranded	
	1		Header, 0.100" two-pin	Digi-Key WM6002-ND
	4		Wood screws, #4 1/2" Phillips	
	1		2.5" x 2.5" screen mesh, nylon	

servos are sensitive to even a few microseconds of variation in pulse widths, I chose assembly language rather than C code to maintain control over each CPU cycle. Basically, the code executes a big loop 20 times per second that samples the reed switch inputs and, if a train is detected, it alternately flashes the LEDs and sends gradually changing pulse widths to the servos. While this is going on, a timer triggers an interrupt handler that sends sound samples to the PIC's PWM register. The servos move to a position dependent on the width of a pulse that they receive. When the gate moves down, the pulse widths step down 20 times a second, from about 1,500 microseconds to 750 microseconds. The bell waveform data takes up almost all of the PIC's 2K words of program memory. Sampled at 8,000 times per second, this provides a

fleeting quarter second of sound, but it's sufficient for one strike of the bell. It's possible that the system can get out of sync, for instance, if a caboose is mistaken for an approaching engine. Thus, the software utilizes a 30 second time-out to resynchronize things. The exact way all of this works is explained in the meticulously commented



source code that can be downloaded from the article link at **www.nutsvolts.com**.

Conclusion

This crossing system has been successfully installed on both HO and N scale layouts by several members of my railroad club; so the bugs have been worked out. A couple lessons were learned the hard way. Don't connect the PIC's MCLR pin (which can be reconfigured as an input) with a long cable because it's sensitive to electrostatic discharges (ESD). Also, don't assume while laying out traces on the circuit board that you can arbitrarily reassign the PIC's I/O lines to untangle things. Unlike the PIC24, the PIC16 only provides internal pull-ups on port A, not on port C (refer to **Photo 11**).

The auxiliary (AUX) signal was added at the request of a club member who wanted to have a watchman raise a lantern. If you use an inductive load on this signal, there should be a protection diode (an ordinary diode connected backwards across the coil) to suppress any back EMF. The AUX connector also provides a regulated five volts. If you really care about the "little people" on your model railroad, you'll protect them with this road crossing signal!

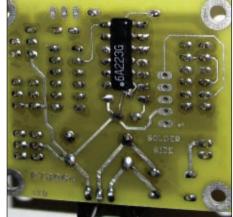
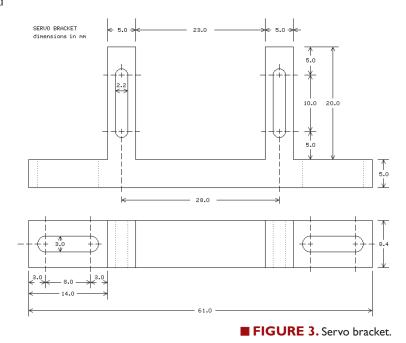


PHOTO II. A lesson learned the hard way was that pull-ups had to be added to the port C inputs. Two leads on RN3 are removed to prevent shorts, and the common lead is jumpered to +5V.











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Partner to the Fuel/Air Mixture Display Kit KC-5485 (see below) and intended to be used with a Bosch wideband LSU4.2 oxygen sensor to accurately measure air/fuel ratios over a wide range from rich to lean. It can be used for precise engine tuning and can be a permanent installation in the car or a temporary connection to the exhaust

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- 12VDC
- PCB and electronic components
- Programmed PIC
- · Case with screen printed lid

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Display your car's air-fuel ratio as you drive. Designed to monitor a wideband oxygen sensor and its associated wideband controller (see KC-5486 above). Alternatively, it can be used for monitoring other types of engine sensors.

- Double-sided plated through PCB
- Programmed PIC
- Electronic components
- · Case with machined and
- screen printed lid
- PCB: 80 x 50mm

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One small board and a handful of parts will allow you to create either a regulated ±15V rail or +15VDC single voltage from a single winding or centre tap transformer (not included).

- · Includes all PCB and components for board, transformer not included
- PCB: 72(L) x 30(W)mm

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Suits tanks up to 2.4m high

• PCB: 104 x 78.5mm

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When you modify your gearbox, diff ratio or change to a large circumference tyre, it may result in an inaccurate speedometer. This kit alters the speedometer signal up or down from 0% to 99% of the original signal. With this improved model, the input setup selection can be automatically selected and it also features an LED indicator to show when the input signal is being received. Kit supplied with PCB with overlay and all electronic components.

• PCB: 105 x 61mm

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This improved model features fast data transfer, capable of transmitting PayTV digital remote control signals using the Pace 400 series decoder. Kit supplied with case, screen printed front panel, PCB with overlay & all electronic components.

• PCB[·] 79 x 47mm

• Requires 9VDC wall adaptor

"Minivox" Voice Operated Relay

KC-5172 \$11.50 plus postage & packing

Voice operated relays are used for 'hands free' radio communications and some PA applications etc. Instead of pushing a button, this device is activated by the sound of a voice. This tiny kit fits in the tightest spaces and has almost no turn-on delay. 12VDC @ 35mA required. Kit is supplied with PCB electret mic, and all specified components.

PCB Dimensions: 47 x 44mm



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(available separately, use RP-3510 \$2.00).

- Kit supplied with PCB and all onboard electronic components
- Suitable enclosure UB3 case, HB-6013 \$2.50

Clifford The Cricket

KC-5178 \$12.50 plus postage & packing

Clifford hides in the dark and chirps annoyingly until a light is turned on - just like a real cricket. Clifford is created on a small PCB, measuring just 40 x 35mm and has cute little LED insect eyes that flash as it sings. Just like a real cricket, it waits a few seconds after darkness until it begins chirping, and stops instantly when a light comes back on.

• PCB, piezo buzzer, LDR plus all electronic components supplied

Hundreds Sold

Universal Voltage Switch

KC-5377 \$23.75 plus postage & packing

This is a universal module which can be adapted to suit a range of different applications. It will trip a relay when a preset voltage is reached. It can be configured to trip with a rising or falling voltage, so it is suitable for a wide variety of voltage outputting devices eg., throttle position sensor, air flow sensor, EGO sensor. You could even use it to trigger an extra fuel pump under high boost, anti-lag wastegate shutoff, and much more. Kit supplied with PCB, and all electronic components.

• PCB: 105 x 60mm

Ultrasonic Antifouling for Boats

KC-5498 \$179.50 plus postage & packing

Marine growth electronic antifouling systems can cost thousands. This project uses the same ultrasonic waveforms and virtually identical ultrasonic transducers mounted in sturdy polyurethane housings. By building yourself (which



includes some potting) you save a fortune! Standard unit consists of control electronic kit and case, ultrasonic transducer, potting and gluing components and housings. The single transducer design of this kit is suitable for boats up to 32ft boats longer than about 45ft will need two transducers and drivers. Basically all parts supplied in the project kit including wiring. ORE IN (Price includes epoxies).

- 12VDC
- Suitable for power or sail
- · Could be powered by a solar panel/wind generator







LEGACY COMMUNICATION WITH THE 32-BIT MICRO EXPERIMENTER

BYTHOMAS KIBALO

Legacy communication is still very prominent and in high use with microcontrollers. This article will explore the use of a legacy communication that exists between a 32-bit Experimenter (as a terminal) and a PC. Legacy communication uses the traditional RS-232 serial, and the communications payload is ASCII characters. Of interest is how the Experimenter PIC32 microcontroller achieves this, and what specific hardware and software is required.

block diagram of the experiment is shown in Figure 1. Figure 2 shows the prototype with an expanded view of the Experimenter display. One end of the RS-232 cable will connect to a PC (not shown). (A future article will discuss how to convert this RS-232 communication to the Experimenter on-board USB using a CDC class of device driver.)

To build a legacy Microcontroller Terminal, we'll use the 32-bit Experimenter (or Experimenter for short), the universal graphics board configured as a 12x8 character display, the Mini-kit (available from the *Nuts & Volts* Webstore) for keyboard and rotary inputs, and an RS-232 serial interface (consisting of both RS-232 cable and interface electronics). We will cover these components and associated software in more detail, in the discussions to follow. Think of all these components as building blocks that are used together for this experiment, but that can also be used individually or in different combinations for use in your own future Experimenter applications.

For example, an application can use the PS/2 keyboard for general operator entry, or use the rotary knob for a precision rotation and direction setting, or replicate the serial port across all five of the available Experimenter UARTs (Universal Asynchronous Receiver Transmitters) for enhanced legacy communications. In a sense, this experiment is an exercise in embedded system integration where diverse components are put together to create an assembled whole. Again, as in other Experimenter applications, some familiarity with C language

FIGURE I. Experimenter terminal setup.

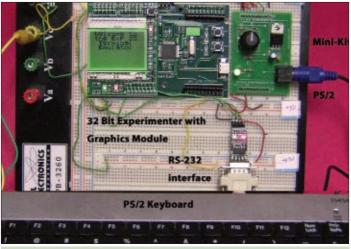


FIGURE 2. Prototype shown without PC.

is required. However, to make code reuse easier, the majority of demo software is written as library functions.

Working with the PC itself is fairly straightforward, as it already supports legacy serial communication with an application like HyperTerminal.

The 12x8 LCD Character Display Operation

The 12x8 character display uses the universal graphics display module that was introduced in the June '11 issue of *Nuts & Volts*. This graphics display has a pixel matrix of

www.nutsvolts.com/index.php?/magazine/article/october2011_Kibalo

3

102 horizontal by 64 vertical, for a total graphics complement of 6,528 pixels. By using a fixed font of 8x8 pixels per character, this display (using new software) emulates a 12x8 LCD character display (see Figure 3).

Here's the LCD display software library. The functions and their use are straightforward. This library is used to write strings onto the graphics display.

- void LCDInit(void) Called once to initialize the display and Experimenter I/O.
- void LCDUpdate(void) Updates the entire display with contents of LCDText[].
- void LCDErase(void) Clears the entire display and content of LCDText[].
- void LCDWriteLine(WORD number, char *line) Updates a designated row (indicated by number 1 to 8) on the display.

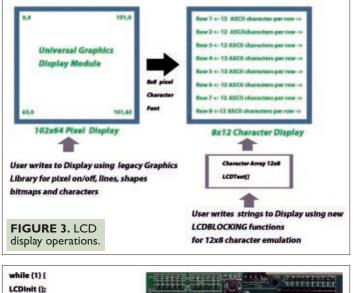
The good news is that the above four functions mimic Microchip's LCD interface for their family of evaluation boards using a character display, and are used extensively throughout their applications library. So, if you want to host their library applications on the Experimenter, just replace their library function LCDBlocking.H and LCDBlocking.C with our Graphics.h and Graphics.c. Everything should compile and work without issue. A demo program is provided that illustrates the use of this LCD library. Open LCDtest.MCP in the LCD folder, (included with the article downloads) build, and then program an Experimenter equipped with a universal graphics module. You should see what's in Figure 4. There are also additional library functions used to write characters onto the display:

- AT (X, Y) Macro to position the cursor on the display before writing a character. X is 0 to 11; Y is 0 to 7.
- **putcV** (char) Writes a character to the display memory at the current cursor position.
- **dumpVmap** () Updates the display with new characters.

Finally, there are functions in the display library that provide independent control of the two LEDS (green/red) on the display module, as well as the LCD display backlights (which are also green and red). The green backlight has a controlled brightness feature. The red backlight is simply on or off. Feel free to activate both and experiment with the color mix to "jazz up" your display operations.

- void redBLON(void) Turn red LCD backlight on.
- void redBLOFF(void) Turn red LCD backlight off.
- void greenBLOFF(void) Turn green LCD backlight off.
- void greenBLON(int setting) Turn green LCD backlight on with dim setting (0 to 64K), where 0 is dimmest and 64K is brightest.
- void greenLEDON(void) Turn green LED on.
- void greenLEDOFF(void) Turn green LED off.
- void redLEDON(void) Turn red LED on.
- void redLEDOFF(void) Turn red LED off.

The graphic module board itself contains SMT



LCDWriteLine (1, "Hello Dwarfs"); LCDWriteLine (2, "Hi ya Dopey"); LCDWriteLine (3, "Hi ya Sneezy"); LCDWriteLine (4, "Hi ya Sleepy"); LCDWriteLine (5, "Hi va Doc"): LCDWriteLine (6, "Hi ya Greedy"); LCDWriteLine (7, " Hi ya Happy"); LCDWriteLine (8, "Hi ya Grump"); while(1); FIGURE 4. LCD string display test demo. components that are pre-assembled. The module does

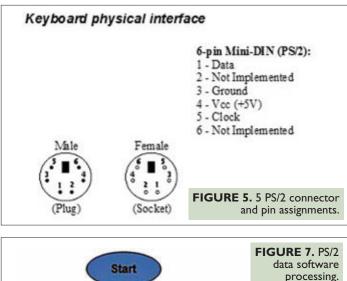
come as a kit (available at **www.nutsvolts.com**). The only assembly requirement is to solder the display and backlight to the board, as well as the headers to the board.

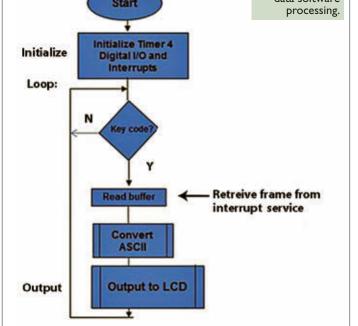
The PS/2 Keyboard

A keyboard provides a powerful input capability for Experimenter applications. With the advent of USB keyboards, the PS/2 versions have pretty much been replaced, but still exist in large guantities and are available at low cost. First, we will discuss PS/2 operation and then how the Experimenter integrates to a PS/2 keyboard.

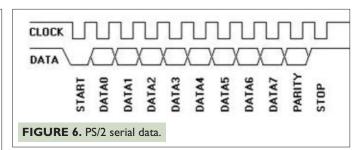
Historically between the PS/2 and the computer, there can be an extensive host-device protocol involved. For our purposes, we will simply stick to receiving key scan data from the keyboard, deciphering these key scans into ASCII, and then displaying this data on the Experimenter LCD display. We'll avoid any of the more difficult command features as they are not necessary for this terminal experiment.

Let's first examine the PS/2 physical interface. The PS/2 keyboard requires +5V to operate and generates its own data clock. PS/2 communications use a framed synchronous serial protocol to transfer data. The communication is "idle" when both lines (data and clock) are high (open-collector). This is the only state where the keyboard is allowed to begin transmitting data. The microcontroller has ultimate control over communications, and may inhibit communication at any time by pulling the clock line low. **Figure 5** shows the physical connector with









pin assignments used by the PS/2.

All data is transmitted "framed" one data byte at a time, and each data byte is sent in a frame consisting of 11-12 bits (see **Figure 6**). These bits are:

- One start bit. This is always 0.
- Eight data bits, least significant bit first.
- One parity bit (odd parity).
- One stop bit. This is always 1.
- One acknowledge bit (microcontroller to device communication only).

The parity bit is set (1) if there is an even number of 1s in the data bits, and reset (0) if there is an odd number of 1s in the data bits. This is used for error detection. Data changes during the clock high state, and is valid when the clock line is low. The baud (or bit) rate is not standard and can change from keyboard to keyboard. Typical clock frequencies can range from 10-16.7 kHz.

PS/2 Operation

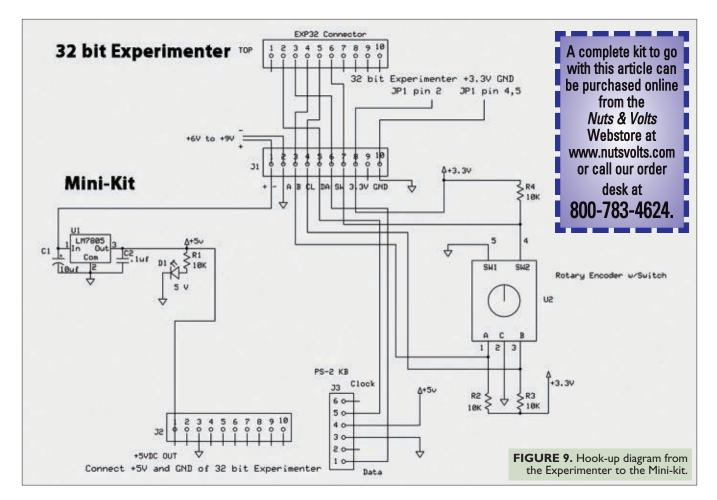
There are a few things we need to do here to retrieve a key that is pressed on the PS/2 and then to display this key as a ASCII character on the Experimenter LCD.

First, the Experimenter needs to sample the PS/2 data line and clock line simultaneously and then determine when keyboard incoming data bits are valid. For the PS/2,

the data valid condition occurs at the falling edge of the clock. We use the Experimenter PIC32 timer 4, as well as two of the PIC32 +5V tolerant digital inputs, to make this happen. The timer is run in an interrupt mode and configured fast enough to ensure that all clock edges are captured.

Here are some important specifics:

- 1. The PS/2 clock can be approximated as a square wave with a 50% duty cycle and a period of about 62.5 to 50 microseconds. If the timer 4 interrupt is set to sample every 25 microseconds, we are guaranteed during a timer 4 interrupt to get at least one sample of the PS/2 clock between falling and rising clock edges (good enough to capture an output PS/2 data sample).
- 2. Once a valid data sample is taken, it needs to be processed as a bit in context of the larger data frame. Once a total of 11 bits is captured, the frame needs to be validated for correct start, stop, and parity. If it passes this test, then it is a legitimate key scan code.



3. A key scan code is the raw scan data that represents the row and column position of the depressed or released key. It is generated by the PS/2 keyboard electronics as a result of scanning its matrix of 80-101 keys.

The final step in this process is to look up the corresponding ASCII representation associated with the key code. Once this is done, the key can then be displayed as ASCII on the LCD display. A flowchart of this operation is shown in **Figure 7**. The functional library elements for this component are:

- **void initKBD(void)** Initialize timer interrupts, digital I/O, and global variables for PS/2 operation.
- Char getcKBD(void) Retrieve and return valid ASCII codes as they have been received from the PS/2 keyboard. The KBDReady flag must be set to indicate that a valid key code has been received from the PS/2 keyboard. The current key code is processed into ASCII. KBDReady must be reset once the key has been processed.

The Mini-Kit

The Mini-kit was introduced in the Dec '10 Nuts & Volts article "Enhanced User Interface for the 16-bit Experimenter." It is recommended for use in this experiment because it provides a convenient small board format with all the necessary electronics and power supply (PS/2 keyboard interface and a +5V DC regulator) to integrate the PS/2

with the Experimenter. The kit also has an additional bonus in that it contains a rotary encoder knob that we can use for an additional data source (see **Figure 8**). We will experiment with both inputs. The +5V regulator is necessary to provide the required voltage levels for the PS/2 interface (to power up the keyboard and provide proper voltage levels for the PS/2 interface). Its 5V output is also available to the user to power up any external +5V hardware. As a side note, if you intend to use the Experimenter from an unregulated supply, consider using the Mini-kit +5V regulator output for the Experimenter. The hook-up diagram shows this.

The Experimenter itself is a +3.3V device but has +5V tolerant lines that are available for connecting and sampling the PS/2 data and the clock. Recommended hook-up lines are shown in **Figure 9**.

Keyboard Demo

Once you've completed the integration of the Mini-kit to the Experimenter hookups (as per **Figure 9**), proceed to the keyboard demo. Open the keyboard folder and then invoke workspace Keyboard.mcp. Compile and program/ download this workspace into the Experimenter. You should see the start up screen as shown in **Figure 10**. This screen times out to a blank screen that accepts any characters typed in from the PS/2 display. Typing beyond the limits of the LCD display will cause the characters to wrap around, starting at the initial display position of 0,0. The F10 key will clear the display and be in position for a new character

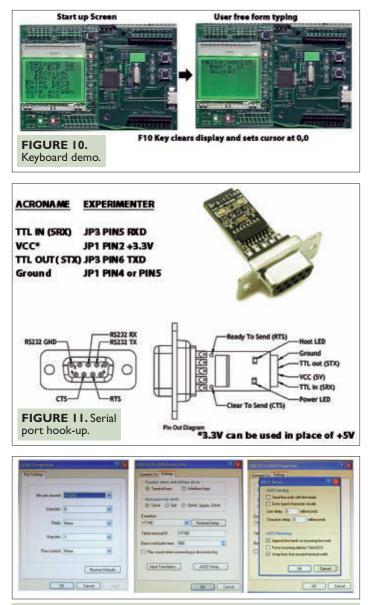


FIGURE 12. HyperTerminal setup.

at 0,0. Upon completion, you should be able to type on the keyboard and have the keyboard characters appear on the Experimenter LCD module (if one is present). Congratulations! You have a working PS/2 interface!

Serial RS-232 Port

Let's continue with our integration for the terminal application. To connect the Experimenter to a PC, a serial port is required. The port uses one of the five possible internal UARTs that are available with the PIC32.

The serial port can drive a PC serial RS-232 port only when the proper physical interface is made available. As a reminder, the PIC32 serial port is +3.3V to 0V out; RS-232 requires a +12 to -12 volt communication. The ACRONAME serial interface connector (**www.acroname.com**) provides an ideal physical interface and is pretty inexpensive. The ACRONAME interface is shown in **Figure 11**.

There are only four connections needed to work with the Experimenter. They are pins 1 and 2 of Experimenter JP2, and +3.3V and GND pins 2 and 4 (or 5) of Experimenter JP1. (Note the hook-up diagram in **Figure 11**.) To facilitate solderless board connection, use two .1" four-pin headers soldered together at a right angle to connect the interface to the board. The serial port library functions used in the data logger are:

- initU3() Initializes the serial port using UART3 of the PIC32 for 19200 baud, eight-bit data, no parity, and one stop bit (or shorthand 192008N1).
- **putU3(character)** Writes ASCII characters out to the serial port.
- **putsU3(string)** Writes an entire ASCII string out to the serial port.
- Character getU3() Waits for a new ASCII character to arrive at the serial port and returns it.

A demo test program is provided in the downloads. First, you must configure your PC with HyperTerminal for communication at 115200 baud, eight bits, no parity, one stop (115200 8N1). Follow the steps shown in **Figure 12**. The other setting is to configure the communications for VT100. With the VT100 configuration, we can issue escape codes to clear the HyperTerminal screen and return the cursor to the home position. In this way, we can mimic the LCD display operations with the F10 key.

Open the serial folder and then invoke workspace Serial.mcp. Compile and program/download this workspace into the Experimenter. You should now be able to type on the PC keyboard and see the results displayed on the Experimenter LCD.

Final Terminal Emulation Step

Let's complete the terminal application by integrating all the diverse components into a cohesive whole. At this point, all the individual pieces should be integrated and tested by the following checklist:

- 1. LCD 12x8 Hook up the universal module and run LCDtest.MCP.
- 2. Mini-kit Assemble and hook up.
- **3.** PS/2 Use assembled/connected Mini-kit; connect the PS/2 keyboard and run Keyboard.mcp.
- **4**. Serial Port Hook up the ACRONAME interface and run Serial.mcp.

Once you have successfully integrated and tested all the components, is it now time to run the final terminal application. Open the folder terminal.mcp, and build and download code into the Experimenter.

For the final test, we should be able to type on the PS/2 interface and see updates to both the LCD interface and the PC HyperTerminal window simultaneously. Examples are shown in **Figure 13**.

What About Rotary Inputs?

The other Experimenter input associated with the

Mini-kit is the rotary encoder. Let's incorporate that into our terminal emulation. The rotary encoder theory and operation were covered in detail in the Dec '10 issue article "Enhancing the User Interface With the 16-Bit Experimenter." The lessons learned there still apply here.

With a rotary encoder, the Experimenter can determine rotation direction and amount of rotation. These two physical phenomena can be used as an operator input. In this demo, we will use the rotary encoder rotations and direction to display a string on both the LCD display and HyperTerminal screen simultaneously. The PS/2 keyboard inputs will not be used. No change in hardware is required if the earlier hook-up diagram provided is followed.

Open the terminal folder and then invoke workspace rotary.mcp. Compile and program/download this workspace into the Experimenter. Results should appear as shown in Figure 14.

What's Next

We have now successfully integrated all the required functional components to realize a legacy communication operation with the 32-bit Experimenter. There are a number of components that can be used for other applications. At a later time, we'll show you a USB application that essentially does the same function but without the legacy RS-232. Until next time, happy 32-bit computing! **NV**

Experimenter
CAR HILLING CONTINUE
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FIGURE 13. Terminal emulation demo.

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FIGURE 14. Rotary operations.



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BY FRED EADY INVASION OF THE CHIPKIT MAX32

With the introduction of the Arduino-speaking chipKIT Max32, about the only world-changing thing left for me to see in my lifetime is the arrival of aliens. There have been lots of intrusions into AVR-controlled Arduino airspace. However, the arrival of the 32-bit Microchip flagship microcontroller in the Arduino sector is akin to the Vulcan scout ship T'Plana Hath's first contact on Earth.

FIRST CONTACT

The chipKIT Max32 parked in **Photo 1** is capable of quickly ferrying Arduino shield payloads with minimal or zero modifications. The chipKIT Max32's warp drive is based on a Microchip PIC32MX795F512L 32-bit microcontroller clocked at 80 MHz. The PIC32MX795F512L is rich in on-chip peripherals. However, external components to support its on-chip peripherals are only required when a particular peripheral is put into use. For instance, the PIC32MX795F512L supports Ethernet and CAN. If you do not need Ethernet capability, the Ethernet PHY IC need not be mounted. The same goes for CAN. There is no need to include the MCP2551 CAN transceiver if the PIC's internal CAN hardware is to remain idle in your application.

Arduino is a programmer's paradise that shields the complexities of the hardware from the programmer. To that end, a specially modified Multi-Platform IDE (MPIDE) has been developed to assist in chipKIT Max32 sketch development. However, most "programmers" these days are also interested in what they are loading their precious code into. With that, we're going to board that chipKIT Max32 you see in **Photo 1** and check out the instrumentation.

The chipKIT Max32's controlling microcontroller needs no introduction. The PIC32MX795F512L is the biggest, baddest, fastest microcontroller in the PIC family of devices. Every logical embedded peripheral a designer would need



for an embedded application is integrated into the PIC32MX795F512L's silicon fabric.

This silicon embeds an intense USB subsystem. Thus, it would be logical to ascertain that the PIC32MX795F512L has no need for any help from an external USB-enabling peripheral IC. So, why is there an FTDI FT232RQ standing behind the chipKIT Max32's Mini-B USB connector? Answer me this. Does your brand new laptop have a nine-pin male RS-232 connector? The correct answer is "nope." Okay. Does your brand new laptop have a USB connector? The answer is "Yep. More than one." An RS-232 portal may not be physically mounted on your PC frame, however, RS-232 is not totally dead yet. So, if you don't have a regulation RS-232 interface on your laptop, how do you establish RS-232 type connectivity? Easy. You use one of the USB portals and a COM port driver. Okay. Then, why is that FTDI FT232RQ mounted on the chipKIT Max32 printed circuit board (PCB)? The answer lies in Schematic 1. Note that the FT232RQ's TXD and RXD signal lines are physically tied to the PIC32MX795F512L's UART1 pins.

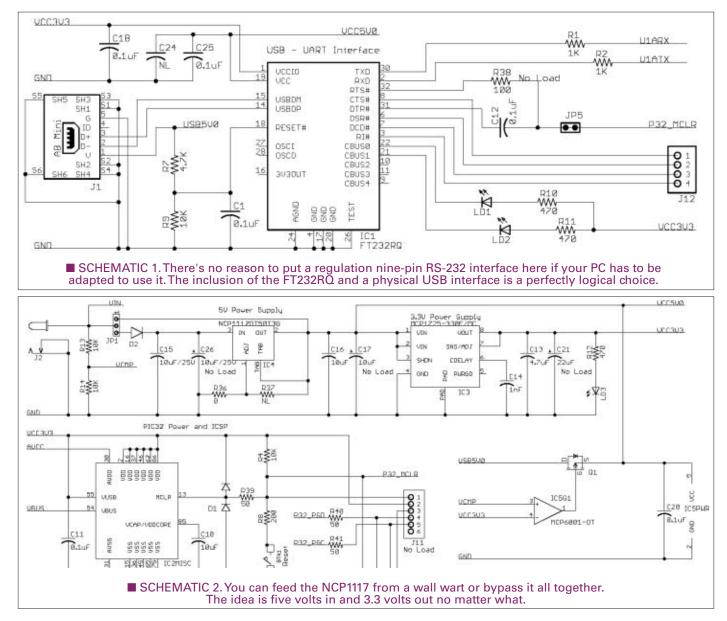
The chipKIT Max32's FT232RQ-based USB portal serves a dual purpose. From a communications standpoint, the Max32's USB portal provides a data path between the PIC and the MPIDE. Power for the Max32 can also be derived from the USB connection. Data activity is visually indicated by a pair of LEDs that sit directly behind the RESET button. The only other remarkable support components onboard

the Max32 are the power supply electronics. The chipKIT designers took a page from the Vulcan play book. If it works and it has worked for a very long time, continue to use it. The 5.0 volt regulator in **Schematic 2** is a descendant of the caveman LM317 adjustable voltage regulator. In the case of the Max32, the Jetsons NCP1117 variant is a fixed voltage adaptation of its LM317 predecessor. The NCP1117's

■ PHOTO 1. It really doesn't matter if you lean towards Arduino or not. The chipKIT Max32 can work in the Arduino environment or fall back on its Microchip MPLAB/C32 C compiler roots. Either way, you have direct access to the resources of Microchip's most powerful microcontroller.

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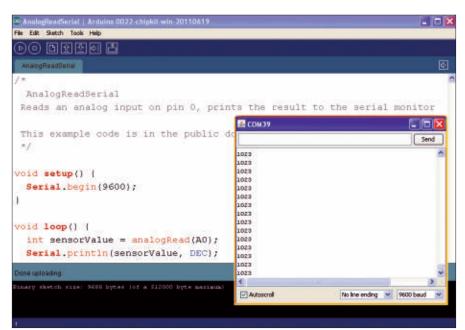
DESIGN CYCLE



job is to bring the raw chipKIT Max32 input voltage (7 to 15 VDC) down to an acceptable level (under six volts) for the MCP1725. The NCP1117 is a low dropout version of the standard LM317 and has a lower dropout voltage than its contemporary, the LM1117. In addition, the NCP1117 can supply 1.0 amperes of current versus the 800 mA offered by the LM1117. The "MCP" tells us that the 3.3 volt MCP1725 low dropout voltage regulator is a Microchip product. The MCP1725's entourage includes a minimum of relatively small and inexpensive stabilizing components. In fact, capacitors C17, C21, and C26 in Schematic 2 are not mounted on the chipKIT Max32 PCB. The big picture is to use the NCP1117 to drop the input voltage to an acceptable level for presentation to the MCP1725 which will provide a very stable 3.3 volts at up to 500 mA for the Max32's PIC32MX795F512L and 3.3 volt peripherals.

The NCP1117's role depends on you. You can feed the NCP1117 from a wall wart, an external power supply, or totally ignore it by way of the POWER SELECT jumper. The

5.0 volts supplied by the NCP1117 to the MCP1725 can also be usurped by a host PC's USB portal. As you can see in Schematic 2, a Microchip MCP6001 comparator samples the voltages presented at the VCMP voltage divider junction and VCC3V3 output of the MCP1725. The output of the MCP6001 comparator drives the gate of a P-channel MOSFET which blocks or passes the USB portal's 5.0 volts to the VCC5V0 output. The comparator hookup makes sure that the external input voltage always wins the power sourcing battle when a USB power source is attached. If vou're wondering why we need +5.0 volts other than to feed the MCP1725, it's the shields, Scotty, the shields. Many of the native AVR-based shields want to see a 5.0 volt power source. The chipKIT Max32 is designed to be compatible with Arduino shields. The Max32 digital I/O pins are five volt tolerant. However, the PIC32MX795F512L's analog-capable pins are not five volt tolerant. To keep the magic smoke contained within the PIC silicon, every PIC analog-capable pin is protected by a clamp diode and



■ SCREENSHOT 1. A jumper between the chipKIT Max32's A0 analog-to-digital input and +3.3 volts maxes out the PIC32MX795F512L's 10-bit analog-to-digital converter.

current limiting resistor. Regardless of the protection, the maximum analog input voltage remains at +3.3 volts. Scanning the Max32 landscape does not reveal any *Area 51* PIC32MX795F512L programming hardware. Instead, the standard set of six ICSP pads lie between the power connector and USB Mini-B interface. The ICSP pads are configured to interface directly to a PICkit3 programmer/debugger. If you accidentally find a way to eliminate the factory-loaded bootloader, the ICSP pads and an external PIC32MX795F512L programmer are the only way to reload the bootloader code. The ICSP portal is also your door to loading and debugging your home-grown PIC code.

The Max32 physically presents us with 83 I/O pins, 16 analog inputs, a 10 Mbps/100 Mbps Ethernet MAC, a USB 2.0 Full Speed OTG (On The Go) controller, a pair of CAN controllers, RS-232 modem control handshaking signals (DCD, CTS, etc.), an SPI portal interface, and a power portal. The length of the Max32 I/O access list translates into a gaggle of associated I/O pins and female headers. Fortunately for us, the Arduino boilerplate logically sorts them all out for us. The Max32 provides female headers that supply power to the shields and establish data paths between the shield electronics and the PIC32MX795F512L. The Arduino system dictates that the Max32 I/O pins range from zero to 85. Each I/O and analog input is clearly identified by silkscreen legend. We've walked around, on, and over most every major component that comprises the base chipKIT Max32. It's time to take our Max32 out of space dock and do some maneuvering.

OUT-OF-THE-BOX ARDUINO

Your Max32 will come preloaded with a bootloader **58 NUTS** VOLTS October 2011

and a simple LED blinker application. I grew out of the LED blinker program rather quickly and decided to try out the MPIDE and Max32 with a simple analogto-digital (A-to-D) application. So, I started the MPIDE, selected the Max32 as my target platform, and loaded the AnalogReadSerial example sketch.

Arduino is an intuitively obvious system. So, I won't insult your intelligence with a blow-by-blow description of how I loaded and ran the example program. Everything you need to see concerning my test run is displayed in **Screenshot 1**. I placed a jumper between the Max32's A0 A-to-D input and the 3.3 volt power pin on J10. The PIC32MX795F512L's Ato-D subsystem has 10 bits of resolution, and the 1023 decimal (0x3FF) result is what one would expect to see. I used the MPIDE's serial monitor tool to display the

A-to-D readings in Screenshot 1.

Running those simple Arduino demos was enough to convince me that the PIC32MX795F512L could truly speak Arduino. I'll leave it to you to experiment with the Max32 at this level. If you're new to Arduino or the PIC32MX795F512L, I suggest stopping here, whipping out your Max32 hardware, firing up your copy of MPIDE, and doing some getting acquainted. You'll find lots of Max32-adapted Arduino example code within the MPIDE framework. Once you get comfortable with the Max32 and Arduino, you can pick up and continue from here.

THE ULTIMATE chipKIT Max32 SHIELD

Back in my corporate days, I was the communications specialist for our sales team. My sales pitches were purely technical and focused on the importance of setting up and maintaining a robust data network. I would always emphasize the point that no matter how big, how fast, or how expensive the CPU was, it was a useless piece of junk if it could not communicate with its peers and slaves via some sort of network. The same axiom holds true for small embedded devices like the Max32. With that networking thought in mind, get an eye full of **Photo 2**. Recall that Ethernet PHY IC we mentioned earlier? It's here. Remember that pair of MCP2551 CAN transceivers that were needed for the Max32 to participate in a CAN network mount? They're here too. The chipKIT Network Shield is an extension of the PIC32MX795F512L's internal communications subsystems. By fitting the Network Shield to the Max32's female headers, we gain total access to the PIC32MX795F512L's 10 Mbps/100 Mbps Ethernet engine and its CAN networking subsystem. The Ethernet PHY is made up of an SMSC LAN 8720 PHY IC and a Bel Stewart Maglack. The chipKIT Network Shield sources everything the PIC32MX795F512L's Ethernet MAC requires, right down to the 25 MHz crystal. The Ethernet

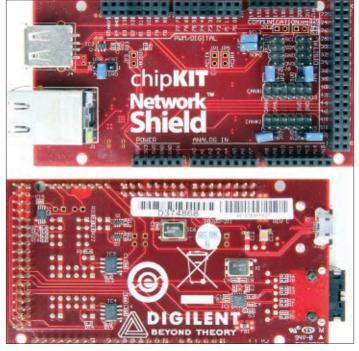
DESIGN CYCLE

PHOTO 2. The chipKIT Network Shield complements the PIC32MX795F512L's internal communications subsystems by providing the physical means to access them.

PHY portion of the Network Shield is graphically depicted in **Schematic 3**.

The MCP2551 CAN transceiver circuits are inked in **Schematic 4**. Note that the transceivers require a +5.0 volt supply. If you're CAN challenged, those 120Ω resistors are CAN termination resistors. The termination resistors reside at each end of the CAN network. Portals for accessing the I²C bus and an onboard 24LC256 I²C EEPROM are also revealed by the installation of the Network Shield. If you plan to use your Shield in an I²C network, you must use bus pull-up resistors in accordance with the I²C specification. Note the 2.2K Ω pull-up resistors on the 24LC256 SCL and SDA lines in **Schematic 5**.

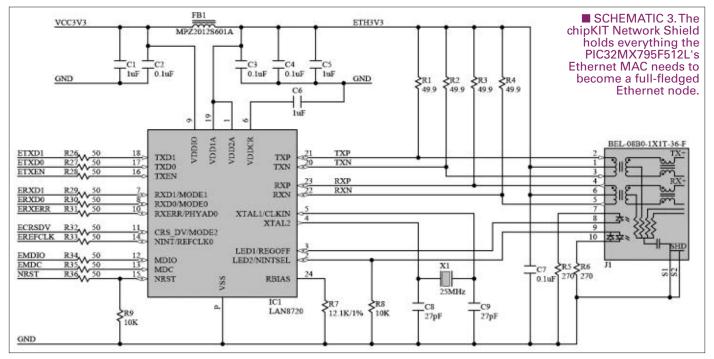
To utilize the PIC32MX795E512L's internal RTCC (Real Time Clock Calendar), all we really need from a hardware point of view is a 32.768 kHz clock source. The Network Shield is loaded with a 32.768 kHz crystal oscillator which we can use to keep up with the time and date. The Network Shield is equipped with a pair of USB interfaces. The Max32 can be configured as a USB host, a USB device, or a USB OTG device. When configured to act as a USB host, the Network Shield must have a means of supplying +5.0 volts at the USB host interface. Take a look at Schematic 6. The TPS2051B is a solid-state power switch that is under the control of the Max32's VBUSON signal. When the TP2051B's EN input is logically high, VCC5V0 flows to the TP2051B's OUT pin. The TP2051B's OUT pin voltage is manually directed to one of the USB interfaces via a jumper. An overcurrent condition is reported to the PIC32MX795F512L via the TP2051B's active-low OC pin. We've established one thing for certain: the Max32 loaded with a Network Shield is

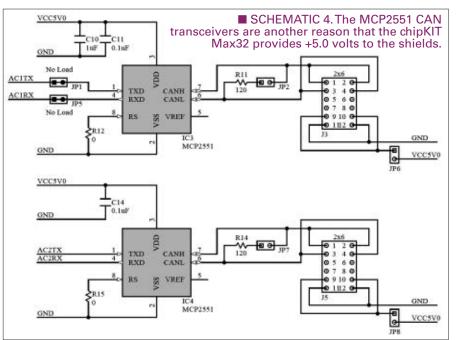


definitely not a piece of junk. The Max32/Network Shield combination has the ability to communicate with most any embedded or cloud-based computing platform. The Max32 can even talk to your car.

BILINGUAL SILICON

The Max32 was designed to support Arduino with a Microchip 32-bit microcontroller. However, the Max32 and the Network Shield combine to form a perfect PIC development platform. If you want to use the Max32 as a development tool, you'll need to take control of the





Max32's FTDI-based USB portal. We'll do this using the PIC32MX UART peripheral library. The Arduino *Serial.begin(9600);* translates to this PIC32MX UART Peripheral Library snippet:

#define GetSystemClock()
#define GetPeripheralClock()
#define GetInstructionClock()
(GetSystemClock() / 2)

80000000UL 40000000UL

UARTConfigure(UART1A, UART_ENABLE_PINS_TX_RX_ONLY); UARTSetLineControl(UART1A, UART_DATA_SIZE_8_BITS | UART_PARITY_NONE | UART_STOP_BITS_1); UARTSetDataRate(UART1A, GetPeripheralClock(), 9600); UARTEnable(UART1A, UART_ENABLE_RX_TX);

The PIC32MX UART Peripheral Library mnemonics are self-commenting and are as easy to read and follow as an Arduino sketch. Now that we have the Max32 UART's attention, sending a byte (0x55) of data is as easy as:



Loosely translated, the PIC32MX UART Peripheral Library serial transmit looks like this Arduino statement:

Serial.write(0x55);

Odds are if you need to transmit, you may also need to receive. In Arduino speak, receiving data is as easy as:

```
int character;
if (Serial.available() > 0)
{
character = Serial.read()
```

Like the Arduino receive code, the Peripheral Library mnemonics are an easy read. There isn't much difference between the Arduino and the Peripheral Library code. As far as the PIC32MX795F512L is concerned, if you can code in C, you can code in Arduino, and vice versa:

UINT8 character;

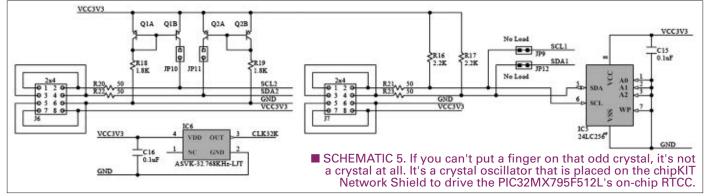
```
if(UARTReceivedDataIsAvailable(UART1A))
{
    character = UARTGetDataByte(UART1A);
}
```

Whether they are legacy RS-232 or USB-based, serial portals are staples in embedded computing. So, it's no surprise that Arduino has a strong set of serial routines in its bag.

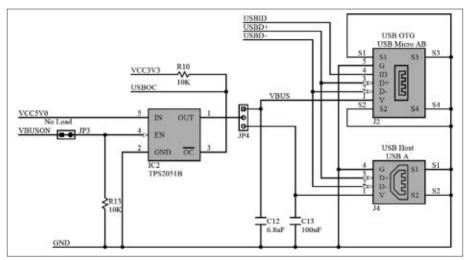
Some of the Max32 peripherals cannot be accessed bilingually using C and Arduino. However, at this very moment, there are guys and gals out there porting more and more PIC32MX code to the Arduino platform.

THE chipKIT Max32 AND THE MICROCHIP APPLICATION LIBRARIES

I decided to throw caution to the wind and load up my Max32 with a Microchip Application Libraries thumb drive application. Using my PICkit3, I blew up the Max32 factory bootloader and replaced it with the Mass Storage simple demo which is part of the Application Libraries.



DESIGN CYCLE



■ SCHEMATIC 6. The PIC32MX795F512L contains all of the USB interface circuitry. All that we need to supply are the physical connectors. If our chipKIT Max32 is to act in the USB host capacity, throw in a switchable +5.0 volt power source with those connectors.

Here's the heart of the application:

//Opening a file in mode "w" will create the file if it // doesn't exist. If the file does exist it will delete // the old file and create a new one that is blank. myFile = FSfopen("test.txt","w");

//Write some data to the new file.
FSfwrite("Bringing Arduino to the PIC32MX with chipKIT
Max32",1,50,myFile);

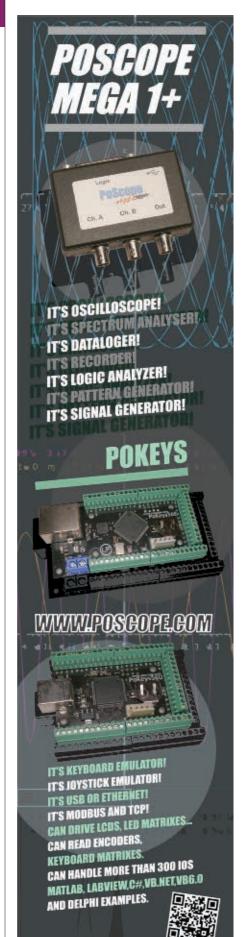
//Always make sure to close the file so that the data gets
// written to the drive.
FSfclose(myFile);

My contribution to the Mass Storage application is rather obvious. Before compiling the Application Libraries thumb drive code, I set the USB host power jumper to feed the larger Type A USB connector on the Network Shield and plugged the Shield into the Max32. I used the Max32's Mini-B USB portal to power the Max32/Network Shield combination from my laptop's USB portal. Next, I opened the Mass Storage application with MPLAB, attached a PICkit3 to the Max32, and compiled the thumb drive application using Microchip's C32 C compiler. The Application Libraries Mass Storage application is bare bones and doesn't give any indication that it is working or not. However, when I attached the thumb drive to the Network Shield's Type A USB connector, the thumb drive's activity LED began to blink. After a few moments, the thumb drive's activity LED ceased to blink and remained illuminated. So, I detached the thumb drive and inserted it into my laptop's USB portal. Sure enough, there was a file created on the thumb drive called *test.txt*. I opened *test.txt* with a text editor application to find that the file contained the following: Bringing Arduino to the PIC32MX with chipKIT Max32. Enough said.

MORE chipKIT Max32 ARDUINO STUFF

We've shown the Max32 and Network Shield for what they are: excellent development tools for both Arduino and native 32-bit PIC32MX applications. If you've been on the fence about a Microchip presence in AVR land, check out the chipKIT Max32 forums. I think you'll be pleasantly surprised. You can access the chipKIT forums and purchase your Max32 equipment from the Digilent website (**www. digilent.com**).

Meanwhile, I've got some Arduino-to-chipKIT Max32 porting to do! NV





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0100

DELAY_MS(100) WRITE_COMMAND(0x01): //SOFTWARE RESET WRITE_COMMAND(0x01) COMMAND_WRITE(OXEO,OXO1); //START PLL COMMAND_WRITE(OXEO,OXO3); //LOCK PLL WRITE_COMMAND(0X80): //SET LCD MODE SET TFT 188ITS MODE WRITE_DATA(OXOC): //SET TFT MODE & HSYNC+VSYNC+DEA WRITE_DATA(0x01): //SET HORIZONTAL SIZE=320-1 HIGH WRITE DATA(0x3F): //SET HORIZONTAL SIZE=320-1 LOWBYT WRITE_DATA(0x00); //SET VERTICAL SIZE=240-1 HIGHTBYTE WRITE_DATA(OXEF); //SET VERTICAL SIZE=240-1 LOWBYTE WRITE DATA(0x00); //SET EVEN/ODD LINE RGB SEQ.=RGB COMMAND_WRITE(0x3A,0x60): // SET R G B FORMAT = 6 6 6 WRITE_COMMAND(0xE6); //SET PCLK FREQ=4.94MHZ ; PIXEL WRITE_DATA(0x02) WRITE_DATA(OXFF); WRITE_DATA(OXFF); WRITE_COMMAND(OX84): //SET HBP, WRITE_DATA(0x01); //SET HSYNC TOTAL 440 WRITE_DATA(OX88); WRITE_DATA(0x00); //SET HBP 68 WRITE_DATA(0x44); WRITE_DATA(OXOF): //SET VBP 16=15+1 WRITE DATA(0x00) WRITE DATA(0x00): //SET HSYNC PULSE SUBPIXEL START PO TION WRITE_COMMAND(OX86): //SET VBP,

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SETIO G, OXFOOO ;SET LOCATION FOR DISPLAY SETIO U, OXFAOO ;SET LOCATION FOR USB DRIVE LOAD "FONT.HEX" ;GET FONT TABLE FROM USB SETGLCD I ;INITIALIZE THE GRAPHICS LCD SETGLCD C ;CLEAR THE DISPLAY SETGLCD D1 ;SELECT LARGE FONT

DO

SETGLCD P,(100,200);SET CURSOR POSTION

; DISPLAY TIME GLCDOUT USING(2),!H,":",!M,":",!S," ",

; DISPLAY DATE GLCDOUT USING 2),!MO,"/",!D,"/",!Y

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■ BY LOUIS E. FRENZEL W5LEF

WHATEVER HAPPENED TO CITIZENS BAND RADIO? Wireless Throwback Or Still A Viable Communication Alternative?

I was talking with some friends about ham radio and some other wireless communications methods recently and someone asked, "Whatever happened to Citizens Band (CB) radio?" One of the younger guys in the group asked "What is CB radio?" Whoa! I guess I forgot how old a technology this is. It certainly is not as visible as it used to be. In any case, I had to explain that CB radio is a two-way radio communications system sanctioned by the Federal Communications Commission (FCC) that does not require a license. Anyone can use it if you buy the licensed and approved equipment. In case you didn't know, CB radio is still around. If you haven't tried it, you just might find it fits one of your communications needs.

CB BACKGROUND

The FCC established a personal communications service back in the late 1940s. Most of the frequency assignments were in the 450 MHz UHF range and it never became popular. In 1958 and in the 1960s, the 27 MHz spectrum was assigned to the newly created Class D radio service. This is what has become known as CB radio. The 27 MHz spectrum (also known as 11 meters) was initially divided into 23 fixed frequency channels. Early transceivers used tubes and had to use two crystals for each channel: one for the transmitter and another for the receiver's local oscillator. Crystals were expensive back then, so many of the lower priced units only covered a few channels.

The 1960s brought about the solid-state movement, so CB radios quickly adopted transistor circuitry making them smaller, less expensive, and easily battery operated. Phase-locked loop (PLL) frequency synthesizers came along making it possible to synthesize all channels from one or two crystals. Later, the FCC opened up more channels, making a total of 40 available. These are the channels used today (see **Table 1**).

Incidentally, while all channels are used, some are

much more popular and useful than others. Channel 9 is an emergency calling frequency. It is not used for regular conversation, but instead is used only when someone needs help. Some police and highway patrol units monitor it on a regular basis. Channel 19 has become a general calling frequency for anyone wanting to establish a

Channel	Frequency (MHz)	Channel	Frequency (MHz)
1	26.965	21	27.215
2	26.975	22	27.225
3	26.985	23	27.255
4	27.005	24	27.235
5	27.015	25	27.245
6	27.025	26	27.265
7	27.035	27	27.275
8	27.055	28	27.285
9	27.065	29	27.295
10	27.075	30	27.305
11	27.085	31	27.315
12	27.105	32	27.325
13	27.115	33	27.335
14	27.125	34	27.345
15	27.135	35	27.355
16	27.155	36	27.365
17	27.165	37	27.375
18	27.175	38	27.385
19	27.185	39	27.395
20	27.205	40	27.405
Table 1			

connection. Usually, those who link up switch to another clear channel.

CB radio really became popular in the 1970s. It was initially adopted by long distance truckers who wanted to know about gasoline sources, speed traps, heavy traffic, accidents, and the like. Back then, there was a gasoline shortage and a 55 MPH highway speed limit that drove the truckers nuts. CB radio really made it easier for all of them. Many other people put CBs in their cars and connected with the truckers on long highway trips.

CB radio became a major hobby. It was for those who wanted to be

amateur radio operators but didn't want to learn the code or electronics to pass the FCC exams. The FCC issued licenses and call signs back then, but dropped both after it could not keep pace with the millions that wanted them. The license requirement and call signs were dropped. That led CB operators to adopt a name or "handle" that could identify them and present a radio personality.

CB was also popularized by many movies and songs back in the 1970s. Movies like *Smokey and the Bandit*, TV shows like *Dukes of Hazzard*, and songs like *Convoy* made CB a real phenomenon. And don't forget all the colorful expressions like "good buddy," "breaker, breaker," plus the 10 codes like 10-4 (received clearly) and 10-20 (location) were extremely popular.

The CB frenzy eventually faded, but it still continued to be popular. Over the years as new technology has come along, CB has now become a secondary communication choice, if it is known at all. There are so many other communications options today, CB has become an anachronism.

<u>CB TECH</u>

CB radios come in two basic form factors: mobile units for under-dash mounting and handheld units. **Figure 1** shows some typical units. The mobile units take their power from the car battery, while the handhelds have an internal rechargeable battery. The mobile units rely on a rather large (six to 10 feet) whip antenna, whereas the handhelds have their own telescoping whip or "rubber ducky" antenna. Some companies still make a basestation unit with a built-in AC power supply.



The basic transmission mode is plain old amplitude modulation (AM). A more sophisticated version uses single sideband (SSB) — a form of AM that uses only one sideband to conserve spectrum space. Maximum output power to the antenna is limited to four watts for AM and 12 watts peak envelope power (PEP) for SSB. The receivers are superheterodynes and all 40 channels are frequency synthesized. Prices run from about \$50 to \$150 for a typical unit, making CB one of the most affordable forms of two-way radio for personal or business use.

As for radio wave propagation, the 11 meter band has some unusual characteristics. It is only good for a one to five mile range in local coverage. However, frequencies in the 27 MHz range use sky waves to communicate over huge distances. Such waves go up to the ionosphere – an ionized region miles above the earth, where they are refracted or bent back to earth many miles away. Such propagation permits worldwide communications even with low power. This makes coast to coast communications possible, not to mention contacts with CB operators in other countries that use those frequencies. However, be forewarned. The FCC Part 95 rules do not permit contacts beyond 155 miles.

THE FRS ALTERNATIVE TO CB

FRS means Family Radio Service – another personal communications service created by the FCC. It uses the 460 MHz UHF spectrum and (like CB) specific channel frequencies are assigned (see **Table 2**).

Channels 8 through 14 are strictly for FRS use, but Channels 1 through 7 are shared with the General Mobile

Channel Number	Frequency (MHz)	
1	462.5625	
2	462.5875	
3	462.6125	
4	462.63.75	
5	462.6625	
6	462.6875	
7	462.7125	
8	467.5625	
9	467.5875	
10	467.6125	
11	467.6375	
12	467.6625	
13	467.6875	
14	467.7125	
Table 2		

Radio Service (GMRS) - alicensed service available to business.

Virtually all FRS radios are handhelds that are battery powered. FRS radios use frequency modulation (FM), so have better noise immunity and fidelity than CB. The transmit power is restricted to 0.5 watts. Furthermore, the antenna must be affixed to the unit. The low power and antenna restrictions limit practical range to less than a mile. With a clear line-of-sight path, several miles are possible. Mountain tops and high buildings make good platforms for long range communications. UHF

signal propagation is LOS and any path blockage by walls, buildings, trees, cars, or whatever severely limits range. The units are best used for short distances indoors or out.

The FRS radios serve a different purpose than CB radios. They are great for just keeping in touch in close situations. Car to car, shopping centers, stadiums, parks, and other large areas make it possible for people to stay in contact. Boating is another use.

You can get a pair of FRS transceiver handhelds for less than \$50 at RadioShack and other electronic dealers.

THE DECLINE OF TWO-WAY RADIO

I just recently got out my RadioShack CB handheld and recharged it. I tried listening on all the channels and got little for my effort. I heard a few weak signals on Channel 11 but nothing on Channel 19. I suspect my small rubber ducky antenna is part of the problem. A longer, higher outdoor antenna would bring in more traffic, I think. The band was essentially deserted. But then again, that is a good thing as it is open for business. Most CB units (like my handheld) have a VHF weather radio built in. I was able to hear the local weather station loud and clear. Check your local RadioShack for both CB and FRS radios.

While CB and FRS radios are still used, their presence is not as great as it once was. Bet you can guess why – cell phones. Pretty much everyone carries one these days and they basically cover our personal communications needs. However, every now and then, a two-way radio makes sense so keep it in mind when you need to communicate. And remember, CB can still be a hobby of sorts. The good thing about these services is that the equipment is cheap and available to play around with. So go ahead and give them a try, good buddies. 10-4. **NV**







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C PROGRAMMING - HARDWARE - PROJECTS

by Joe Pardue

Digital I/O – Part 1

#39

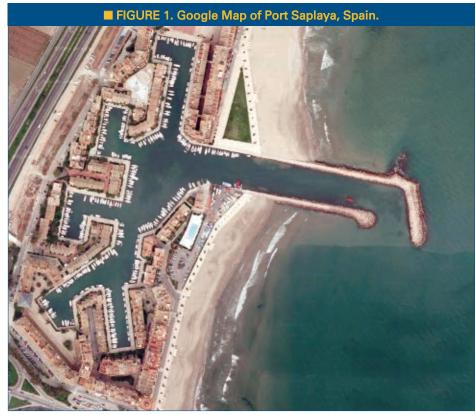


Recap

Last month, we continued looking at tools to put in our avrtoolbox and added the C Standard Library, paying special attention to the venerable printf() functions. We finished by developing a command-line interpreter tool. This month, we are going to dive into digital I/O (Input/Output). I've checked the depth, so you probably won't hit your head on any rocks.

Ports?

If you ever wondered why a group of pins on a microcontroller might be called a port, you need look no



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farther than **Figure 1** – a Google Map of Port Saplaya, Spain. The sea can be a wild and dangerous place where ships batten down and fight their way through storms until they get to the safety of a port. In a port, the waves are broken and the ship is safely tied to a dock where goods can be moved (ported) to and from warehouses and highways.

A microcontroller port is analogous to a seaport in that the pins of a port buffer the sensitive inner workings of the microcontroller from the chaotic outer world, allowing you to safely move digital data to and from the processor core. In practice, we apply the term port to several different kinds of input and output: We have various serial ports such as the USART or SPI, and we

> have the topic of this month's Workshop: digital I/O ports, which are arrays of eight pins considered as a unit.

Why eight pins? This is because the ports are peripherals that are viewed by the AVR processor core as being memory locations, and the AVR addresses memory as locations for eight bits of data. The AVR core knows nothing about the electrical vagaries of the outside world, leaving the conditioning electronics to sort that out. The core just sees a port as a memory location that it can read data from or write data to.

Pins?

The Arduino folks made a radical (okay, that's my opinion) decision to ignore the port concept and present the novice with pins as individually numbered entities independent of any other pin. That is a great simplification for the novice, and maybe that isn't even so radical since the AVR does guarantee that you can change and use any pin of a port without affecting any other pin on that port.

So, let's go with this idea of pins as separate entities and ignore ports for a moment, and present a pin-based Arduino-like elementary digitalio library for our avrtoolbox. This library is written for the ATmega328 (which I tested on an Arduino board) and the AVR Butterfly which uses the ATmega169.

Avrtoolbox Digitalio Library

The avrtoolbox digitalio library – like most of our elementary libraries – looks a lot like something from the Arduino. This is intentional and provides a transition for the novice Arduino user to progress to using the more professional AVR C programming

resources: AVRStudio, WinAVR, avrlibc, avr-gcc, etc. Once you decide you are ready to move beyond the Arduino, you have the elementary avrtoolbox libraries and associated source code to let you see how this is done in C (okay, you get to see how *I* think it ought to be done in C). Don't get me wrong. I love the Arduino and use it all the time, even in 'professional' projects, but it was never intended to be a full-fledged development environment, so I'm trying to provide the tools that will help make the transition easier.

Digitalio Functional Requirements Specification

Digitalio Initialization Function: Allows the user to set a pin as either input or output, with the option of using the output pull-up resistor or not.

Digitalio Input Function: Allows the user to read the digital state of an input pin.

Digitalio Output Function: Allows the user to set the digital state of an output pin.

Digitalio Application Programmer's Interface pin_mode() – Digitalio Initialization Function

Description: Sets the specified pin to either input or output.

Syntax: void pin_mode(uint8_t pin, uint8_t mode)
Parameters:

Pin: Pin to set the mode for. Mode: Either INPUT or OUTPUT. **Returns:** Nothing.

Example: pin_mode(9, OUTPUT);

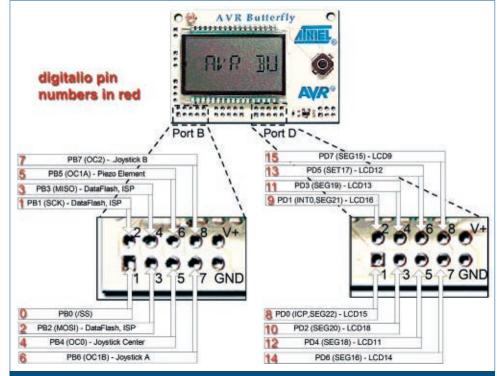


FIGURE 2. Butterfly pin numbers.

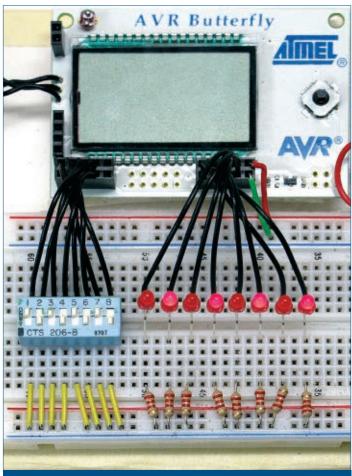
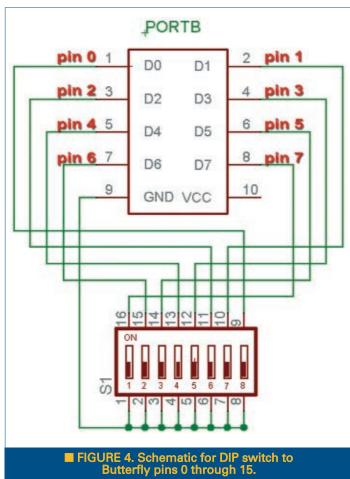


FIGURE 3. Testing digitalio using the Butterfly.



digital_read() - Digitalio Input Function

Description: Reads the state of the indicated input pin.

Syntax: int digital_read(uint8_t pin)

Parameters: Pin: The number of the pin to read. **Returns:** HIGH, LOW, or ERROR.

Example:

// wait until the monkey pushes the

// button then give him some candy

while(digital_read(8) == HIGH); // wait forever for the
button push

digital_write(9,HIGH); // open the candy door

digital_write() - Digitalio Output Function

Description: Sets the state of the indicated output pin.

Syntax: void digital_write(uint8_t pin, uint8_t value) **Parameters:**

Pin: Pin to set either HIGH or LOW.

Value: HIGH or LOW.

Returns: Nothing. **Example:**

if(digital_read(8) == LOW) // did somebody trip the door sensor?

digital_write(7) = HIGH; // if so, turn on the lights

The Source Code

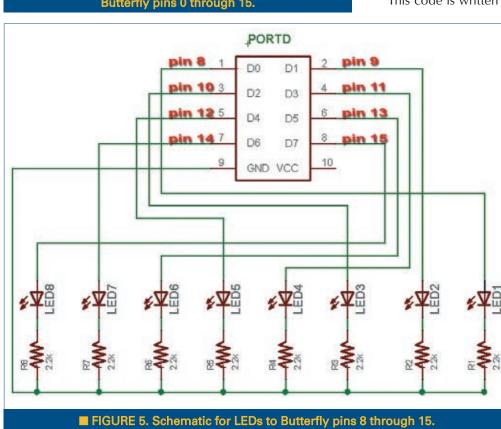
This code is written for both the Butterfly and the

Arduino board using AVRStudio (not the Arduino IDE). If you aren't already familiar with how to do this, you may want to refer to earlier Workshops to review how to compile and upload the code.

The tester is very simple and merely reads the DIP switch and shows the state on the LEDs. Once you build the hardware for this and get it all working, be sure and save it for next month where we will do some more complex things using the same setup.

Using the Digitalio Library With the Butterfly

Since this about input and output, let's dig out that eightbit DIP switch and use it to set the states on eight LEDs. In **Figure 2**, you see the Butterfly pins for ports B and D shown



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as pin numbers to use with the digitalio library. The test program — as usual — is located in the testers directory at **http://code.google.com/p/avrtoolbox/libavr**. The functions are in the /source/elementary/digitalio directory and the tester is in the /testers directory. Rather than use a precompiled object library in digital_c_tester_butterfly, I include the source for the digitalio library functions since these functions are small.

Using the Digitalio Library With the Arduino

Our work is simplified for the Arduino since it already has the pin numbers labeled on the board as shown in **Figure 5** and **6** (if they look vaguely familiar, it is because we used this circuit in Workshop 12). The tester program – like for the Butterfly – is also located in the testers directory at **http://code.google.com/p/avrtoolbox** and is named digitalio_c_tester_atmega328. Remember that this is AVRStudio base code and even though we are using the Arduino hardware, we are not using the Arduino IDE.

Digital I/O Electrical Characteristics

After getting our toes wet with some elementary software, let's go off the high board and look at the hardware. Microcontroller ports (like real ports) are limited in how much protection they can provide and like a seaport, when a hurricane comes along the port can get swamped. In the good old days, you were advised to wear static protection when messing with ICs to prevent static electricity from entering the pin and zapping something important inside the chip. Nowadays, devices like the AVR have input diodes (**Figure 8**) that help protect them.

The first thing to note is the diodes connecting the pin to Vcc and GND. These help protect the pin from things like you walking across a wool carpet on a dry winter's day and then gently touching your micro – 'snap!' – exposing it to a few thousand volts of static electricity. There is also an input capacitance associated with the pin that you might have to deal with in your circuit design,

though it is small and I usually ignore it. Also, there is a pull-up resistor that we can activate to connect the pin to VCC through about 20K ohms – more on this later. **Figure 9** shows more detail.

We know that for something to be digital in our microcontroller binary logic sense it can have two states that can be thought of as true and false, high and low, 1 and 0. From the simplicity of these two states flows the

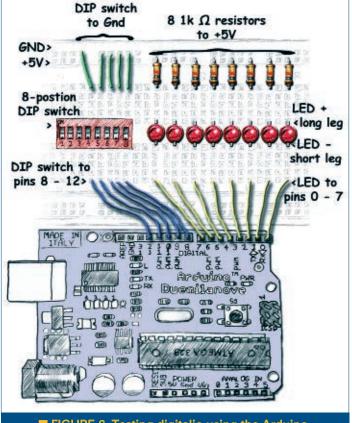
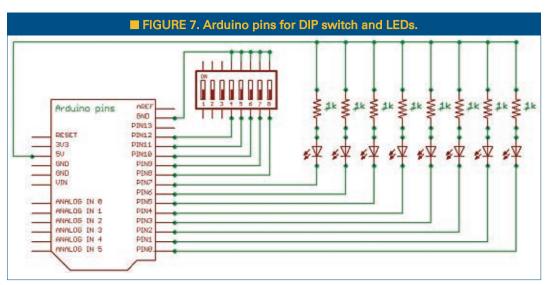
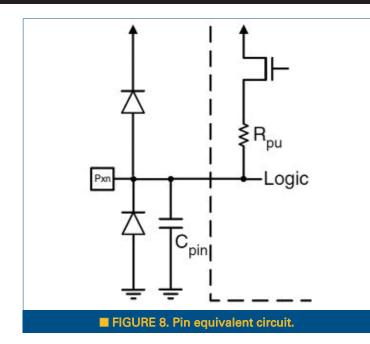


FIGURE 6. Testing digitalio using the Arduino.

entire digital revolution – arguably the most significant change in human history – so it is kind of fun to be able to actually get our hands on these magical 'states' in the form of little silver pins hanging on the sides of our AVR black box. We use digital input when we want to sense an event in the real world such as a button state. We use digital output when we want to change something in the real world such as lighting an LED. [And that is pretty much all we do with microcontrollers: We sense and change things.]

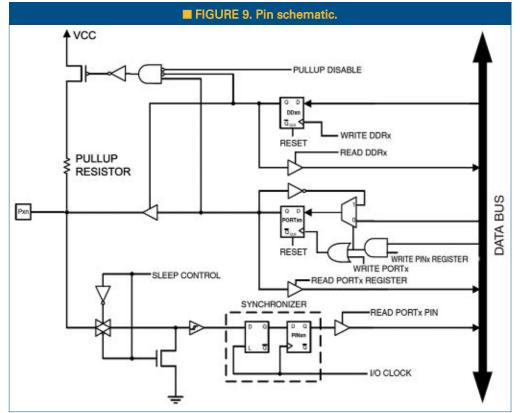




Hysteresis?

In our case, this refers to the concept that we must have above a certain voltage to generate a 1, and below a different lower voltage to generate a 0. For a five volt AVR, we considered any voltage above three volts to be 1, high, or true, and any voltage below two volts to be 0, low, or false. What about voltages between two and three?

If a voltage falls below two, then the pin indicates low



even if the pin immediately rises to 2.1 or 2.9999. It has to go all the way to three to change the pin to high; then if it drops below three, the pin continues to indicate high until it falls below two volts when it changes to low. This is called hysteresis and is a great feature because it allows us to measure signals with some noise on them.

Imagine a signal that is slowly dropping from three to two, but has a bit of \pm 0.2 volts high frequency chaos jumping around. If the pin state changes from high to low and low to high at exactly two volts, then you'd see the pin toggling like crazy, going off and on due to the noise while the real signal (the average of all that noise) we are interested in takes its own good time falling below our threshold plus the noise level. We might read a thousand transitions while the signal voltage falls the extra 0.2 volts noise needed to get completely below the transition threshold.

Logic Input/Output Voltage Levels

In the discussion above, I said that for 5V a value above three was a 1 and below two was 0; that was a simplification to help understand the concept. Looking at the ATmega328 datasheet section 29.8.9 *Pin Threshold and Hysteresis*, we see that for the I/O pin to read 1 when the Vcc is five volts, the input voltage level should be above about 2.6 volts. For the pin to indicate 0 with a Vcc of five volts, then the input voltage should be below about 2.1 volts. So, using five volts we have a

hysteresis of about 0.5 volts. At a Vcc of 3.3V, these values would be above about 1.6 volts for a 1 and below about 1.25 for a 0, with about 0.35 volts hysteresis.

Why do I keep saying 'about?' That is so you won't memorize these figures and think you know how to generate a 1 or 0 on an input pin. Ask yourself: When have you ever gotten exactly five or 3.3 volts for Vcc? What if you are using batteries and you've got 3.2 volts one day and 2.7 a month later? What I'm saying is that you need to consider these thresholds and hysteresis values as ballpark figures for your application. Make sure vou've got some extra voltage if you want to be certain you've got a 1, and that you take it down farther than the indicated low to assure that you are getting a 0.

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The Arduino website [http://arduino.cc/en/ **Reference/Constants**] states that for a digital input high you need to provide greater than 3V, and for a digital input low you need to provide less than 2V. While this is certainly true for the standard 5V Arduino, it isn't true for those Arduinos that use lower voltages such as the 3.3V models. For these, the 3V high will work, but for the low you really need to provide less than 1V. You might think that this would be simple, but look no further than the thread I started on AVRFreaks to see that it isn't:

http://tinyurl.com/3r6aoyu.

If you look at the datasheet section 28.2 DC Characteristics, you get some absolute maximum and minimum electrical parameters to work with. The datasheet Table 28-1 shows the maximum input low:

- For Vcc = 1.8V-2.4V maximum input low is 0.2*Vcc.
- For Vcc = 2.4-5.5V maximum input low is 0.3*Vcc. Minimum input high:
- For Vcc = 1.8V-2.4V minimum input high is 0.7*Vcc.
- For Vcc = 2.4 5.5V minimum input high is 0.6*Vcc.

From which we can calculate:

- For 1.8 volts, the maximum input low is 0.2*1.8V =0.65V.
- For 1.8 volts, the minimum input high is 0.7*1.8V = 1.26V.
- · For 5.5 volts, the maximum input low is 0.3*5V = 1.65V.
- For 5.5 volts, the minimum input high is 0.6*5V = 3.3V.

Okay, this is getting to be way too much so let's promulgate a rule of thumb that applies in all cases: take 0.2*Vcc for low and 0.7*Vcc for high. If this is too restrictive and you want to make the reasonable assumption that your five volt system won't fall below 2.4 volts, then you can use a narrower range of 0.3 Vcc and 0.6 Vcc which for five volts gives you above 0.6*5 = 3V for a high, and below 0.3*5 = 1.5V for a low.

What about digital output? Well, that is a bit clearer but with the complication that it depends on the current it has to provide to the output circuit.

- Output low for 5V at -20 ma is 0.9V or less.
- Output low for 3V at -10 ma is 0.6V or less.
- Output high for 5V at -20 ma is 4.2V or more.

• Output high for 3V at -10 ma is 2.3V or more.

We can see that these digital output high and low values will work with the digital input pins, but if you are using them to drive something other than an AVR compatible digital input, then you might have to take a look at the datasheet.

Questions?

As usual, if you want to be helpful when you find a problem or have a question, you'll need to put on your biohazard suit and start a thread on www.avrfreaks.net with the word 'avrtoolbox' in the title and I probably will see it. Read my blog entry that will tell you why you need the biohazard suit first at http://smileymicros.com/blog/ 2011/01/24/using-an-internet-forum. Next month, we are going to take a deeper look at digital input and output, confronting the datasheet, examining the registers, and learning how to do digital I/O in ordinary C.

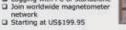
If you just can't wait and want to get a leg up on all this serial stuff and real C programming for the AVR (while helping support your favorite magazine and technical writer), then buy my C Programming book and Butterfly projects kit and/or the Arduino Workshop book and projects kit from the *Nuts & Volts* web shop. **NV**



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ELECTRONICS

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Michael Jay Geier began operating a neighborhood electronics repair service at age eight that was profiled in The Miami News. \$24.95

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Genius gets you started with programming and I/O interfacing right away, and then shows you how to develop a master processor circuit. \$24.95











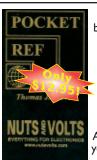




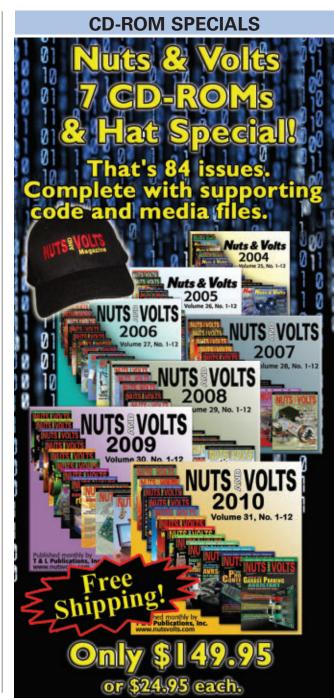
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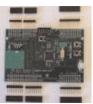
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16-Bit Experimenter Handbook



From the series of articles by Tom Kibalo, The Complete 16-Bit Experimenter Handbook on CD-ROM!

This CD-ROM handbook contains: 200 pages of detailed instructions and 41 experiments with 'C' software examples. Also includes the experiments for the dsPIC33. For a complete list of kits and combos, please visit our webstore. Subscriber's Price **\$64.95** Non-Subscriber's Price **\$69.95**

PROJECTS Piezoelectric Film Speaker Kit

As seen in the November 2010 issue, here is a great project to amaze your friends and to demonstrate a unique way of producing sound. Kit contains one piece of piezoelectric film, speaker film stand,

PCB, components, audio input cable, and construction manual. All you'll need to add is a battery and a sound source. For more info, please visit our website.

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A small power supply with +5V, +12V, and -12V outputs is a handy thing to have around when you're breadboarding circuits with both op-amps and

digital ICs. Kit includes: Enclosure box, accessories, DC-to-DC converter kit, switching regulator kit, and article reprint. For more information, please see the "feature article section" on the of the

Nuts & Volts website. Subscriber's Price **\$76.95** Non-Subscriber's Price **\$83.95**

Transistor Clock Kit



If you like electronic puzzles, then this kit is for you! There are no integrated circuits; all functionality is achieved using discrete transistor-diode logic. The PCB is 10"x11" and harbors more than 1,250 components! For more info, see the November 2009 issue.

Reg \$225.95 Sale Price \$199.95

PCBs can be bought separately.



CHIPINO Kit

The CHIPINO module is an electronic prototyping platform that is used in a series of articles starting with the March 2011 issue of Nuts & Volts Magazine.

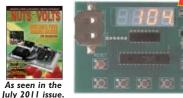
today



Developed by the CHIPAXE Team as a bridge between PICs and Arduinos. The module was designed specifically to match the board outline, mounting holes, connector spacing, and most of the microcontroller I/O functions found on the popular Arduino module. Subscriber's Price \$18.95

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Sorting Counter Kit



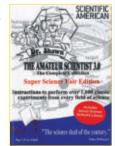
Sorting counters have many uses — keeping score, counting parts, counting people — it is just a handy gadget to have around. This is a very simple project for those who want to learn to solder or are interested in using microprocessors and how they function. No special tools are needed, just a small tip soldering iron. It has no box as it stands alone, therefore there is no drilling.

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READER-TO-READER ECHFORUM

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>>> QUESTIONS

Blinking LED

Could someone come up with a simple schematic for a blinking LED (clear but blinks red)? The item would use an NTE5405 SCR, a high intensity (MSD) LED, and an LSR (light sensitive resistor), all driven by a nine volt battery. The purpose would be for the blinking light to remain out during the day but would come on at night and blink about once every seven seconds. It would then go out during the daytime.

#10111

Cy Fedor via email

PIC16F690

What is the easiest or best way to PWM a PIC16F690? I've seen Chuck Hellebuyck's book, but he uses an external pot to manually do it. I need to do it in software. #10112 Resonance

#10112	Resonance	
	Baraboo, WI	

Basic Compiler

What cheap or free Basic compilers are being used for programming PICs? #10113 Gregg Reber via email

>>> Comments

I just received my September issue and looked at the proposed solution to problem 7113. The solution proposed is quite simple and the analysis of its operation is correct. However, its output is not linear with the value of Rx - as is indicated by the formulas. I believe the request was for a linear response. I have not built the circuit in **Figure 1**, but it looks like it should do what was requested by Brian. U1 must be a rail-to-rail, single supply quad op-amp. A TS914 should work; +V should be between ≈ 12 and +15V.

U1.1 is a current source which I have used and found in a *Linear*

All questions AND answers are submitted by Nuts & Volts readers and are intended to promote the exchange of ideas and provide assistance for solving technical problems. Questions are subject to editing and will be published on a space available basis if deemed suitable by the publisher. Answers are submitted by

Applications handbook by National

Semiconductor. It ensures that the

high when the voltage across Rx is

below the voltage set by VR1, and is

low (ground) when the voltage across

Rx is higher than the voltage set by

gain when the output of U1.2 is at

ground. When the output of U1.2 is

high, the output of U1.3 will be at

Brian wanted +10V when the resist-

zener diode or a 10V reference IC.

the desired maximum resistance value

for the circuit can be derived by

varying any of several resistors. I

would suggest making R8 a 5K poten-

tiometer. Without knowing the actual

resistances Brian wants to measure, I

cannot suggest values for R1-R5.

However, the voltage across Rx should

be kept to < 5V due to the operation

of U1.3 which has a minimum gain of one. If desired, an LED with series

limiting resistor can be added to the

output of U1.2. If the LED circuit is

U1.3 will act as a follower with

U1.4 acts as an inverter since

The +10V can be derived from a

Variable gain to set the voltage for

U1.2 is a comparator. Its output is

voltage across Rx is linear.

VR1.

ground.

ance is low.

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connected to ground, it will indicate when Rx is below the threshold set by VR1. If it is connected to V+, it will indicate when Rx is above the VR1 threshold and is in the linear range of the circuit.

> Larry Cicchinelli - K3PTO via email

>>> ANSWERS [#8111 - August 2011] Backup Alarm

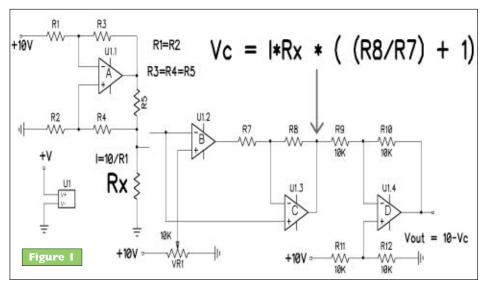
I would like to install a backup alarm on my '07 Mazda Miata with the Mazda jingle (ZOOM ZOOM). I can get the jingle from YouTube but I don't know if it's possible to erase the sounds on a current backup alarm and install the jingle, or build an entirely new alarm.

Get a programmable greeting card at a pharmacy or Walmart, replace the battery with a three volt power supply, and connect the output to a suitable audio amp.

Russell Kincaid Milford, NH

[#8113 - August 2011] Analog S Meter

I have a vintage Realistic DX-400 shortwave radio. The signal meter finally





ANSWERED HERE BY N&V READERS<<<

gave out after all these years. I would like a small circuit to fit in a small confined space, using a 10 bargraph LED. The bargraph should fit perfect in the small window where the needle meter once was.

The LM3914 is made for that application. The datasheet will have a schematic; all you need to do is determine the maximum voltage that is applied to the S meter.

Russell Kincaid Milford, NH

[#9111 - September 2011] **Voltage Problem?**

We would like to plug our credit card terminal into one of our Vodavi Starplus STSe telephone system's SLT ports so that we can use the system's "off-hook preference" feature to find an unused line. However, the credit card terminal needs more power than the 22V provided by the KSU's single line port which drops to 8V when in use. Sometimes the terminal says "no line" and other times it does connect but something inside the terminal is causing a clicking noise – although it does send the DTMF dialing signals in spite of the clicking. We have tried our fax machine on the same KSU port and it works very well. I have tried simply adding a 9V battery in series and it solved the original problem 100% - no clicking, no "no line" message, and the KSU found an unused line just fine. However, the phone company no longer was able to recognize the DTMF dialing.

#1 Being a telecom installer and although I have never worked with the Vodavi STSe, we install credit card machines all the time on IST ports on Comdial and Vertical, and program them for auto group line selection.

Yes, the IST only provides 24 VDC, but does provide about 40 ma loop current. A standard line is 51 VDC on-hook and at 35 ma current, will give you about 6 VDC off-hook. Those off-hook voltages will vary greatly depending on how far you are from the office. Out here where I am (very rural area), I have seen off-hook as low as 3 VDC.

I really am thinking that there is something adrift with your terminal. The terminal should work fine on a 24 VDC on-hook circuit. The voltage is not normally the key; it is the amount of loop current available. Also, it helps to put a P (pause) ahead of the number being dialed. Sometimes when the switch sees off-hook and selects the line, there could be a very short delay. The switch has not yet selected the line when the terminal has already started to dial, hence it will miss sending the first digit or two to the CO.

R. Kauffman Beavertown, PA

#2 I have been in the interconnect business a long time and I know that the voltage from the phone system cannot be increased. It is fixed by specially designed chips, made specifically for the manufacturer of the system. The reason the fax works is it does NOT rely on the line for power. If your credit card machine has an auxiliary power input, that should make it work like the fax.

> **Dennis Hewett** Frontenac, KS

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[#9113 - September 2011] **Increase Resolution**

How do I increase the resolution of a variable resistor (pot)? I have to move the knob so slightly that I can't fine-tune or properly adjust the resistance in the circuit.

There are numerous possible ways to increase the resolution of the potentiometer. Which ones are preferable depends on the external conditions and limitations which are not addressed in the question.

The simplest way is to replace the ordinary potentiometer with a multi-turn potentiometer. If the potentiometer needs to be adjusted often, then this would require a fairly large and relatively expensive unit. If, however, the usage is a "set it and forget it" application, then a "trimmer" potentiometer would be appropriate; then the size and cost would be comparable to an ordinary potentiometer.

Another approach is to divide the potentiometer operation between "coarse" and "fine" potentiometers. Again, several approaches present themselves, depending on whether it's only the resistance that needs to be controlled or if there is a voltage that

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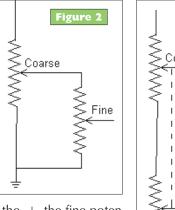
needs to be controlled.

If resistance needs to be controlled, then the circuitry needed is very simple. It is only necessary to wire the two controls in series, with the fine control potentiometer having roughly 1/10 the resistance of the coarse control potentiometer. In this case, the potentiometer should be adjusted by first setting the fine control potentiometer to zero resistance, set the coarse control to have a resistance slightly less than is desired, then finally use the fine control to set the total resistance to the value desired.

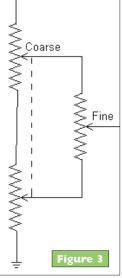
If the voltage needs to be controlled, then the two potentiometers need to be wired together in a slightly more complicated circuit. Figure 2 illustrates one way this can be accomplished. The fine control potentiometer should be enough lower in resistance than the coarse control potentiometer to give the desired amount of expansion; again, a 10-to-1 ratio would seem reasonable. This

configuration is relatively simple but has some disadvantages. First, the total resistance of the combination changes drastically as the coarse control is adjusted. Secondly, as the coarse control is adjusted toward higher settings, the amount of expansion provided by the fine control is reduced. For these reasons, this configuration is practical only if the coarse control is used for the bottom of the range.

A configuration that is free of the disadvantages of the circuit of Figure 2 is shown in Figure 3. This circuit requires two "ganged" potentiometers for the coarse control, i.e., two potentiometers that are adjusted simultaneously with a single knob. Here again, a 10-to-1 ratio of the resistances of the coarse and fine potentiometers would seem to provide a reasonable amount of expansion. The total resistance across the circuit would be very roughly equal to the sum of the resistances of



the fine potentiometer and one of the coarse potentiometers, depending on



the ratio of their resistances. On the other hand, that total resistance would be fairly constant, depending on how well the two ganged potentiometers track each other.

Howard Mark Sufern, NY

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connects to many portable audio sources, MP3s, iPods etc. The on-off switch and volume control are in a 34mm square module attached to the amplifier with a 29" flat cable. The speakers included with the amp are an interesting pair of 1.30" x 1.45" x 0.36" flat-panel speakers incorporating a unique cone-less technology. We don't know for what application these speakers are intended; they might be better for voice or sound effects, but the amplifier sounds a lot better with traditional 4 Ohm cone-type speakers. Operates on 4 AA batteries, not included. CAT# AMP-5

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FAULHABER

#1624E024SNP305. Small, powerful-for-its-size, very-low current, Swiss made gear motor. No-load Ratings: 320 RPM @ 12Vdc/ 2.5mA. 545 RPM @

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LG Innotek Co.# TA09-T3001459AA / Spec MVMF-A346T3. 10mm dia. x 3mm thick. Operates on 1.5 - 4 Vdc. Two 12mm wire leads. Prepped with adhesive patch on one side.

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4 basic functions: CC, CV, CR & CP

8 basic test modes: CCL, CCH, CV, CRL, CRM, CRH, CPV & CPC

Minimum operating voltage is less than 0.6v at the load's full rated current.

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Item# \$49.00 **CSI-Station-3DLF**

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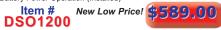
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Multicore Microcontroller Development Board Propeller P8X32A QuickStart

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FEATURES

- 32-bit, 8-core P8X32A microcontroller, up to 160 MIPS
- Buffered USB to serial converter with USB bus power
- Eight resistive touch buttons on I/O pins PO-P7
- Eight blue LEDs on I/O pins P16-P23
- 3.3 V regulator supplies up to 500 mA for projects
- 5 MHz on-board crystal, for DC to 80 MHz operation
- 64 KB EEPROM: 32 KB for code, 32 KB general-purpose
- Pads for sigma-delta A/DC circuitry
- Expansion header to access control signals, Vss, Vdd, and all 32 I/O pins
- Open-source hardware PCB design

KEY SPECIFICATIONS

- Power Requirements: 3.3 or 4 to 9 VDC, up to 500 mA
- Interface: USB, 3.3 V serial, I²C, or others as programmed
- Operating temperature: -40 to +185 °F (-40 to +85 °C)
- Dimensions: 2.0 x 3.0 x 0.36 in (5.0 x 7.6 x 0.84 cm)





The P8X32A QuickStart development board is an open-source hardware project. The bill of materials and the DipTrace schematic and PCB layout files are released under the Creative Commons Attribution 3.0 license. The example code is released under the MIT License. All of these files are available from the P8X32A QuickStart product page at **www.parallax.com**.

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