

- **150-mA Load Current Without External Power Transistor**
- **Adjustable Current-Limiting Capability**
- **Input Voltages up to 40 V**
- **Output Adjustable From 2 V to 37 V**
- **Direct Replacement for Fairchild μA723C**

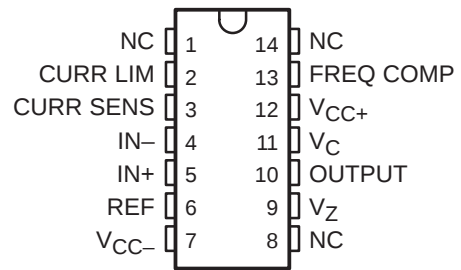
description

The μA723 is a precision integrated-circuit voltage regulator, featuring high ripple rejection, excellent input and load regulation, excellent temperature stability, and low standby current. The circuit consists of a temperature-compensated reference-voltage amplifier, an error amplifier, a 150-mA output transistor, and an adjustable-output current limiter.

The μA723 is designed for use in positive or negative power supplies as a series, shunt, switching, or floating regulator. For output currents exceeding 150 mA, additional pass elements can be connected as shown in Figures 4 and 5.

The μA723C is characterized for operation from 0°C to 70°C.

**D OR N PACKAGE
(TOP VIEW)**

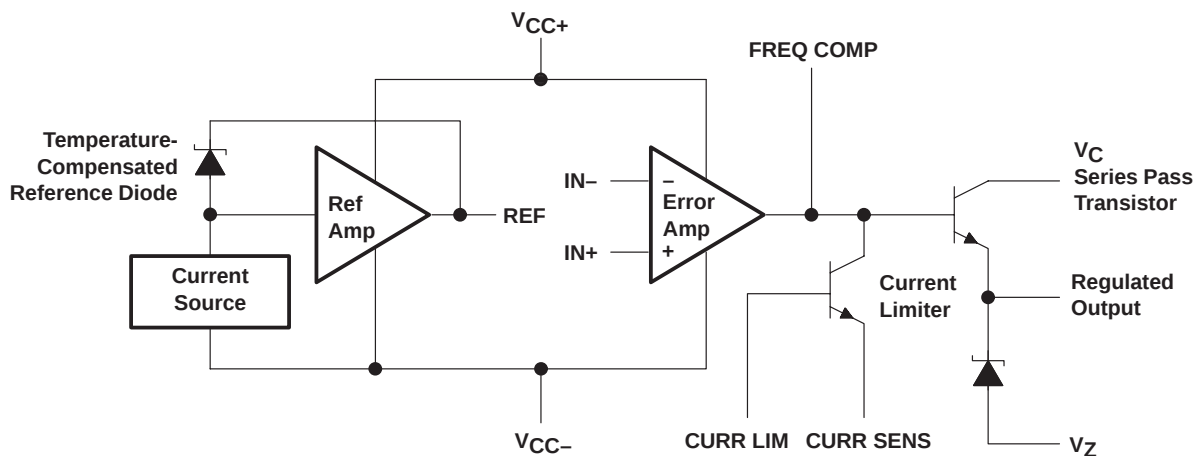


AVAILABLE OPTIONS

T _A	PACKAGED DEVICES		CHIP FORM (Y)
	PLASTIC DIP (N)	SMALL OUTLINE (D)	
0°C to 70°C	μA723CN	μA723CD	μA723Y

The D package is available taped and reeled. Add the suffix R to the device type (e.g., μA723CDR). Chip forms are tested at 25°C.

functional block diagram



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

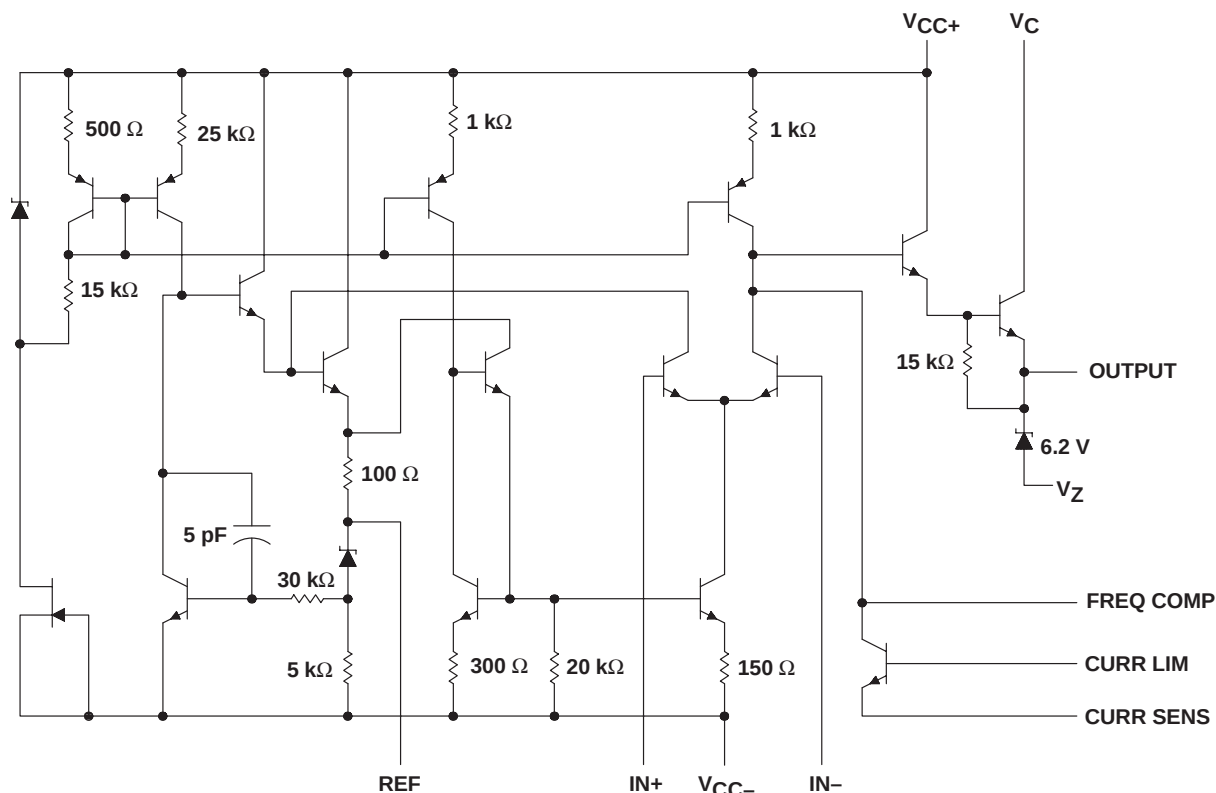
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μA723 PRECISION VOLTAGE REGULATORS

SLVS057D – AUGUST 1972 – REVISED JULY 1999

schematic



Resistor and capacitor values shown are nominal.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Peak voltage from V_{CC+} to V_{CC-} ($t_w \leq 50$ ms)	50 V
Continuous voltage from V_{CC+} to V_{CC-}	40 V
Input-to-output voltage differential	40 V
Differential input voltage to error amplifier	± 5 V
Voltage between noninverting input and V_{CC-}	8 V
Current from V_Z	25 mA
Current from REF	15 mA
Package thermal impedance, θ_{JA} (see Notes 1 and 2): D package	86°C/W
N package	101°C/W
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Storage temperature range, T_{stg}	-65°C to 150°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Maximum power dissipation is a function of $T_J(\text{max})$, θ_{JA} , and T_A . The maximum allowable power dissipation at any allowable ambient temperature is $P_D = (T_J(\text{max}) - T_A)/\theta_{JA}$. Operating at the absolute maximum T_J of 150°C can impact reliability.
2. The package thermal impedance is calculated in accordance with JEDEC 51, except for through-hole packages, which use a trace length of zero.

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, V_I	9.5	40	V
Output voltage, V_O	2	37	V
Input-to-output voltage differential, $V_C - V_O$	3	38	V
Output current, I_O		150	mA
Operating free-air temperature range, T_A	μA723C		0 70 °C

electrical characteristics at specified free-air temperature (see Notes 3 and 4)

PARAMETER	TEST CONDITIONS	T_A	μA723C			UNIT
			MIN	TYP	MAX	
Input regulation	$V_I = 12\text{ V to }V_I = 15\text{ V}$	25°C		0.1	1	mV/V
	$V_I = 12\text{ V to }V_I = 40\text{ V}$	25°C		1	5	
	$V_I = 12\text{ V to }V_I = 15\text{ V}$	0°C to 70°C			3	
Ripple rejection	f = 50 Hz to 10 kHz, $C_{ref} = 0$	25°C		74		dB
	f = 50 Hz to 10 kHz, $C_{ref} = 5\text{ }\mu\text{F}$	25°C		86		
Output regulation		25°C		-0.3	-2	mV/V
		0°C to 70°C			-6	
Reference voltage, V_{ref}		25°C	6.8	7.15	7.5	V
Standby current	$V_I = 30\text{ V}, I_O = 0$	25°C		2.3	4	mA
Temperature coefficient of output voltage		0°C to 70°C		0.003	0.015	%/°C
Short-circuit output current	$R_{SC} = 10\text{ }\Omega, V_O = 0$	25°C		65		mA
Output noise voltage	BW = 100 Hz to 10 kHz, $C_{ref} = 0$	25°C		20		μV
	BW = 100 Hz to 10 kHz, $C_{ref} = 5\text{ }\mu\text{F}$	25°C		2.5		

- NOTES: 3. For all values in this table, the device is connected as shown in Figure 1 with the divider resistance as seen by the error amplifier $\leq 10\text{ k}\Omega$. Unless otherwise specified, $V_I = V_{CC+} = V_C = 12\text{ V}$, $V_{CC-} = 0$, $V_O = 5\text{ V}$, $I_O = 1\text{ mA}$, $R_{SC} = 0$, and $C_{ref} = 0$.
4. Pulse-testing techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

electrical characteristics, $T_A = 25^\circ\text{C}$ (see Notes 3 and 4)

PARAMETER	TEST CONDITIONS	μA723Y			UNIT
		MIN	TYP	MAX	
Input regulation	$V_I = 12\text{ V to }V_I = 15\text{ V}$		0.1		mV/V
	$V_I = 12\text{ V to }V_I = 40\text{ V}$		1		
Ripple rejection	f = 50 Hz to 10 kHz, $C_{ref} = 0$		74		dB
	f = 50 Hz to 10 kHz, $C_{ref} = 5\text{ }\mu\text{F}$		86		
Output regulation			-0.3		mV/V
Reference voltage, V_{ref}			7.15		V
Standby current	$V_I = 30\text{ V}, I_O = 0$		2.3		mA
Short-circuit output current	$R_{SC} = 10\text{ }\Omega, V_O = 0$		65		mA
Output noise voltage	BW = 100 Hz to 10 kHz, $C_{ref} = 0$		20		μV
	BW = 100 Hz to 10 kHz, $C_{ref} = 5\text{ }\mu\text{F}$		2.5		

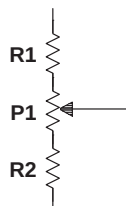
- NOTES: 3. For all values in this table, the device is connected as shown in Figure 1 with the divider resistance as seen by the error amplifier $\leq 10\text{ k}\Omega$. Unless otherwise specified, $V_I = V_{CC+} = V_C = 12\text{ V}$, $V_{CC-} = 0$, $V_O = 5\text{ V}$, $I_O = 1\text{ mA}$, $R_{SC} = 0$, and $C_{ref} = 0$.
4. Pulse-testing techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

APPLICATION INFORMATION

Table 1. Resistor Values (kΩ) for Standard Output Voltages

OUTPUT VOLTAGE (V)	APPLICABLE FIGURES (SEE NOTE 5)	FIXED OUTPUT ±5%		OUTPUT ADJUSTABLE ±10% (SEE NOTE 6)		
		R1 (kΩ)	R2 (kΩ)	R1 (kΩ)	P1 (kΩ)	P2 (kΩ)
3.0	1, 5, 6, 9, 11, 12 (4)	4.12	3.01	1.8	0.5	1.2
3.6	1, 5, 6, 9, 11, 12 (4)	3.57	3.65	1.5	0.5	1.5
5.0	1, 5, 6, 9, 11, 12 (4)	2.15	4.99	0.75	0.5	2.2
6.0	1, 5, 6, 9, 11, 12 (4)	1.15	6.04	0.5	0.5	2.7
9.0	2, 4, (5, 6, 9, 12)	1.87	7.15	0.75	1.0	2.7
12	2, 4, (5, 6, 9, 12)	4.87	7.15	2.0	1.0	3.0
15	2, 4, (5, 6, 9, 12)	7.87	7.15	3.3	1.0	3.0
28	2, 4, (5, 6, 9, 12)	21.0	7.15	5.6	1.0	2.0
45	7	3.57	48.7	2.2	10	39
75	7	3.57	78.7	2.2	10	68
100	7	3.57	105	2.2	10	91
250	7	3.57	255	2.2	10	240
–6 (see Note 7)	3, 10	3.57	2.43	1.2	0.5	0.75
–9	3, 10	3.48	5.36	1.2	0.5	2.0
–12	3, 10	3.57	8.45	1.2	0.5	3.3
–15	3, 10	3.57	11.5	1.2	0.5	4.3
–28	3, 10	3.57	24.3	1.2	0.5	10
–45	8	3.57	41.2	2.2	10	33
–100	8	3.57	95.3	2.2	10	91
–250	8	3.57	249	2.2	10	240

- NOTES: 5. The R1/R2 divider can be across either V_O or $V_{(ref)}$. If the divider is across $V_{(ref)}$, use the figure numbers without parentheses. If the divider is across V_O , use the figure numbers in parentheses.
6. To make the voltage adjustable, the R1/R2 divider shown in the figures must be replaced by the divider shown below.



Adjustable Output Circuit

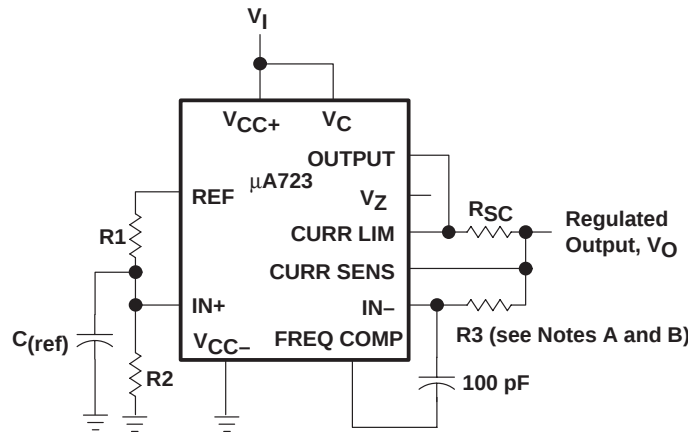
7. For Figures 3, 8, and 10, the device requires a minimum of 9 V between V_{CC+} and V_{CC-} when V_O is equal to or more positive than –9 V.

APPLICATION INFORMATION

Table 2. Formulas for Intermediate Output Voltages

OUTPUTS FROM 2 V TO 7 V SEE FIGURES 1, 5, 6, 9, 11, 12 (4) AND NOTE 5	OUTPUTS FROM 4 V TO 250 V SEE FIGURE 7 AND NOTE 5	CURRENT LIMITING
$V_O = V_{(ref)} \times \frac{R2}{R1 + R2}$	$V_O = \frac{V_{(ref)}}{2} \times \frac{R2 - R1}{R1}$ $R3 = R4$	$I_{(limit)} \approx \frac{0.65 \text{ V}}{R_{SC}}$
OUTPUTS FROM 7 V TO 37 V SEE FIGURES 2, 4, (5, 6, 9, 11, 12) AND NOTE 5	OUTPUTS FROM -6 V TO -250 V SEE FIGURES 3, 8, 10 AND NOTES 5 AND 7	FOLDBACK CURRENT LIMITING SEE FIGURE 6
$V_O = V_{(ref)} \times \frac{R1 + R2}{R2}$	$V_O = -\frac{V_{(ref)}}{2} \times \frac{R1 + R2}{R1}$ $R3 = R4$	$I_{(knee)} \approx \frac{V_O R3 + (R3 + R4) 0.65 \text{ V}}{R_{SC} R4}$ $I_{OS} \approx \frac{0.65 \text{ V}}{R_{SC}} \times \frac{R3 + R4}{R4}$

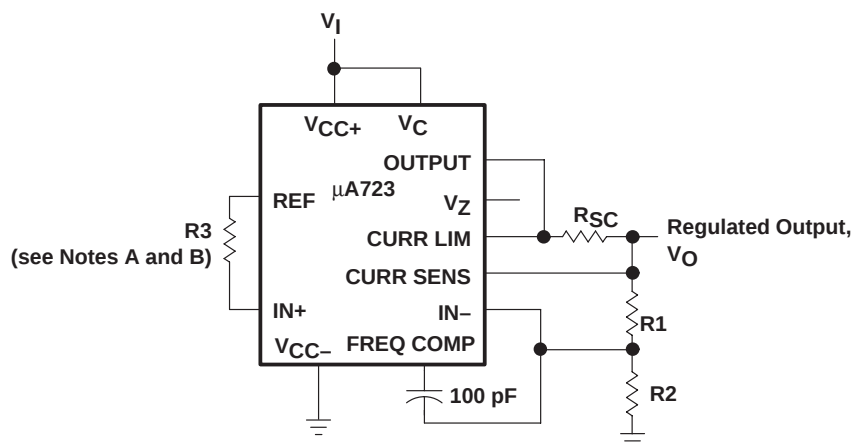
- NOTES: 5. The R1/R2 divider can be across either V_O or $V_{(ref)}$. If the divider is across $V_{(ref)}$, use figure numbers without parentheses. If the divider is across V_O , use the figure numbers in parentheses.
7. For Figures 3, 8, and 10, the device requires a minimum of 9 V between V_{CC+} and V_{CC-} when V_O is equal to or more positive than -9 V.



- NOTES: A. $R3 = \frac{R1 \times R2}{R1 + R2}$ for a minimum α_{V_O}
- B. $R3$ can be eliminated for minimum component count. Use direct connection (i.e., $R3 = 0$).

Figure 1. Basic Low-Voltage Regulator ($V_O = 2 \text{ V to } 7 \text{ V}$)

APPLICATION INFORMATION



- NOTES: A. $R_3 = \frac{R_1 \times R_2}{R_1 + R_2}$ for a minimum α_{V_O}
 B. R_3 can be eliminated for minimum component count. Use direct connection (i.e., $R_3 = 0$).

Figure 2. Basic High-Voltage Regulator ($V_O = 7\text{ V to }37\text{ V}$)

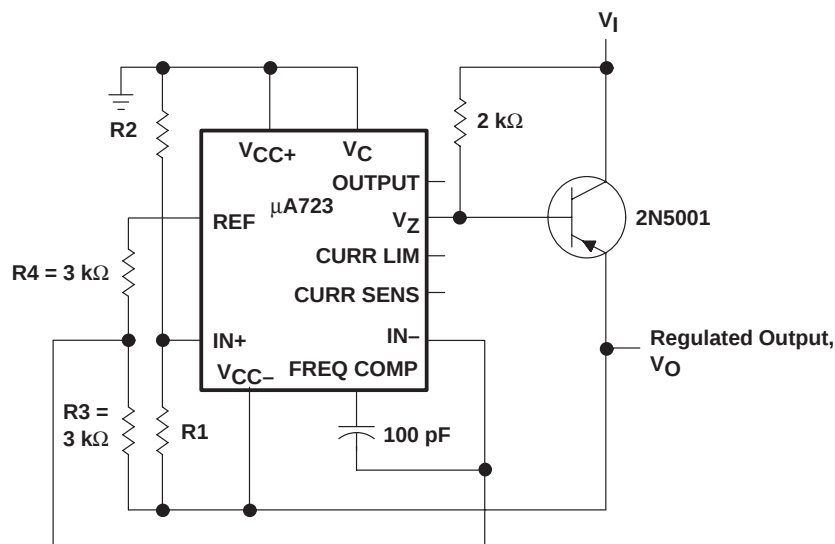


Figure 3. Negative-Voltage Regulator

APPLICATION INFORMATION

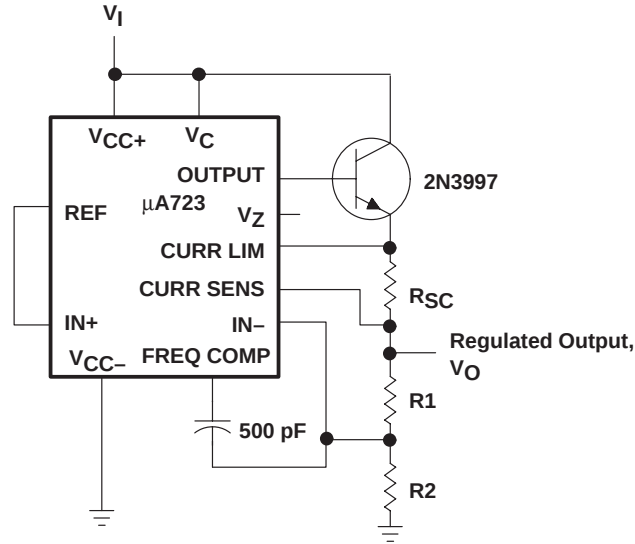


Figure 4. Positive-Voltage Regulator (External npn Pass Transistor)

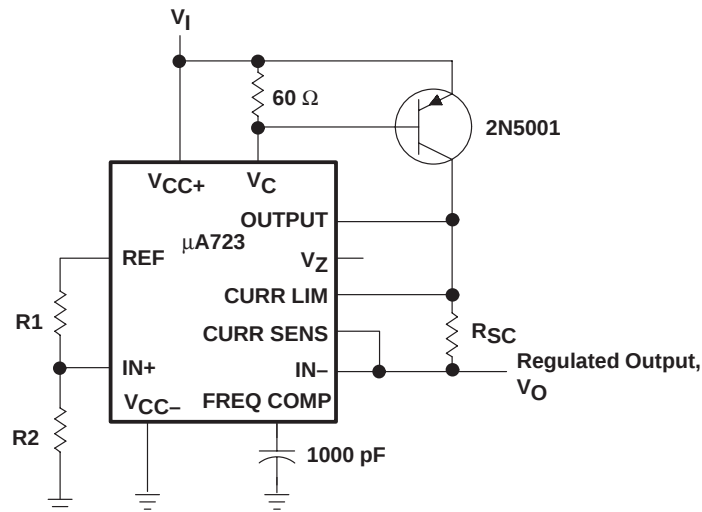


Figure 5. Positive-Voltage Regulator (External pnp Pass Transistor)

APPLICATION INFORMATION

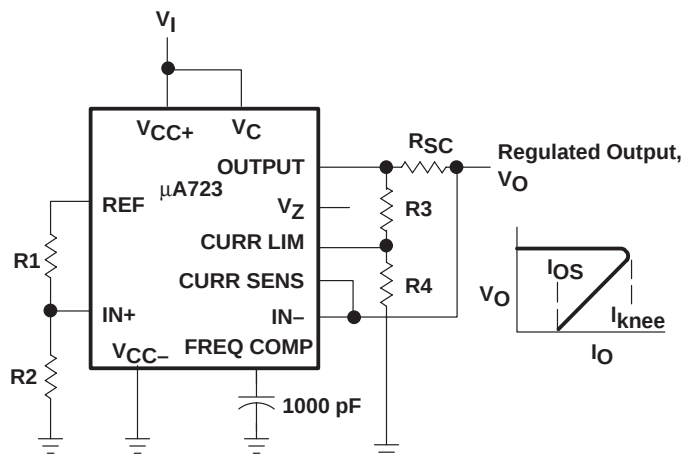


Figure 6. Foldback Current Limiting

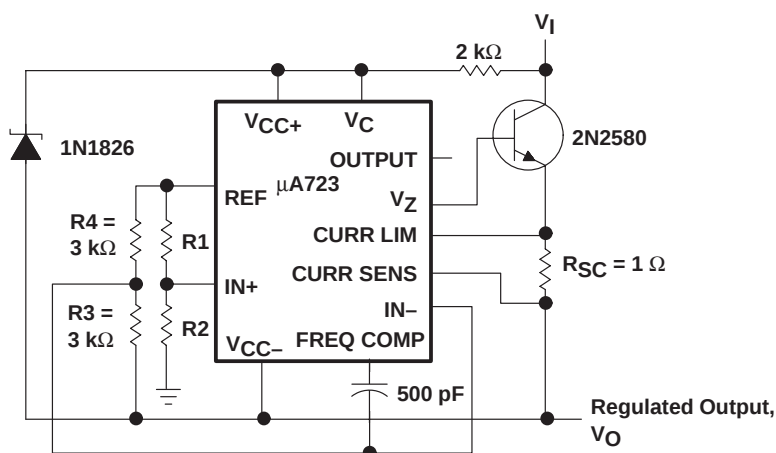


Figure 7. Positive Floating Regulator

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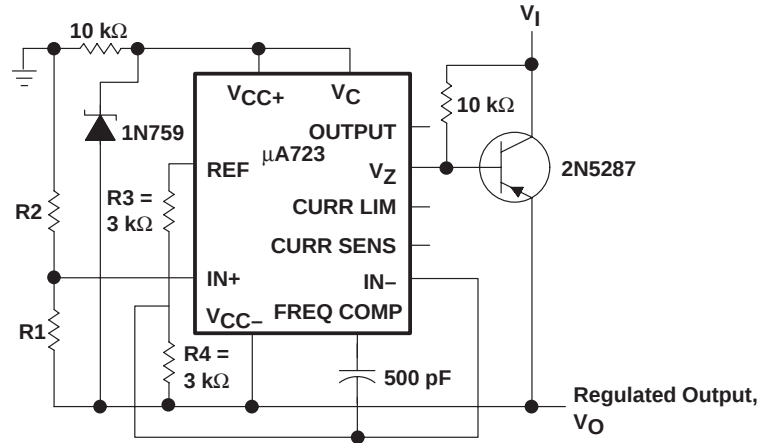
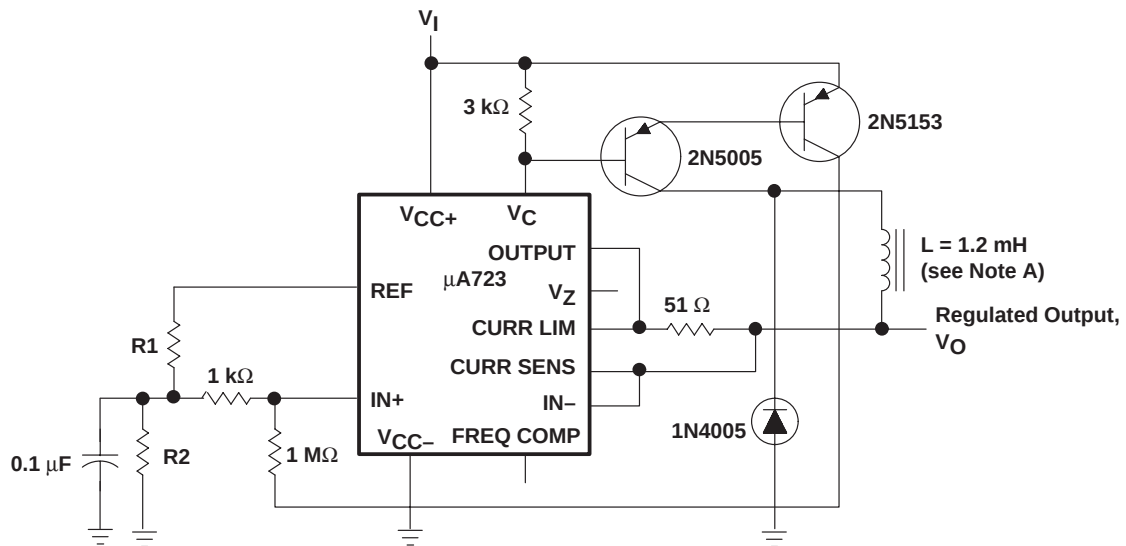


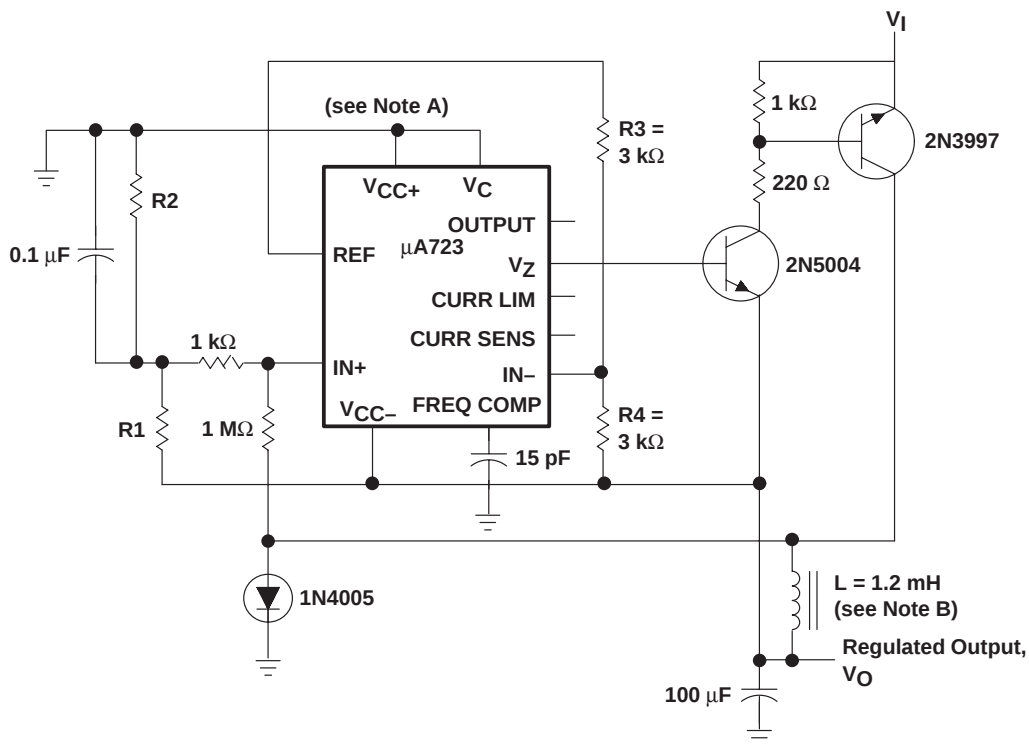
Figure 8. Negative Floating Regulator



NOTE A: L is 40 turns of No. 20 enameled copper wire wound on Ferroxcube P36/22-3B7 potted core, or equivalent, with a 0.009-inch air gap.

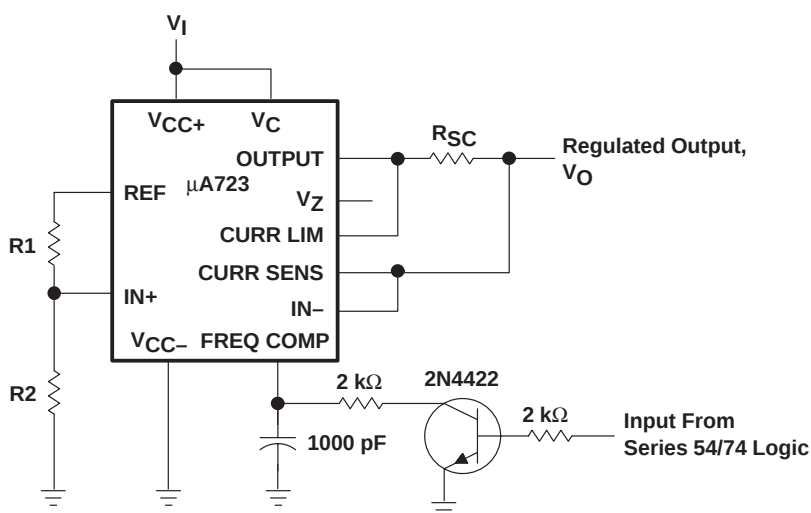
Figure 9. Positive Switching Regulator

APPLICATION INFORMATION



- NOTES: A. The device requires a minimum of 9 V between V_{CC+} and V_{CC-} when V_O is equal to or more positive than -9 V.
B. L is 40 turns of No. 20 enameled copper wire wound on Ferroxcube P36/22-3B7 potted core, or equivalent, with a 0.009-inch air gap.

Figure 10. Negative Switching Regulator



NOTE A: A current-limiting transistor can be used for shutdown if current limiting is not required.

Figure 11. Remote Shutdown Regulator With Current Limiting

APPLICATION INFORMATION

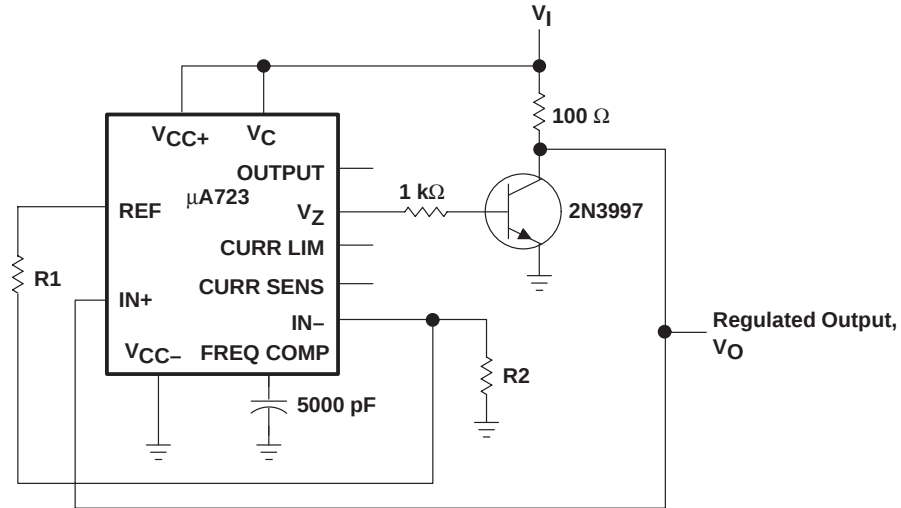


Figure 12. Shunt Regulator

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
UA723CD	ACTIVE	SOIC	D	14	50	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1 YEAR/ Level-1-235C-UNLIM
UA723CDR	ACTIVE	SOIC	D	14	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1 YEAR/ Level-1-235C-UNLIM
UA723CJ	OBSOLETE	CDIP	J	14		None	Call TI	Call TI
UA723CN	ACTIVE	PDIP	N	14	25	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
UA723CNSR	ACTIVE	SO	NS	14	2000	Pb-Free (RoHS)	CU NIPDAU	Level-2-260C-1 YEAR/ Level-1-235C-UNLIM

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - May not be currently available - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

None: Not yet available Lead (Pb-Free).

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Green (RoHS & no Sb/Br): TI defines "Green" to mean "Pb-Free" and in addition, uses package materials that do not contain halogens, including bromine (Br) or antimony (Sb) above 0.1% of total product weight.

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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J (R-GDIP-T**)

14 LEADS SHOWN

CERAMIC DUAL IN-LINE PACKAGE



PINS ** DIM	14	16	18	20
A	0.300 (7,62) BSC	0.300 (7,62) BSC	0.300 (7,62) BSC	0.300 (7,62) BSC
B MAX	0.785 (19,94)	.840 (21,34)	0.960 (24,38)	1.060 (26,92)
B MIN	—	—	—	—
C MAX	0.300 (7,62)	0.300 (7,62)	0.310 (7,87)	0.300 (7,62)
C MIN	0.245 (6,22)	0.245 (6,22)	0.220 (5,59)	0.245 (6,22)



4040083/F 03/03

- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. This package is hermetically sealed with a ceramic lid using glass frit.
 - D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.
 - E. Falls within MIL STD 1835 GDIP1-T14, GDIP1-T16, GDIP1-T18 and GDIP1-T20.

N (R-PDIP-T**)

16 PINS SHOWN

PLASTIC DUAL-IN-LINE PACKAGE



PINS ** DIM	14	16	18	20
A MAX	0.775 (19,69)	0.775 (19,69)	0.920 (23,37)	1.060 (26,92)
A MIN	0.745 (18,92)	0.745 (18,92)	0.850 (21,59)	0.940 (23,88)
MS-001 VARIATION	AA	BB	AC	AD



4040049/E 12/2002

NOTES:

- A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
-  Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
 The 20 pin end lead shoulder width is a vendor option, either half or full width.

D (R-PDSO-G14)

PLASTIC SMALL-OUTLINE PACKAGE



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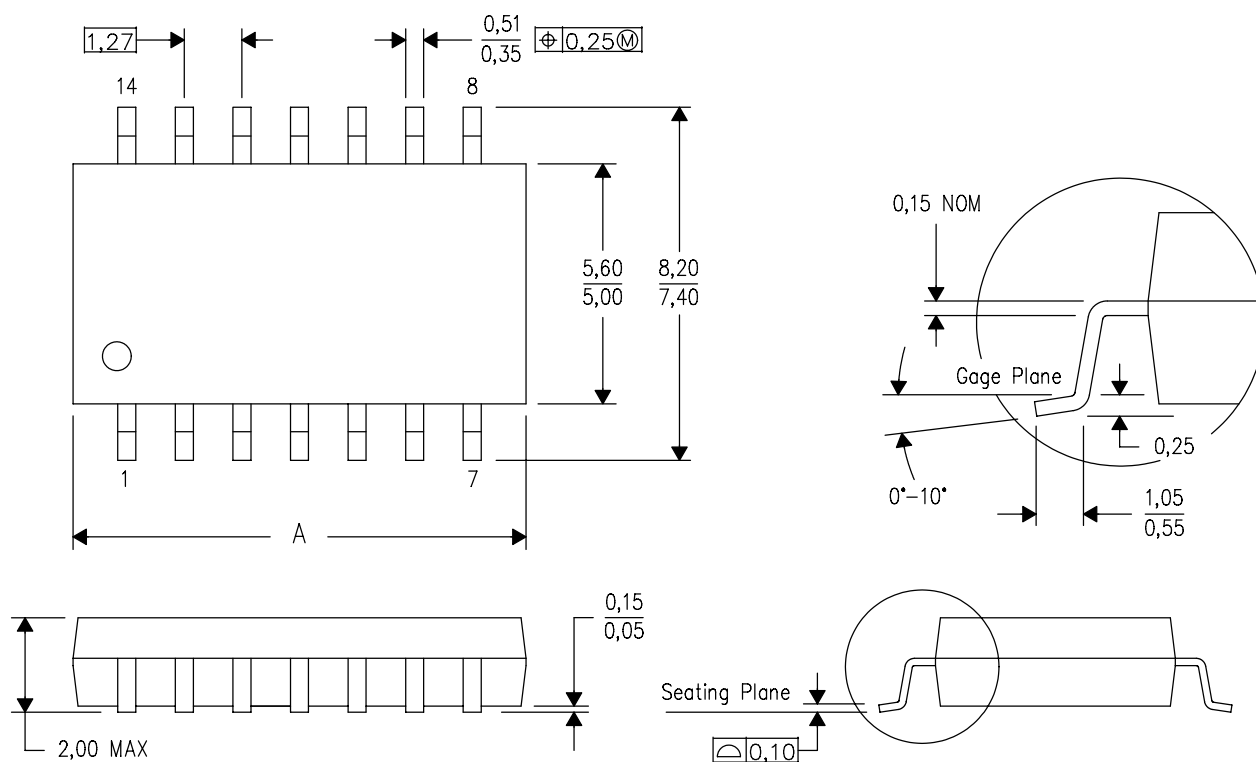
- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
 - D. Falls within JEDEC MS-012 variation AB.

MECHANICAL DATA

NS (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14-PINS SHOWN



DIM \ PINS **	14	16	20	24
A MAX	10,50	10,50	12,90	15,30
A MIN	9,90	9,90	12,30	14,70

4040062/C 03/03

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

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Mailing Address: Texas Instruments
Post Office Box 655303 Dallas, Texas 75265

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